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Title	LDPC coding for OFDMA PHY	
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Source(s)	Brian Classon Yufei Blankenship Motorola	brian.classon@motorola.com yufei.blankenship@motorola.com
	Jerry Kim Panyuh Joo Gyubum Kyung Hongsil Jeong Sang-Hyo Kim Hanju Kim DS Park Jaeho Jeon Samsung	kimjy@samsung.com panyuh@samsung.com gyubum.kyung@samsung.com hongsil.jeong@samsung.com sanghyo7.kim@samsung.com hanju.kim@samsung.com dspark@samsung.com jhjeon@samsung.com
	Eric Jacobsen Bo Xia Intel	eric.a.jacobsen@intel.com bo.xia@intel.com
	Kyuhyuk Chung Min-seok Oh LG Electronics	kyuhyuk@lge.com minoh@lge.com
	Brian Johnson Alek Purkovic Nina Burns Sergey Sukobok Nortel Networks	brjohnso@nortelnetworks.com apurkovi@nortelnetworks.com nburns@nortelnetworks.com ssukobok@nortelnetworks.com
	Eli Sasha Zion Hadad Yigal Leiba Itzik Kitroser Yossi Segal Oren Elkayam Runcom Technologies	shasha@runcom.co.il zionh@runcom.co.il yigall@runcom.co.il itzikk@runcom.co.il yossis@runcom.co.il Elkayam@runcom.co.il

	Simon Litsyn Eran Sharon Tel-Aviv University.	litsyn@eng.tau.ac.il eransh@eng.tau.ac.il	
	Dale Hocevar Anuj Batra Texas Instruments	hocevar@ti.com batra@ti.com	
	Victor Stolpman Jianzhong (Charlie) Zhang Nico Van Waes Nokia	Victor.Stolpman@nokia.com Charlie.Zhang@nokia.com	
	Masoud Olfat Nextel Communications	masoud.olfat@nextel.com	
Re:	IEEE P802.16-REVe/D5-2004, sponsor ballot		
Abstract	This contribution contains additional text output from an informal LDPC group.		
Purpose	Provide additional LDPC specification text.		
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standard being developed within the IEEE 802.16 Working Group. The Chair will disclose this notification via the IEEE 802.16 web site http://ieee802.org/16/ipr/patents/notices>.

Overview

An informal LDPC group has been working on the goal of achieving consensus on a proposed LDPC code design as an optional advanced code for the OFDMA PHY. Many excellent code designs have been submitted. The codes have been qualitatively and quantitatively characterized, and it is clear that a LDPC code with excellent flexibility and performance, as well as low encoding and decoding complexity, can be defined for 802.16e. This contribution provides additional LDPC specification text.

Recommended Text Changes:

Add the following text to 802.16e_D5, adjusting the numbering as required:

<Add a new paragraph to section 8.4.9.2.5.1 Code Description, the new paragraph to precede the last paragraph of the section " \mathbf{H}_{b} is partitioned into two sections...">

The permutations used are circular right shifts, and the set of permutation matrices contains the $z \times z$ identity matrix and circular right shifted versions of the identity matrix. Because each permutation matrix is specified by a single circular right shift, the binary base matrix information and permutation replacement information can be combined into a single compact model matrix \mathbf{H}_{bm} . The model matrix \mathbf{H}_{bm} is the same size as the binary base matrix \mathbf{H}_{b} , with each binary entry (i,j) of the base matrix \mathbf{H}_{b} replaced to create the model matrix \mathbf{H}_{bm} . Each 0 in \mathbf{H}_{b} is replaced by a blank or negative value (e.g., by -1) to denote a $z \times z$ all-zero matrix, and each 1 in \mathbf{H}_{b} is replaced by a circular shift size $p(i,j) \ge 0$. The model matrix \mathbf{H}_{bm} can then be directly expanded to \mathbf{H} .

<Add the material below to the end of section 8.4.9.2.5.2 LDPC encoding, directly after the sentence "The LDPC codes are defined such that very low complexity encoding directly from **H** is possible." > The following informative subsection shows two such methods.

