

# **ERL and Parameter Recommendations (updated from mellitz\_3ck\_adhoc\_01a\_030420) for Clause 162, 163 and Annex 120F, and 120G**

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# Putting ERL, Return Loss, and Reflections into Perspective

*Much of this is like finding faces in the clouds when all we care about is it raining or not.*

# Agenda

- ❑ ERL comments
  - i.e. the purpose is to resolve comments
- ❑ COM Tracking ERL
- ❑ ERL parameters recommendations
- ❑ Clause 162 COM only shows correlation to II and not ISI
  - Recommendation: The best tact here it to set the ERL limit close to the lowest ERL for channels near 3 dB COM
- ❑ Annex 120F an 120G
  - Recommendation based on Clause 163

# Comments and References

<b>Clause</b>	<b>comment</b>	<b>type</b>
162.9.3	10003	Tx Host
162.9.3.4	10009	Tx Host
162.9.4	10009	Rx Host
162.9.4.5	10011	Rx Host
162.11.3	10013	CA
162.11.3	10012	CA
163.9.1.1	10020	Tx Chip
163.9.1.1	10021	Tx Chip
163.9.2.1	10022	Rx Chip
163.10.2	10024	Channel

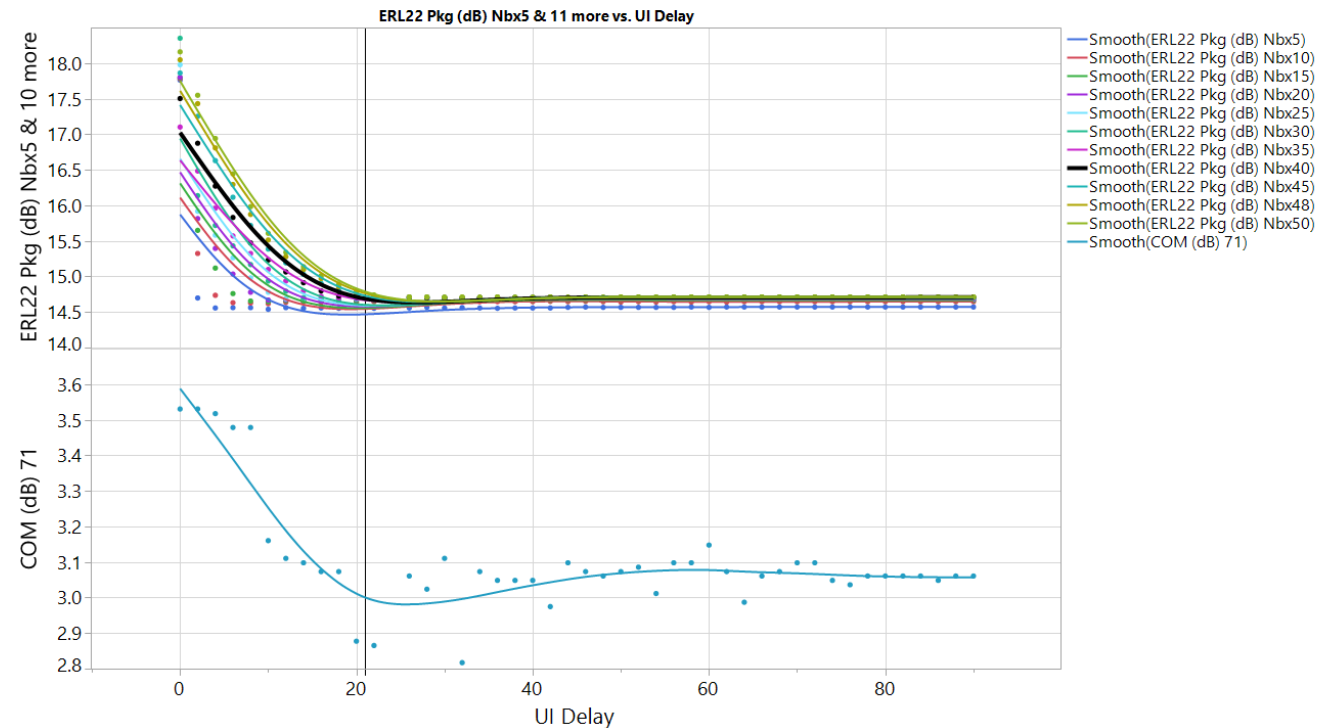
# How Well Does COM Track ERL?

❑ COM tracks ERL when channel reflections are near the device transmitter or receiver.

