# 400GBASE-SR4.2 optical penalties 

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## Introduction

- 400GBASE-SR4.2 is proposed to go $50 \%$ further on OM5 than 50GBASE-SR / 100GBASE-SR2 / 200GBASE-SR4 / 400GBASESR8
-150 m vs. 100 m on $\mathrm{OM} 5,100 \mathrm{~m}$ vs. 100 m on OM4
- With more chromatic dispersion caused by the extra distance, we need to revisit the mode partition noise penalty
- When the combination of all the other impairments is too high, modal noise becomes significant too


## Method of estimating penalties

- The next slide starts with the well-researched 10GBASE-SR specification and model
- Scales for spectral width, frequency, FEC, PAM4 and reach
- Recognises recent investigations into mode partition noise k factor
- Shows example ways of bringing the total penalties back to a very high but plausible level
- Unlike dawe 3cd 01b 0918 this calculation fully includes the Pcross effect
- Unlike dawe 3 cm adhoc 01092718 this calculation follows the OgawaAgrawal equations as in the 10 Gigabit Ethernet link model spreadsheet for scaling modal noise, and it includes recent small improvements in the fibre's specified chromatic dispersion
- ( $\sim-108 \mathrm{ps} / \mathrm{nm} / \mathrm{km}$ instead of ${ }^{\sim}-118 \mathrm{ps} / \mathrm{nm} / \mathrm{km}$ )
- Like the draft, it assumes faster lasers for $>100 \mathrm{~m}$


# Estimates of budget with minor noise penalties 

| 10GBASE-SR |  |  |  | 100GBASE-SR4 |  | 400GBASE-SR8 |  |  | 400G-4.2 D1.0 |  | $\begin{aligned} & \text { 400G-4.2 } \\ & \text { better Tx } \end{aligned}$ |  | $\begin{gathered} 400 \mathrm{G}-4.2 \\ 125 \mathrm{~m} \end{gathered}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | eadsh | example | Estima two k | es for values | $\begin{gathered} \text { As in } \\ \text { P802.3cd } \end{gathered}$ | Pessimistic | Optimistic | Pessimistic | Optimistic | Pessimistic | Optimistic | Pessimistic | Optimistic |
| PAM- (no. levels) |  |  | 2 |  |  | 4 |  |  |  |  |  |  |  |  |
| No. eyes |  |  | 1 |  |  | 3 |  |  |  |  |  |  |  |  |
| Qmin |  |  | 7.0345 | 3.8906 |  | 3.414 |  |  |  |  |  |  |  |  |
| TDP, TDEC or TDECQ |  | dBo | 3.9 | 4.3 |  | 4.5 |  |  |  | . 5 | 3.84 | 4.06 | 4.00 | 4.16 |
| Total penalty |  | dBo | 4.2 | 4.3 | 4.11 | 4.60 | 4.95 | 4.80 | 5.41 | 5.05 | 4.50 | 4.50 | 4.50 | 4.50 |
| Signalling rate |  | GBd | 10.3125 | 25.78125 |  | 26.5625 |  |  |  |  |  |  |  |  |
| Rea |  | m | 300 | 100 |  | 100 |  |  | 150 |  | 150 |  | 125 |  |
| Spectral width |  | nm | 0.29 | 0.6 |  | 0.6 | 0.6 |  | 0.6 |  | 0.6 |  | 0.6 |  |
| MPN penalty |  | dBo | 0.1 | 0.14 | 0.02 | 0.02 | 0.15 | 0.09 | 0.55 | 0.31 | 0.41 | 0.25 | 0.25 | 0.15 |
| MN penalty |  | dBo | 0.3 | 0.11 | 0.03 | 0.08 | 0.30 | 0.22 | 0.36 | 0.24 | 0.25 | 0.19 | 0.25 | 0.19 |
| Combined |  | dBo | 0.4 | 0.24 | 0.05 | 0.10 | 0.45 | 0.30 | 0.91 | 0.55 | 0.66 | 0.44 | 0.50 | 0.35 |
| MPN $k$, also used for MN |  |  | 0.3 | 0.3 | 0.1 | 0.0296 | 0.1 | 0.075 | 0.1 | 0.075 | 0.1 | 0.075 | 0.1 | 0.075 |
| TDP, TDEC | ECQ | Pmpn | 3.8 | 3.92 | 4.04 | 4.5 | 4.5 | 4.5 | 4.5 | 4.5 | 3.84 | 4.06 | 4.00 | 4.16 |
| Rate*reach*spectral width |  |  | 897 | 1547 |  | 1594 |  |  | 2391 | 2391 | 2391 | 2391 | 1992 | 1992 |
| MPN noise | rel. O | uter | 0.01247 | 0.0257 | 0.0086 | 0.0030 | 0.0090 | 0.0068 | 0.0171 | 0.0128 | 0.0171 | 0.0128 | 0.0130 | 0.0098 |

