## Interim Channel Models for G2 MMDS Fixed Wireless Applications

Document Number:

802.16.3p-00/49

Date Submitted:

The date the document is contributed, in the format 2000-11-07

### Source:

K. V. S. Hari	Voice:	650-724-3640
Stanford University	Fax:	650-723-8473
Department of EE	mailto: har	@rascals.stanford.edu
Packard #223		
Stanford University		
Stanford, CA 94305-9510		

Venue:

IEEE 802 plenary meeting, November 6-9, 2000 at Tampa, Florida

Base Document:

802.16.3c-00/49r2 URL http://ieee802.org/16/groups/dot16web/sub11/contrib/802163c-00\_49r2.pdf

Purpose:

This is to present work by Stanford University on MMDS channel models with requirements from Sprint.

#### Notice:

This document has been prepared to assist IEEE 802.16. It is offered as a basis for discussion and is not binding on the contributing individual(s) or organization(s). The material in this document is subject to change in form and content after further study. The contributor(s) reserve(s) the right to add, amend or withdraw material contained herein.

#### Release:

The contributor grants a free, irrevocable license to the IEEE to incorporate text contained in this contribution, and any modifications thereof, in the creation of an IEEE Standards publication; to copyright in the IEEE's name any IEEE Standards publication even though it may include portions of this contribution; and at the IEEE's sole discretion to permit others to reproduce in whole or in part the resulting IEEE Standards publication. The contributor also acknowledges and accepts that this contribution may be made public by IEEE 802.16.

#### IEEE 802.16 Patent Policy:

The contributor is familiar with the IEEE 802.16 Patent Policy and Procedures (Version 1.0)  $< \underline{http://ieee802.org/16/ipr/patents/policy.html>}$ , including the statement "IEEE standards may include the known use of patent(s), including patent applications, if there is technical justification in the opinion of the standards-developing committee and provided the IEEE receives assurance from the patent holder that it will license applicants under reasonable terms and conditions for the purpose of implementing the standard."

Early disclosure to the Working Group of patent information that might be relevant to the standard is essential to reduce the possibility for delays in the development process and increase the likelihood that the draft publication will be approved for publication. Please notify the Chair <<u>mailto:r.b.marks@iece.org</u>> as early as possible, in written or electronic form, of any patents (granted or under application) that may cover technology that is under consideration by or has been approved by IEEE 802.16. The Chair will disclose this notification via the IEEE 802.16 web site <<u>http://ieee802.org/16/ipr/patents/letters></u>.

## Interim Channel Models for G2 MMDS Fixed Wireless Applications

K. V. S. Hari, Stanford University Khurram P. Sheikh, Sprint BWG Carl Bushue, Sprint CTL

## G2 MMDS Scenario\*

- Cells are < 4 miles in radius,
- Under-the-eave/window broadbeam directional antennas (8-15ft) at the CPE
- 50-120ft BTS antennas.

\* G2 MMDS:Generation2 Multi-channel Multi-point Distribution System

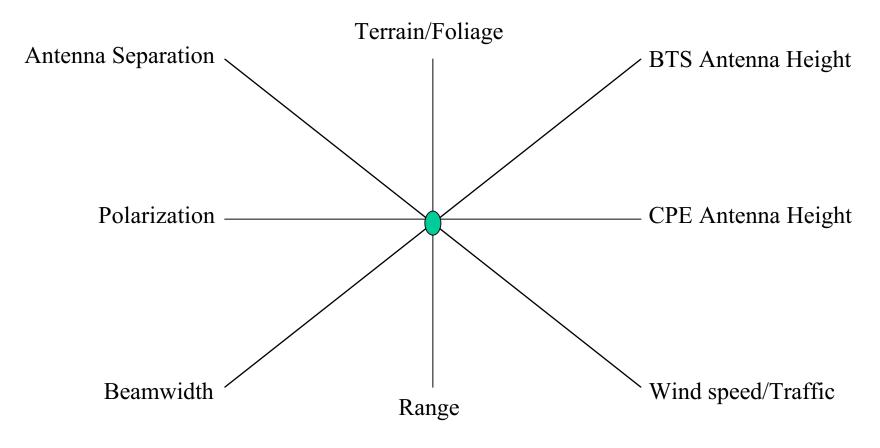
## Impairments

- Higher Path Loss (as compared to super cell architecture)
- Fading: Macroscopic (due to shadowing) and Microscopic (due to multipath).
- Co-channel and Adjacent channel Interference
- More severe multipath delay spread
- Doppler spread

## **Channel Model**

- Channel Model shall describe
  - Path Loss
  - Multipath Fading and correlation
  - Interference

