
802.3ae Transmitter & Dispersion Penalty (TDP) and Stressed Eye sensitivity tests

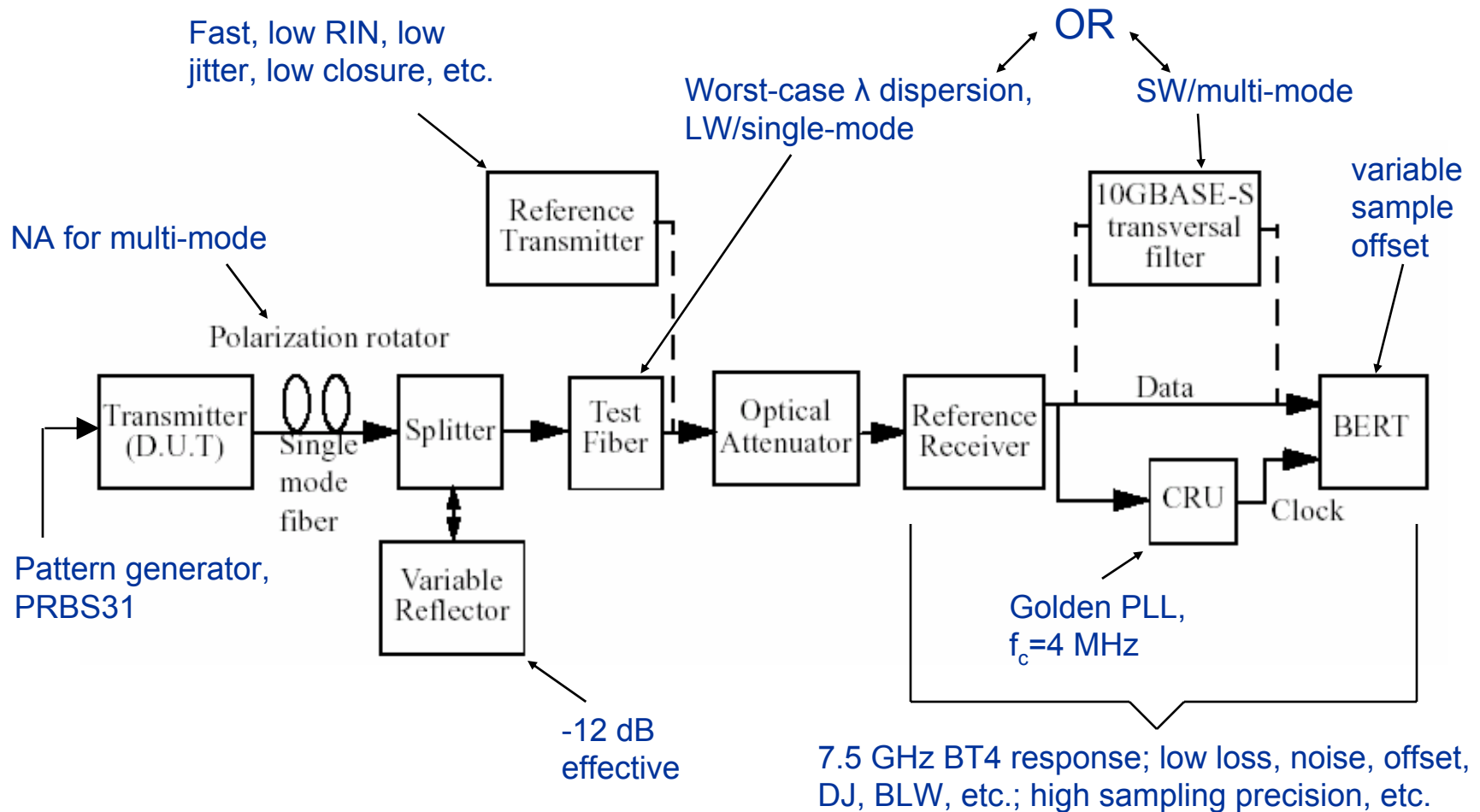
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Contents of these slides courtesy
of SyntheSys Research, Inc.

TDP - description

- Primary measure of Tx quality
 - TDP measures to 1E-12 probabilities
 - Also eye mask at TP2
 - Eye mask is high probability, qualitative test
- At virtual TP3: measured at output of test setup that includes worst case dispersion
- Take worse case of sampling ± 0.05 UI from eye center
 - Eye center defined by mid of 1E-3 points
 - Forces at least that much horizontal eye opening at TP3
- Bottom line spec, inherently includes parametric tradeoffs
 - Includes amplitude penalties

TDP – setup, Figure 52-12



TDP - procedure

1. Ref Tx, short fiber; adjust attenuator to 1E-12 BER, measure OMA*
 - BERT threshold at average power, sample time at eye center
2. DUT, dispersion fiber (or transversal filter); adjust attenuator to 1E-12 BER, measure OMA*
 - BERT threshold at average power, sample time +/-0.05 UI from eye center, take result with less sensitivity
3. TDP = difference in OMAs