Start with the penalties and $k$ factor in 10GBASE-SR. Compare 100GBASE-SR4, where estimated Pmpn and Pmn are low because PAM2 and FEC
In right hand columns, assume $k$ is 0.1 or 0.075 . Scale the $10 \mathrm{G} / 25 \mathrm{G}$ noises and predict the penalties for 802.3 cm MMF: around 0.4 to 0.9 dB , bringing the total penalty to around 4.9 dB to 5.4 dB , which is too high. There is only 0.1 dB in the budget for these penalties
The modal noise penalty could be higher or lower for all columns together - need new information

## Pcross effect



MPN and MN penalties become bad only when TDECQ (without them) is bad

## Discussion

- These links are dispersion-limited not power-limited
- It's about the penalties, not so much about the budget
- Mode partition noise is a concern and modal noise is a contributor
- The classic theory of mode partition noise may not be accurate enough
- Equations used here may under-estimate MPN for an equalised link and high total penalties
- Up-to-date information on modal noise is needed, e.g. from experiments
- See also pepeljugoski 11104 pepeljugoski 011212 mmf and dawe 040114 optx


## Way forward for the spec

- We know how to account for mode partition noise and modal noise because we did it in 802.3bm (100GBASE-SR4)
- In 138.8.5, insert:
- Equation (138-1) is used in place of Equation (12111).
- $R=\sqrt{ }\left(\sigma_{G}{ }^{2}+\sigma_{S}{ }^{2}-M^{2}\right)$
- where $M=0.0065 P_{\text {ave }}$
- [Need to agree and/or refine the number 0.0065. $P_{\text {ave }}$ is already defined in 121.8.5.3]
- In 138.8.10 Stressed receiver sensitivity, insert:
- the value of $M$ in Equation (138-1) is set to zero, and


## Discussion 1

- This addresses modal noise
- This is simpler than 95.8.5.2 which uses two terms
- Also more optimistic than 95.8.5.2 which uses a much higher value
- Mode partition noise could be handled as in 95.8.5.2 or with a fixed allocation in the budget


## Discussion 2

- This would be appropriate if modal noise is not significantly affected by the equalizer
- The mode partition noise theory already assumes an equalized signal
- However, it seems probable that modal noise can have a similar or wider spectrum as RIN, so undergoes noise enhancement like receiver noise or RIN
- The next slide shows a simple alternative fix to take noise enhancement into account


## Alternative way forward

- In 138.8.5, insert:
- Equation (138-1) is used in place of Equation (12111).
- $R=\sqrt{ }\left(\sigma_{G}^{2}+\sigma_{S}^{2}-\left(M\left(C_{\mathrm{eq}}\right)^{2}\right)\right.$
where $M=0.0065 P_{\text {ave }}$
- [Need to agree and/or refine the number 0.0065. $P_{\text {ave }}$ is already defined in 121.8.5.3]
- In 138.8.10 Stressed receiver sensitivity, insert:
- the value of $M$ in Equation (138-1) is set to zero, and


## Alternative in context

### 138.8.5 Transmitter and dispersion eye closure for PAM4 (TDECQ)

... Compensation may be made for any deviation from an ideal fourth-order Bessel-Thomson response.

- Equation (138-1) is used in place of Equation (121-11).

$$
\begin{aligned}
& \underline{R=\sqrt{ }\left(\sigma_{G}{ }^{2}+\sigma_{\underline{s}}{ }^{2}-\left(M / C_{\text {eq }}\right)^{2}\right)} \\
& \underline{\text { where } M=0.0065 P_{\text {ave }}}
\end{aligned}
$$

- The reference equalizer to be used for TDECQ for 50GBASE-SR, 100GBASE-SR2, and 200GBASE-SR4 is specified in 138.8.5.1.


### 138.8.10 Stressed receiver sensitivity

- The SECQ of the stressed receiver conformance test signal is measured according to 138.8 .5 , except that the value of $M$ in Equation (138-1) is set to zero, and the combination of the $\mathrm{O} / \mathrm{E}$ converter and the oscilloscope has
[In the first way forward, " $/ C_{\text {eq }}$ " would be omitted]


## Effect of under-estimating MN and MPN in D3.5 - without their noise enhancement

What we really allow vs. D3.5 TDECQ


D3.5 measures the blue lines
In service receiver may experience the orange lines
Significantly worse penalty when TDECQ is bad
This slide assumes these noises do not undergo noise enhancement

# Effect of under-estimating MN and MPN in D3.5 - with full noise enhancement 

What we really allow vs. D3.5 TDECQ


D3.5 measures the blue lines
In service receiver may experience the orange lines
Significantly worse penalty when TDECQ is bad
This slide assumes these noises undergo full noise enhancement