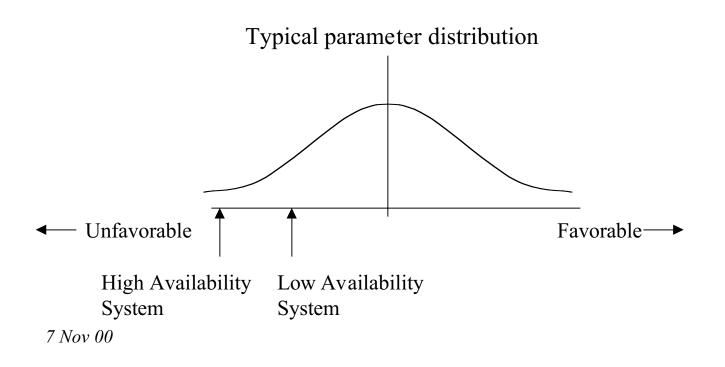
## Channel Has Many Dimensions



## **Channel Model Variability**

Channel parameters are RANDOM quantities We need Statistical characterization

- Cumulative Distribution Function (CDF)



## Path Loss Model

- Hata, COST 231 Hata path loss models are for very high BTS and are not suitable for G2 MMDS scenario
- Erceg Model is the appropriate model

## G2 MMDS Path Loss Model Median Path Loss (Erceg model):

$$PL(dB) = A + 10\gamma \log_{10}(d / d_0) + s + \Delta PL_f + \Delta PL_h$$
  
for  $d > d_0$ 

## where

$$A = 20\log_{10}(4\pi d_0/\lambda)$$
 (free space path loss)

$$\gamma = \overline{a} - bh_b + \frac{c}{h_b} \sqrt{10} \text{ meters} < h_b < 80 \text{ meters}$$
(mean path loss exponent)

 $\lambda$  is the wavelength

7 Nov 00

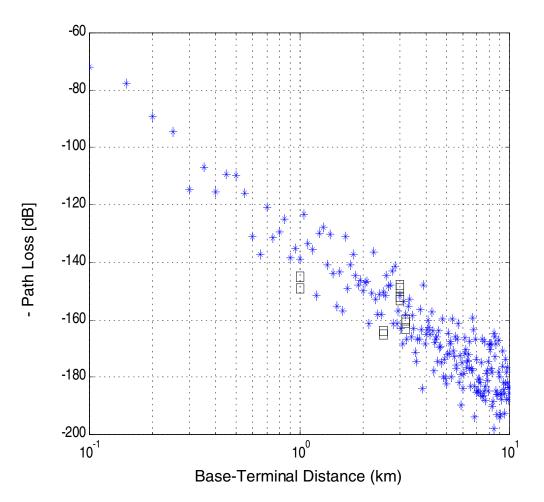
# Path Loss Model (contd.)

- *s* is a lognormal shadow fading
  - zero mean
  - terrain dependent standard deviation
- $h_b$  is the BTS height in meters
- *a*, *b*, *c* are constants dependent on the terrain category
- $d_o$  is chosen as 100m (reference distance)
- *d* is the distance from BTS

## **Correction Terms**

- Frequency correction terms  $\Delta PL_f = 5.7 \log \frac{f}{2000} \sqrt{f \text{ in MHz}}$
- CPE height correction term (> 2 meters)  $\Delta PL_{h} = -10.8 \log(\frac{h_{CPE}}{2}) 1 \text{ meter} < h_{CPE} < 8 \text{ meters}$

## Path Loss Scatter Plot



<sup>D</sup> SU Measurements

\* From Erceg Model

7 Nov 00

## K-Factor

- K-Factor is the ratio of power in the fixed component to the power in the variable component
- It depends upon
  - BTS, CPE heights and beamwidths
  - Distance from the antenna
  - Scattering environment
  - Wind, traffic, season

## K-factor Model

- Erceg model for K-factor  $K = F_s F_h F_b K_o d^{\gamma} u$
- $F_s$  is a seasonal factor
  - 1.0; summer (leaves)
  - -2.5; winter (no leaves)
- *F<sub>h</sub>* is the height factor
  - $-(h/3)^{0.46}$  (h is the CPE height in meters)

## K-factor Model (contd.)

- $F_b$  is the beamwidth factor - $F_b = (b/10)^{-0.62}$ ; (b in degrees)
- $K_o$  and  $\gamma$  are regression coefficients -  $K_o = 10$ ;  $\gamma = -0.5$
- *u* is a lognormal variable
  - zero mean
  - std. deviation of 8.0 dB

## K-factor and Reliability

- K-factors are highly variable
- To ensure 99.9% reliability, systems must be designed for zero K-factor (Rayleigh fading)