Direct Encoding (Informative)

For the two methods, described below, section \mathbf{H}_{b2} is further partitioned into two sections, where vector \mathbf{h}_{b} has odd weight, and $\mathbf{H}\boldsymbol{\xi}_{2}$ has a dual-diagonal structure with matrix elements at row *i*, column *j* equal to 1 for *i*=*j*, 1 for *i*=*j*+1, and 0 elsewhere:

$$\mathbf{H}_{b2} = \begin{bmatrix} \mathbf{h}_{b} \mid \mathbf{H}_{b2}' \end{bmatrix}$$
$$= \begin{bmatrix} h_{b}(0) \mid 1 & & \\ h_{b}(1) \mid 1 & 1 & \mathbf{0} \\ \vdots & & 1 & \ddots \\ \vdots & & \ddots & 1 \\ \vdots & & \mathbf{0} & 1 & 1 \\ h_{b}(m_{b}-1) \mid & & & 1 \end{bmatrix}$$

The base matrix has $h_b(0)=1$, $h_b(m_{\underline{b}}-1)=1$, and a third value $h_b(j)$, $0 < j < (m_b-1)$ equal to 1. The base matrix structure avoids having multiple weight-1 columns in the expanded matrix.

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In particular, the non-zero submatrices are circularly right shifted by a particular circular shift value. Each 1 in $\mathbf{H}\boldsymbol{\xi}_2$ is assigned a shift size of 0, and is replaced by a $z \times z$ identity matrix when expanding to **H**. The two 1s located at the top and the bottom of \mathbf{h}_b are assigned equal shift sizes, and the third 1 in the middle of \mathbf{h}_b is given an unpaired shift size. The unpaired shift size is 0.

Method 1

Encoding is the process of determining the parity sequence **p** given an information sequence **s**. To encode, the information block **s** is divided into $k_b = n_b - m_b$ groups of *z* bits. Let this grouped **s** be denoted **u**,

$$\mathbf{u} = \begin{bmatrix} \mathbf{u}(0) & \mathbf{u}(1) & \cdots & \mathbf{u}(k_b - 1) \end{bmatrix},$$

where each element of \mathbf{u} is a column vector as follows

$$\mathbf{u}(i) = \begin{bmatrix} s_{iz} & s_{iz+1} & \cdots & s_{(i+1)z-1} \end{bmatrix}^T$$

Using the model matrix \mathbf{H}_{bm} , the parity sequence \mathbf{p} is determined in groups of *z*. Let the grouped parity sequence \mathbf{p} by denoted \mathbf{v} ,

$$\mathbf{v} = \begin{bmatrix} \mathbf{v}(0) & \mathbf{v}(1) & \cdots & \mathbf{v}(m_b - 1) \end{bmatrix}$$

where each element of \mathbf{v} is a column vector as follows

$$\mathbf{v}(i) = \begin{bmatrix} p_{iz} & p_{iz+1} & \cdots & p_{(i+1)z-1} \end{bmatrix}^T$$

Encoding proceeds in two steps, (a) initialization, which determines $\mathbf{v}(0)$, and (b) recursion, which determines $\mathbf{v}(i+1)$ from $\mathbf{v}(i)$, $0 \le i \le m_b-2$.

An expression for $\mathbf{v}(0)$ can be derived by summing over the rows of \mathbf{H}_{bm} to obtain

$$\mathbf{P}_{p(x,k_b)}\mathbf{v}(0) = \sum_{j=0}^{k_b-1} \sum_{i=0}^{m_b-1} \mathbf{P}_{p(i,j)}\mathbf{u}(j)$$
(1)

where x, $1 \le x \le m_b - 2$, is the row index of \mathbf{h}_{bm} where the entry is nonnegative and unpaired, and \mathbf{P}_i represents the $z \times z$ identity matrix circularly right shifted by size *i*. Equation (1) is solved for $\mathbf{v}(0)$ by multiplying by $\mathbf{P}_{p(x,k_b)}^{-1}$, and

 $\mathbf{P}_{p(x,k_b)}^{-1} = \mathbf{P}_{z-p(x,k_b)}$ since $p(x,k_b)$ represents a circular shift.

Considering the structure of $\mathbf{H}\boldsymbol{\xi}_2$, the recursion can be derived as follows,

$$\mathbf{v}(1) = \sum_{j=0}^{k_b-1} \mathbf{P}_{p(i,j)} \mathbf{u}(j) + \mathbf{P}_{p(i,k_b)} \mathbf{v}(0), \quad i = 0,$$
(2)

$$\mathbf{v}(i+1) = \mathbf{v}(i) + \sum_{j=0}^{k_b-1} \mathbf{P}_{p(i,j)} \mathbf{u}(j) + \mathbf{P}_{p(i,k_b)} \mathbf{v}(0), \quad i = 1, ..., m_b - 2$$
(3)

where

$$\mathbf{P}_{-1} \equiv \mathbf{0}_{z \times z}$$

Thus all parity bits not in $\mathbf{v}(0)$ are determined by evaluating Equation (2) for $0 \le i \le m_b - 2$.

