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Re:	DL MIMO Rapporteur Group Discussions			
Abstract	To clarify the concerns on non-unitary precoding.			
Purpose	For review and discussion in the Project 802.16m DL MIMO Rapporteur Group			
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On PAPR, CQI mismatch and inter-cell interference of non-unitary precoding

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5 Abstract: To clarify some concerns on non-unitary precoding regarding PAPR, CQI mismatch and variance of inter-cell interference, corresponding simulations are included. Firstly, it is shown that the non-unitary 6 7 precoding and amplitude uniformity of the precoding vector (matrix) has no impact on PAPR. In other words, 8 PAPR should not be an issue to differentiate between unitary and non-unitary precoding. Secondly, it is shown 9 that the CQI mismatch of non-unitary precoding is rather trivial. With a proper codebook design, existing CQI 10

estimation of non-unitary precoding is sound. Thirdly, the variances of inter-cell interference (ICI) associated with unitary and non-unitary precoding are investigated. Simulations show that the variances of ICI

corresponding to unitary and non-unitary precoding are almost identical.

1. PAPR

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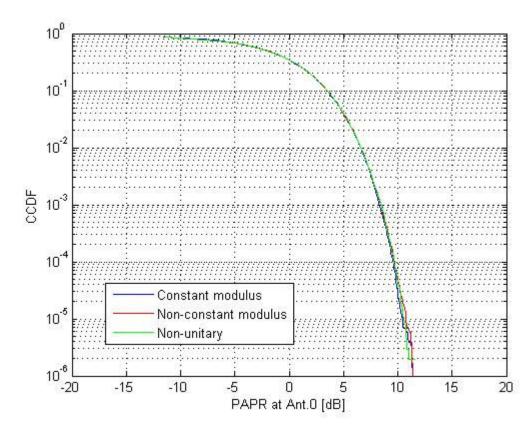
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The peak value s_m is defined such that the probability of the signal power greater than s_m equals to p_0 .

$$\Pr\{S_n \ge S_m\} = p_0$$

As a result, PAPR is defined by $PAPR = 10 \log_{10} \left(\frac{S_m}{E(S_n)} \right)$.



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Simulation Parameters			
Number of transmit antenna	4		
FFT size	1024		
RB size	1 precoding vector per 72 subcarriers		
Modulation	QPSK		
Codebook	Uniform amplitude: 4 bits LTE		
	Unbalanced amplitude: V(4,1,6) in Wimax		

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It is clearly shown in Fig. 1 that the amplitude uniformity of precoding vector (matrix) has very little impact of PAPR. As a result, PAPR should not be a factor to differentiate between unitary and non-unitary precoding.

2. CQI mismatch (without inter-cell interference)

- 8 The real and instantaneous SINR of MS j is denoted by $SNR_{j,real}^{BS}$, given as
- 9 $SINR_{j,real}^{BS} = \frac{\left|hw_{j}^{*}\right|^{2}}{\frac{\sigma^{2}K}{P} + \sum_{\substack{i=1\\i\neq j}}^{K} \left|hw_{i}^{*}\right|}$
- 10 , in which $h = U^*(:,1)H$, where $H = U\Sigma V^*$ is channel matrix of MS j, K is the number streams and
- 11 $w_i, i \in [1,..,K]$, is the precoding vector associated with the i-th stream, P is total transmit power and σ^2 is the
- 12 power spectrum density of additive noise.
- 13 The estimated SINR denoted by $SINR_j^{MS}$ is obtained by the MS and fed back to the BS. $SINR_j^{MS}$ is given by [1]
- $SINR_{j}^{MS} = \frac{\cos^{2} \theta}{\frac{\sigma^{2} N_{t}}{P |h|^{2}} + \sin^{2} \theta}$
- Where $\cos \theta = \frac{\left| h \hat{h}^* \right|}{|h|}$ denotes the quantization error between h and \hat{h} , where \hat{h} is the quantized h based on a
- predefined codebook, N_{\star} represents the number of transmit antenna at BS.
- An alternative way is that MS feeds back SNR denoted by SNR_j^{MS} , given by
- $SINR_{j}^{MS} = \frac{\cos^{2} \theta}{\frac{\sigma^{2} N_{t}}{P|h|^{2}}}$
- 19 At the BS, based on user grouping information, $SINR_{j}^{MS}$ or SNR_{j}^{MS} , is refined to $SINR_{j,refined}^{BS}$. As a result,
- $SINR_{j,refined}^{BS}$ is obtained as

$$1 \qquad SINR_{j,refined}^{BS} = \frac{N_t}{K} \left| \hat{h} w_j^* \right|^2 SINR_j^{MS} = \frac{\cos^2 \theta \left| \hat{h} w_j^* \right|^2}{\frac{\sigma^2 K}{P \left| h \right|^2} + \frac{K}{N_t} \sin^2 \theta}$$

