OFDM-based PHY proposal for 802.16 TG3

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Purpose:

To present an OFDM based PHY proposal for 802.16.3 TG3

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OFDM-based PHY Proposal for 802.16.3 PHY

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Overview

- ¥ Very short OFDM intro
 - -GI, FFT size, constellations, coding
 - —See 802163p-00_30.ppt for more details
- ¥ Relation to the MAC philosophy
 - -Framing, Adaptive Rate, Concatenation, Polling, Random Access, Headers
- ¥ High Efficiency modes
 - —Improved ECC, longer FFT

Why OFDM?

- ¥ Multipath robustness
- ¥ Incorporated in data-oriented standards —802.11a, HIPERLAN/2: WLAN with QoS
- ¥ Incorporated in broadcast standards—DAB, DVB-T
- ¥ Facilitates smart antenna techniques in multipath environment
- ¥ Enables fast parallel polling
- ¥ Enables OFDMA



Solution 1 - Equalization

- ¥ Equalization is building an inverse filter
- ¥ If channel has nulls, you cannot inverse
- ¥ Decision Feedback Equalizer (DFE) can handle also channels with nulls
 - -In coded systems past decisions may be unreliable (not solved by Freq. Domain Equ.)
- ¥ In long channels complexity prblem



Solution 2 — Parallel Channels

- ¥ Send several long symbols in parallel
- ¥ OFDM uses complex exponential waveforms



Guard Interval and FFT Interval



Multipath effect on subcarriers

¥ Each subcarrier is scaled according to the channel, but they still do not interfere with each other



Modulation Constellations

- ¥ Use square QAM constellations only
- ¥ BPSK+4/16/64 QAM on downlink
 - —256 QAM optional
- ¥ BPSK+4/16 QAM on uplink

-64,256 QAM optional



Rx - QAM to metrics conversion

 \mathbf{F} Bit metric is generated based on

—proximity to nearest 0 and 1

—Subcarrier strength

- \blacksquare Gray coding \rightarrow One bit unreliable at a time
- \blacksquare Square constellation \rightarrow separate I & Q processing

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Error Correction Coding

¥ Convolutional code shall be used as a baseline mandatory mode.

—K=7, R=1/2, 2/3, 3/4; terminated tail

- ¥ Interleaver is needed to avoid adjacent faded bits
- ¥ Turbo Codes shall be used as an option



Turbo Codes

- ¥ Turbo codes are like crosswords
 - Solve rows, then use reliable bits to solve columns, then use the new reliable bits for rows etc. etc. till convergence
- ¥ TPC are proposed for compatibility with 802.16.1 and implementation maturity

—Better performance at very low BER



Convolutional Turbo Codes



Pilot subcarriers



- ¥ Pilot subcarriers are needed for carrier tracking
- ¥ Several are needed for:
 - —Fading Diversity \rightarrow spread pilots in frequency

—SNR accumulation

- ¥ 802.11a uses 4 pilots per 48 data subcarriers
- ¥ For 256pt FFT eight pilots are sufficient

—Smaller overhead



Preamble Structures

¥ The preamble is used to estimate

- —Antenna diversity selection and AGC convergence
- —Coarse, then fine **frequency** offset
- —Coarse, then fine **timing** offset

-Channel response

- ¥ More prior knowledge allows shorter preambles
 - —Gain preadjusted by transmit power control
 - -Coarse frequency offset known from prior transmissions

—Timing preadjusted by ranging and timing advance

	Coarse training sequence	Fine training s	equence	DATA1	DATA2	DATA3
/	t1 t2 t3 t4 t5 t6 t7 t8 t9 t10	GI2 T1	T2	θI	ĢI .	GI L

Preambles for Initial and Re-Acquisition



Or: DATA is received at RATE set by the MAC

FFT size tradeoffs

- ¥ GI is dictated by multipath duration
- ¥ Short FFT advantages
 - —Shorter training sequences
 - -Lower payload size granularity
 - —Phase noise tolerance
- ¥ Long FFT advantages
 - -Lower GI overhead and pilot symbol overhead
 - -Steeper spectrum falloff
 - -Facilitates OFDMA

Throughput vs. FFT size

¥ 64 pt FFT mode

-48 data subcarriers, 4 pilot subcarriers

¥ 256 pt FFT mode (optional gear shift)

—208 data subcarriers, 8 pilot subcarriers

¥ Faster spectral falloff is utilized to increase the fraction used.