## **Delay Spread Model**

• Spike-Plus-Exponential Model (Erceg)

$$P(\tau) = A\delta(\tau) + B \sum_{i=0}^{\infty} e^{-i\Delta\tau/\tau_o} \delta(\tau - i\Delta\tau)$$
A, B,  $\tau_o$  and  $\Delta\tau$  are experimentally determined
$$\Delta\tau$$

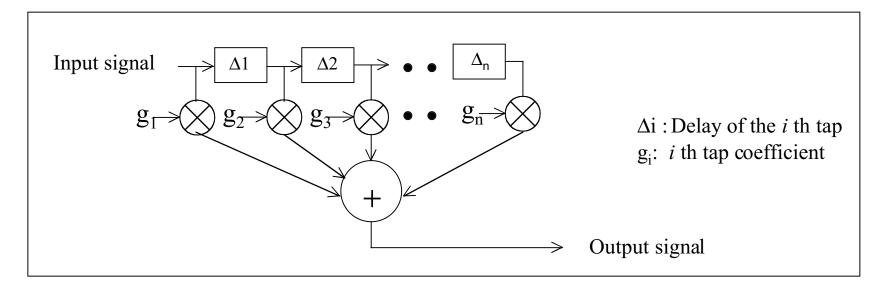
$$T_{rms} = \frac{\Delta \tau}{e^{\Delta \tau/2\tau_o} - e^{-\Delta \tau/2\tau_o}}$$

Good Model for directive antennas

7 Nov 00

## Multipath Fading Model

## Modeled by a Tapped Delay Line (TDL)

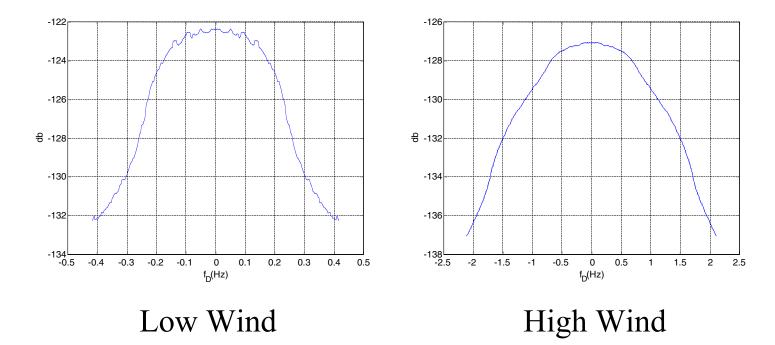


Output signal is a linear combination of time-shifted, attenuated copies of the input signal

# Multipath Fading Model (cont'd)

- Number of taps
- Delay values
- Tap Coefficients (random, uncorrelated)
  - K-Factor
  - Gain
  - Doppler spectrum

## **Doppler Power Spectrum**



Rounded Spectrum with  $f_D \sim 0.1$ Hz- 2Hz

## Antenna Correlation

For SIMO, MIMO channels, correlation between multiple channels depends on

- Spacing between antennas
- Height of the antennas
- Beamwidth
- Polarization

## Antenna Gain Reduction Factor

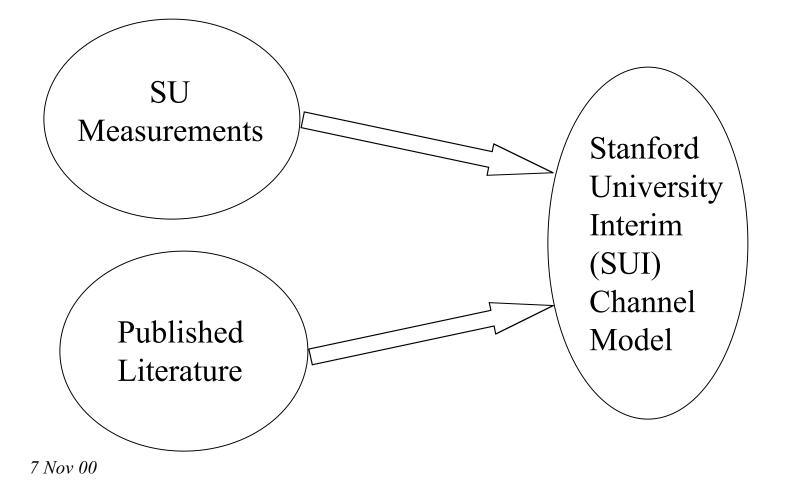
- The effective gain of a directional antenna is lower than the actual gain in a scattering environment
- The Gain reduction factor needs to be considered in the link budget

## Interference Model

- Co-channel Interference
  - Microscopic fading independent of the primary channel
  - C / I depends on
    - Reuse factor
    - Antenna beamwidth
- Adjacent channel Interference
  - Own system / ITFS channel