Equations (1) and (2)to (3) completely describe the encoding algorithm. These equations also have a straightforward interpretation in terms of standard digital logic architectures. Since the non-zero elements p(i,j) of \mathbf{H}_{bm} represent circular shift sizes of a vector, all products of the form $\mathbf{P}_{p(i,j)}\mathbf{u}(j)$ can be implemented by a size-*z* barrel shifter.

2004-11-12 *Method 2*

For efficient encoding of LDPC, **H** are divided into the form

$$H = \begin{pmatrix} A & B & T \\ C & D & E \end{pmatrix}$$
(1)
where A is $(m-z) \times k (N_p - g) \times N_k$, B is $(m-z) \times z (N_p - g) \times g$, T is $(m-z) \times (m-z) (N_p - g) \times (N_p - g)$,
 C is $z \times k \cdot g \times N_k$, D is $z \times z \cdot g \times g$, and finally, E is $z \times (m-z) \cdot g \times (N_p - g)$. $\begin{pmatrix} B \\ D \end{pmatrix}$ and D correspond to the
expanded \mathbf{h}_b and $h_b (m_b - 1)$, respectively. The basic structure of the H matrix is



Further, all these matrices are sparse and T is lower triangular with ones along the diagonal. B and D part have the column degree 3 and D has shift value of a (a is an integer, $0 \le z \le 1$). B is with the shift value a of the first entry and shift value 0 in the middle of the column. This other entry is non-zero.

Let $v = (u, p_1, p_2)$ that where u denotes the systematic part, p_1 and p_2 combined denote the parity part, p_1 has length \underline{zg} , and p_2 has length $\underline{(m-z)}(N_p - g)$. The definition equation $H \cdot v^t = 0$ splits into two equations, as in equations (2) and (3) 3 and 4 namely

$$A\boldsymbol{u}^{T} + \boldsymbol{B}\boldsymbol{p}_{1}^{T} + \boldsymbol{T}\boldsymbol{p}_{2}^{T} = \boldsymbol{0}$$
 (2)

and

$$\underbrace{\left(\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{A}+\boldsymbol{C}\right)\boldsymbol{u}^{T}+\left(\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{B}+\boldsymbol{D}\right)\boldsymbol{p}_{1}^{T}=\boldsymbol{0}\left(\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{A}+\boldsymbol{C}\right)\boldsymbol{u}^{T}+\left(\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{B}+\boldsymbol{D}\right)\boldsymbol{p}_{1}^{T}=\boldsymbol{0}}_{(3)}$$

Define $\mathbf{f} := \mathbf{E}\mathbf{T}^{-1}\mathbf{B} + \mathbf{D}\mathbf{f} := -\mathbf{E}\mathbf{T}^{-1}\mathbf{B} + \mathbf{D}^{-1}$ and when we use with the parity check matrix as indicated appendix we can get, $\mathbf{f} = \mathbf{I}$. Then from (34) we conclude that

$$\underline{p_1^T = \left(ET^{-1}A + C\right)u^T} \underline{p_1^T = \left(ET^{-1}A + C\right)u^T}$$
(45)

and

$$\boldsymbol{p}_{2}^{T} = \boldsymbol{T}^{-1} \left(\boldsymbol{A} \boldsymbol{u}^{T} + \boldsymbol{B} \boldsymbol{p}_{1}^{T} \right).$$
 (56)

As a result, the encoding procedures and the corresponding operations can be summarized below and illustrated in Fig. 1.

Encoding procedure

Step 1) Compute Au^{T} and Cu^{T} . Step 2) Compute $ET^{-1}(Au^{T})$. Step 3) Compute p_{1}^{T} by $p_{1}^{T} = ET^{-1}(Au^{T}) + Cu^{T}$. 2004-11-12 **Step 4**) Compute \boldsymbol{p}_2^T by $\boldsymbol{T}\boldsymbol{p}_2^T = \boldsymbol{A}\boldsymbol{u}^T + \boldsymbol{B}\boldsymbol{p}_1^T$.



Fig. <u>12</u> Block diagram of the encoder architecture for the block LDPC code.