¥ The 256 subcarrier mode provides

—27% rate improvement with 16 pt GI, 18% with 8 pt GI

¥ Little is bought by longer FFTs

Data Rates and Sensitivities

3.5 MHz wide channels, 52 or 216 subcarriers, 8 pt guard interval

Modulation	ECC rate	Rate, 64pt	Rate, 256pt	Sensitivity
BPSK	R=1/2	1.33 Mbit/s	1.57 Mbit/s	-94 dBm
BPSK	R=3/4	2.00 Mbit/s	2.36 Mbit/s	-93 dBm
QPSK	R=1/2	2.66 Mbit/s	3.15 Mbit/s	-91 dBm
QPSK	R=3/4	4.00 Mbit/s	4.73 Mbit/s	-89 dBm
16QAM	R=1/2	5.33 Mbit/s	6.30 Mbit/s	-86 dBm
16QAM	R=3/4	8.00 Mbit/s	9.45 Mbit/s	-82 dBm
64QAM	R=2/3	10.67 Mbit/s	12.60 Mbit/s	-78 dBm
64QAM	R=3/4	12.00 Mbit/s	14.18 Mbit/s	-77 dBm
256QAM	R=2/3	14.22 Mbit/s	16.81 Mbit/s	-73 dBm
256QAM	R=3/4	16.00 Mbit/s	18.91 Mbit/s	-71 dBm

Sensitivity assumes NF=6 dB and 4 dB implementation loss

Mixing 64 and 256 FFT modes

- ¥ Reuse 64 pt mode preambles and signaling
 - —Compatibility
 - —Smaller overhead
- ¥ Payloads may be sent at 256pt mode
 - -Channel estimation can be interpolated from the 64pt mode estimate
 - —The carrier tracking loop achieves better frequency accuracy during the 64pt mode, then gears into the 256pt mode
- ¥ Concatenated payloads may use either 64 point or 256 point mode, depending on CPE s capabilities

Time-frequency view

Training

Data



Mixing FFT and modulation modes



Subcarrier based parallel polling

- ¥ Fourier Transform allows simultaneous detection of multiple subcarriers sent by multiple users
 - -Extreme case of OFDMA combined with On-Off Keying with 1 subcarrier per user.
 - —Demands preranging
- ¥ CDMA-like, but preserves orthogonality
- ¥ Concentrates power, allows higher SNR
- ¥ Permute frequencies in each superframe to avoid prolonged fades

Subcarrier based polling











Several subcarriers can be sent for higher detection reliability in each attempt or convey a few-bit message, at expense of polling time

Optional Advanced Techniques

¥ OFDMA

- -The OFDM preserves orthogonality between transmissions of different users
- —Allows survival at higher path loss
- ¥ Space-Time coding
 - —The decoupling between equalization and coding plays important role in making those techniques practical
 - —New preambles need to be designed for training of response from multiple antennas

BRZE s OFDM proposal Summary

- ¥ Parameters draw on 802.11a+HIPERLAN/2 standards
 - -Available technology
- ¥ Improved performance modes
 - -Longer FFTs, improved ECCs
- ¥ Fast Parallel Polling for fast demand discovery
- ¥ Ready for advanced antenna and multiaccess techniques

Peak2Avg Problem- How bad?

- \mathbf{Y} Worst case peaks are *kN* times the average
 - -N is the number of subcarriers
 - -k is constellation dependent, about 3 dB for QAM
 - —20 dB for *N*=52, 26 dB for *N*=216
- ¥ Central Limit Theorem (sum of many small contributions) → amplitude is Rayleigh
- ¥ Worst peak in a typical packet is +10 dB
- ¥ Some clipping can be tolerated!!
 - -OFDM spreads clips over subcarriers
 - -Error Correction Coding improves robustness
- ¥ Typical PA backoff 7-9 dB

—Depends on constellation and on regulatory masks



PAPR — back of an envelope



- ¥ Central Limit Theorem (sum of many small contributions) → amplitude is Rayleigh
- $P(P>P_{th})=exp(-P/P_{avg})$
- ¥ Typical packet is about 20,000 samples
 - —Look for threshold with crossing probability 5*10⁻⁵
- $P_{th}/P_{avg} = -\ln(5*\ 10^{-5}) = 9.9 = 10 \text{ dB}$
- ¥ Remember that some clipping can still be tolerated

Freq. Domain.Equ — Uplink Option

- ¥ The CPEs have the most to gain at PA
- \mathbf{F} The processing burden is at BST
- ¥ Frees CPE chip manufacturers to work