*OMA defined as averaged and settled pk-pk values with low frequency square

TDP - history

- 802.3z spec'd eye mask and jitter at $1E-12$ BER, both at TP2
- 802.3ae was with eye mask at TP2 and bathtub jitter at TP3
- TDP proposed instead of jitter bathtub, Jan/02
 - At the time, BERT Rx unable to accurately measure jitter at 10G
 - Insufficient sensitivity, too much distortion, setup/hold, scanning inaccuracies, etc.
 - TDP is a (relative) power penalty test
 - Stronger relationship to link model
 - Weak relationship of jitter to vertical amplitude penalties
 - Goal: provide some compensation for shortcomings of Rx tester via difference measurement
- Aggressive 802.3ae schedule, TDP was not thoroughly tested or understood

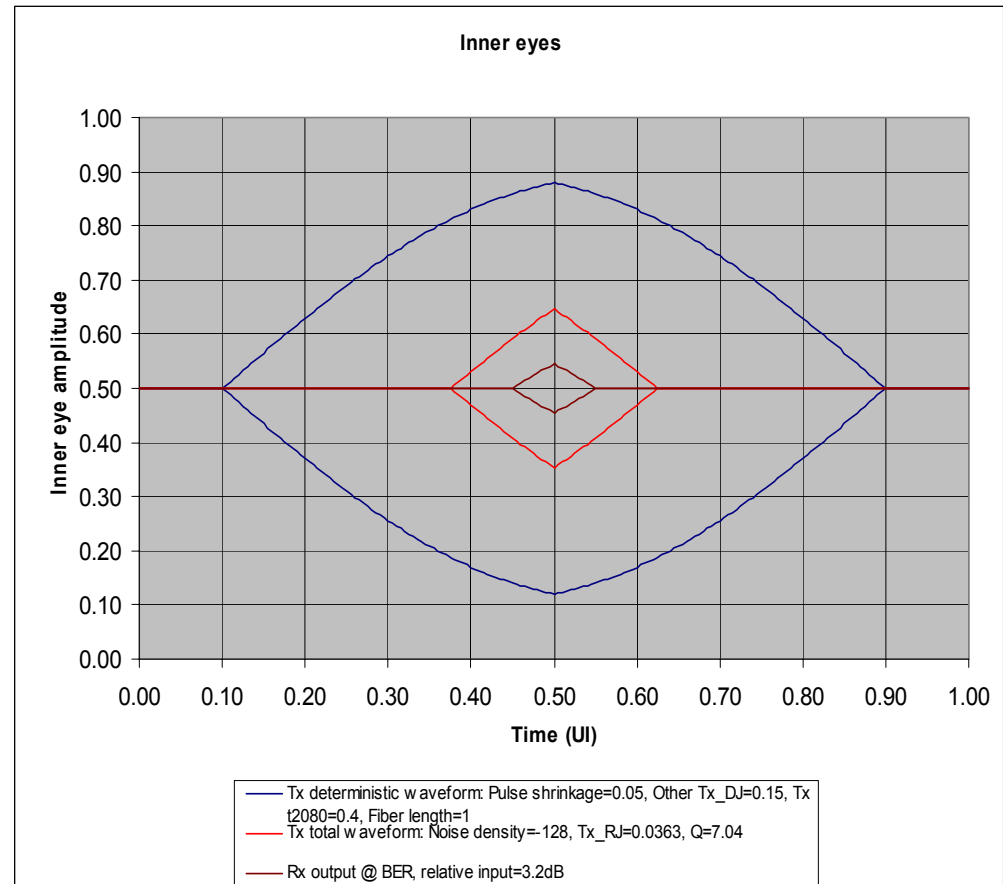
TDP – limitations (see 03-387v0)

- Still suffers from equipment “challenges”
- Still uses BER Rx, *but they are getting much better...*
 - Poor Ref Rx equipment assumed caused by DUT, stresses yields
- Yet spec allows too much jitter from DUT
 - If Ref Rx exceeds requirements, inadequate jitter control
 - Before, jitter was insufficiently related to penalties, but now TDP is insufficiently related to jitter
 - TDP allows ~0.8 UI pk-pk jitter at TP3 output
 - Exceeds jitter of Stressed eye for Rx test
 - Exceeds jitter tolerance of most CDRs!
 - Poor Ref Tx reduces TDP reference, allows lower DUT quality