2 The CQI mismatch is denoted by $10 \log_{10} \left(\frac{SINR_{j,refined}^{BS}}{SINR_{j,real}^{BS}} \right)$

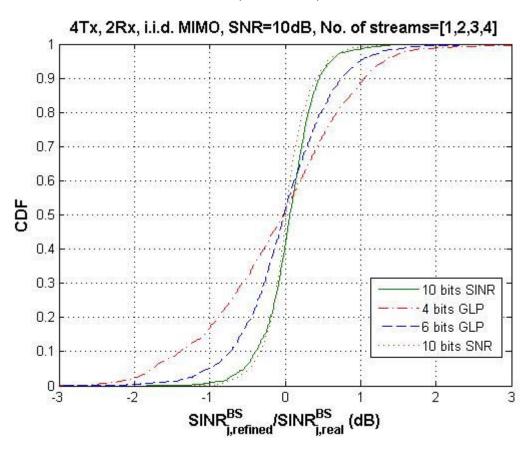


Figure 2: CQI mismatch of ZFBF with various size of codebook

It is shown in Fig. 2 that when the codebook is properly selected (for example the selected 10 bits codebook), the CQI mismatch of non-unitary precoding is very limited. This can be illustrated by the near vertical CDF in Fig. 2. Moreover, if the channel is correlated, a smaller space needs to be quantized. As a result, an even smaller codebook can also achieve very sound CQI estimation.

3. Inter-cell interference

The inter-cell interference is defined as

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$$P_{ICI} = \sum_{i=1}^{N} u \left(\sum_{j=1}^{M} H_i v_{i,j} \right)$$

- where N is the number of dominant sources, u is a combining vector at MS, H_i represents the channel matrix
- between MS and the i-th interference BS and $v_{i,j}$ denotes the precoding vector of the j-th layer at the i-th
- interference BS.

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Simulation Parameters			
Number of transmit antenna	2, 4		
Number of dominant interference	2		
Resource allocation	Same dedicated MU-MIMO zone for all BSs.		
Scheduler	Independent scheduling of each BS		
Precoding Scheme	Same precoding scheme for all BSs		
Codebook	Unitary: 4 bits randomized unitary codebook		
	Unbalanced amplitude: 3bit GLP(2Tx) 6 bit		
	GLP(4Tx)		



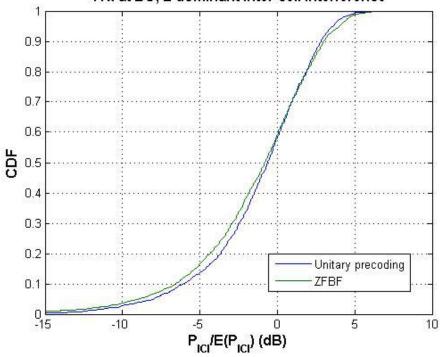


Figure 3 Inter-cell interference comparison (4Tx) based on various precoding schemes.

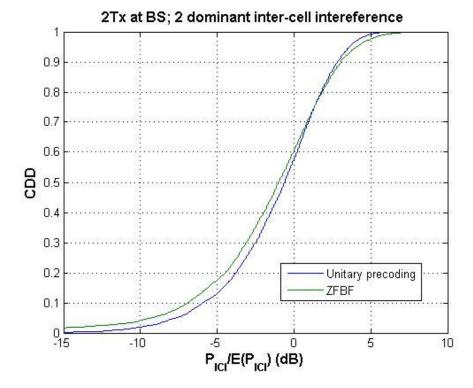


Figure 4 Inter-cell interference comparison (2Tx) based on various precoding schemes.

It has been shown in Figure 3 and 4 that the variance of the inter-cell interference is almost identical for both unitary precoding and ZFBF.

Conclusions:

- (1) The impact of amplitude uniformity of precoding vector (matrix) is rather trivial. As a result, PAPR should not be a factor to differentiate between unitary and non-unitary precoding.
- (2) The performance of current existing CQI estimation scheme for non-unitary precoding is sound. With proper codebook selection, CQI mismatch is not an issue for non-unitary precoding.
- (3) Compared to the unitary precoding, non-unitary precoding (e.g. ZFBF) does not result in a bigger variance of inter-cell interference.

References:

[1] Philip R1-062483 "Comparison between MU-MIMO codebook-based channel reporting techniques for LTE downlink" 3GPP TSG RAN WG1 Seoul, South Korea, 9th October – 13th October 2006