# Stanford University Interim (SUI) Generic Channel Models for G2 MMDS

## **Channel Modeling**



# Rationale for SUI Channel Models

- Many parameter combinations possible
- We picked 6 models related to deployment and Terrain scenarios typical of CONUS

Terrain	SUI Model
C Flat/Light Tree Density	SUI-1, SUI-2
B Flat/Moderate Tree Density	SUI-3, SUI-4
A Hilly/Moderate to Heavy Tree Density	SUI-5, SUI,6

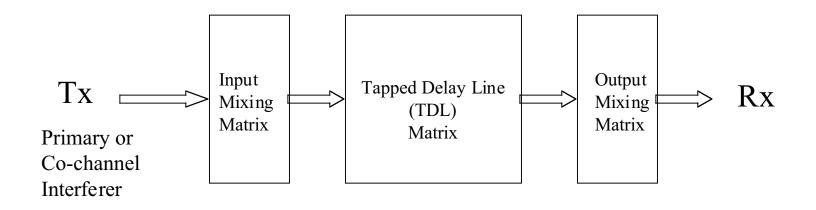
# Parametric View of SUI Channel models

	K: Low	Delay Spread		
		Low	Moderate	High
Doppler	Low	SUI-3		SUI-6
	High		SUI-4	SUI-5

K: Moderate/High	Delay Spread
------------------	--------------

		Low	Moderate	High
Doppler	Low	SUI-1,2		
	High			

## **SUI Channel Structure**



•A generic channel structure for a MIMO (2x3) channel (SISO, SIMO are subsets of this structure)

- •The mixing matrices correlate the signals at Tx and Rx
- TDL matrix introduces fading. Each TDL has 3 taps
- Antenna correlations are assumed to be the same everywhere

## Model Assumptions

- A cell size of 4 miles (6.4 km)
- BTS Antenna height: 50ft
- CPE antenna height: 10ft
- BTS Antenna beamwidth: 120 deg
- CPE Antenna Beamwidth: 50 deg
- Vertical Polarization only

## SUI-1 Channel

Terrain Category C

Ant Corr = 0.7				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	0.4	0.8	$\mu_{S}$
Power (P)	0	-15	-20	dB
K factor	18	0	0	
Doppler (D)	0.4	0.4	0.4	Hz

• RMS Delay Spread =  $0.1 \mu s$ , Overall K = 10

## SUI-2 Channel

Terrain Category C

Ant Corr = 0.5				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	0.5	1	μ <sub>S</sub>
Power (P)	0	-12	-15	dB
K factor	10	0	0	
Doppler (D)	0.4	0.4	0.4	Hz

• RMS Delay Spread =  $0.2 \mu s$ , Overall K = 5

## SUI-3 Channel

Terrain Category B

Ant Corr = 0.25				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	0.5	1	$\mu_{S}$
Power (P)	0	-5	-10	dB
K factor	0	0	0	
Doppler (D)	0.4	0.4	0.4	Hz

• RMS Delay Spread =  $0.3 \ \mu s$ 

## SUI-4 Channel

Terrain Category B

Ant Corr = 0.25				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	2	4	$\mu_{S}$
Power (P)	0	-4	-8	dB
K factor	0	0	0	
Doppler (D)	1	1	1	Hz

• RMS Delay Spread =  $1.3 \ \mu s$ 

## **SUI-5** Channel

Terrain Category A

Ant Corr = 0.25				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	5	10	$\mu_{S}$
Power (P)	0	-5	-10	dB
K factor	0	0	0	
Doppler (D)	2	2	2	Hz

• RMS Delay Spread =  $3 \mu s$ 

## **SUI-6** Channel

Terrain Category A

Ant Corr = 0.25				
	Tap 1	Tap2	Tap3	
Delay ( $\Delta$ )	0	14	20	$\mu_{S}$
Power (P)	0	-10	-14	dB
K factor	0	0	0	
Doppler (D)	0.4	0.4	0.4	Hz

• RMS Delay Spread =  $5.2 \ \mu s$ 

# Summary

- Interim Channel Models presented for G2 MMDS scenario
- SUI Channel Models developed based on measurements and published literature
- 6 typical channels presented

## Example: 3 Taps vs 6 Taps

Comparison of Frequency responses

3-taps

Dly =  $[0 \ 2 \ 5] \mu s$ Gain =  $[0 \ 4 \ -10] dB$ 

Frequency

6-taps

Dly =  $[0 \ 1 \ 2 \ 3 \ 4 \ 5] \ \mu s$ Gain =  $[0 \ -2 \ -4 \ -6 \ -8 \ -10] \ dB$ 

Frequency

Frequency Response (dB)

Frequency Response (dB)