TDP – example of excessive jitter

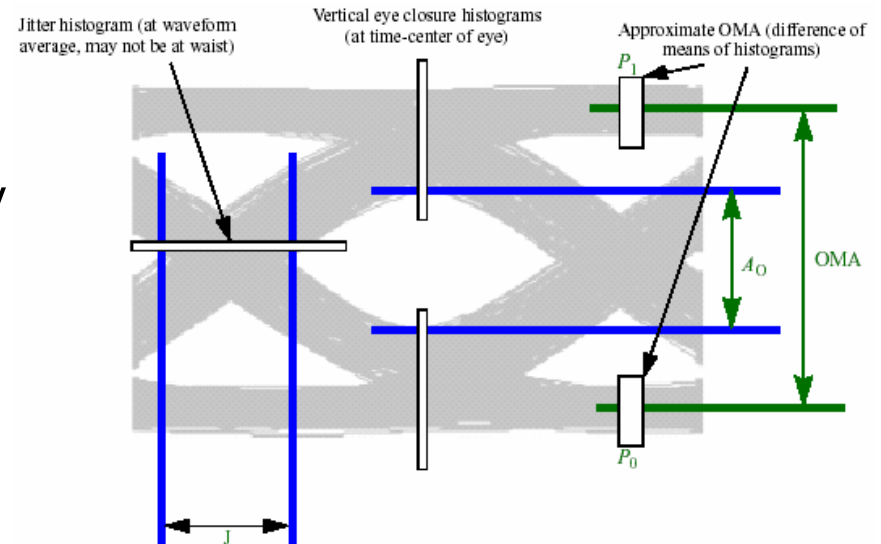
- 3.2 dB TDP
 - per 10GBASE-L
- 0.76 UI at TP3 (red)
 - Not worst case...
- Perfect test equipment
 - See previous slide

Navy is deterministic eye at TP3
Red is random/total eye at TP3
Brown is random/total eye at TP4



Stressed eye - description

- Primary measure of receiver quality
- TP3 input
 - Same interface as TDP (output)
- Rx must receive with BER < 1E-12
- Properties per link model
 - OMA
 - High quality pattern generator and E/O
 - Minimum Tx power - fiber loss - pk-pk random penalties ($2 \cdot Q = 14$)
 - VECP (dB optical power opening between 0.1% cumulative hit regions)
 - Deterministic closure calculated from dispersion and slow Tx edge rates
 - Accomplished with 3-4 GHz BT4 filter & sine interference
 - J = deterministic jitter (time opening between 1% cumulative hit regions)
 - Accomplished with filter (minimal), sine interference and sine jitter



Stressed eye – stress mechanisms

- Filter - clean, vertical closure with minimal phase distortion & jitter
- Sine interference (SI)
 - Amplitude closure, also translates to jitter via waveshape
 - Jitter is modulated pulse-width (bit shrinkage)
 - Tough on Rx BW and sampling blocks
 - No phase shift, so cannot be tracked with PLL
 - SI frequency range (sweeping is optional, but *customer* is allowed to...)
 - 0.1 – 2 GHz
 - Above AC coupling, so must be dissipated
 - Deterministic, so can be measured on scope
- Sine jitter (SJ)
 - 1/f roll-up from 4 MHz down to 40 kHz
 - Verifies CDR tracking
 - Flat from 4 MHz to 10x DUT_loop_bandwidth, 0.05 - 0.15 UI
 - Also translates to amplitude closure via waveshape
 - Deterministic, so can be measured on scope
- High freq SJ amplitude, filter and SI co-adjusted to meet $J = 0.3 \text{ UI}$
- Attenuator – reduce OMA to spec

Stressed eye - procedure

1. Interactively calibrate VECF, J, minimum SJ, pulse shrinkage (5-15 psec)
 - Short patterns may be used if frequency loss has been minimized
2. Adjust optical power (attenuator)
3. Set test pattern
4. Sweep SJ frequency range
 - Sweep SI frequency range (optional)
5. Confirm BER <1E-12

Stressed eye - history

- First used in 1G Ethernet (~1998)
 - Purpose – to be sure receivers have sufficient bandwidth to meet worst case waveform
 - Max BW already understood and controlled
- Adopted by FC-P1, FC-P1-2 (1,2,4G Fibre Channel, 1999-2003)
- Adopted and modified by 10GE (2002)
 - Clearer relationship of stressed eye to link model
 - Built from deterministic elements for better visibility and calibration
- Spec was rushed to meet schedule, not thoroughly understood

Stressed eye – limitations

- Stressed eye may not satisfy interoperability needs
 - Stressed eye minimizes random noises and jitter in test
 - Done so random signals can be observed reliably on scope; no low probability 10G instruments were sufficiently accurate at the time
 - However, random jitter can exist in application (clock jitter, RIN, etc.)
 - $J = 0.3$ UI pk-pk jitter calibrated at TP3 input
 - Some RJ will exist, but TJ likely < 0.8 UI (from TDP spec)
- Limit on SI amplitude *may* prevent achieving $J = 0.3$
 - Spec requires $> 2/3$ VECP from filter, yet ideal filter produces negligible jitter
 - Remaining $1/3$ from SI and SJ (0.15 max) produces ~ 0.26 max
- Expect variation across industry
 - in settings (too many degrees of freedom)
 - in accuracy of calibration equipment