- ERL and COM vs. added delay track fit shapes
- The  $N_{bx}$  parameter is related of the equalizer (DFE) reach
- See back up for details

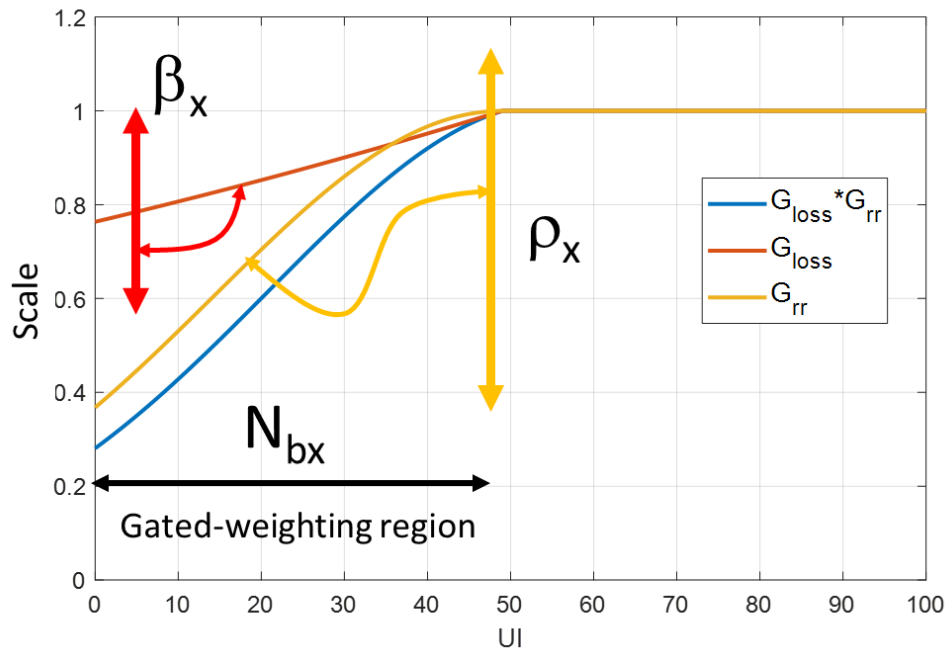
❑  $N_{bx}$  is not tied to DFE reach for computed ERL when the test point is not near TP0 or TP5

- See back up for details



# Parameter Proposal

- 0 value for  $\beta_x$
- 0.618 value for  $\rho_x$



See mellitz\_3ck\_adhoc\_01a\_030420 for more details

$$G_{rr}(t) = \begin{cases} 0 & t < T_{fx} \\ \rho_x(1 + \rho_x) \exp\left(-\frac{[(t - T_{fx})f_b - (N_{bx} + 1)]^2}{(N_{bx} + 1)^2}\right) & T_{fx} \leq t < T_{fx} + \frac{N_{bx} + 1}{f_b} \\ 1 & t \geq T_{fx} + \frac{N_{bx} + 1}{f_b} \end{cases}$$

$$G_{loss}(t) = \begin{cases} 0 & t < T_{fx} \\ 10 \frac{\frac{\beta_x}{f_b} [(t - T_{fx})f_b - (N_{bx} + 1)]}{20} & T_{fx} \leq t < T_{fx} + \frac{N_{bx} + 1}{f_b} \\ 1 & t \geq T_{fx} + \frac{N_{bx} + 1}{f_b} \end{cases}$$

Adopt N is based on electrical length from Mike Li

<b>Clause/Annex</b>	<b>type</b>	<b>N</b>
162.9.3.4	Tx Host	800
162.9.4.5	Rx Host	800
162.11.3	CA	7000
163.9.1.1	Tx Chip	200
163.9.2.1	Rx Chip	200
163.10.2	Channel	3500
120F.3.1.1	Tx Chip	200
120F.3.2.1	Rx Chip	200
120F.4.3	Channel	2000
120G.3.1.3	Host/Module	800

# $N_{bx}$ recommendations

Clause/Annex	type	$N_{bx}$
162.9.3.4	Tx Host	0
162.9.4.5	Rx Host	0
162.11.3	CA	0
163.9.1.1	Tx Chip	TBD
163.9.2.1	Rx Chip	TBD
163.10.2	Channel	TBD
120F.3.1.1	Tx Chip	TBD
120F.3.2.1	Rx Chip	TBD
120F.4.3	Channel	TBD
120G.3.1.3	Host/Module	0



$N_{bx}$  suggestions to use for consistency when reporting ERL when TBD's have not been resolved

Clause/Annex	type	$N_{bx}$
162.9.3.4	Tx Host	0
162.9.4.5	Rx Host	0
162.11.3	CA	0
163.9.1.1	Tx Chip	35
163.9.2.1	Rx Chip	35
163.10.2	Channel	35
120F.3.1.1	Tx Chip	12
120F.3.2.1	Rx Chip	12
120F.4.3	Channel	12
120G.3.1.3	Host/Module	0

$ERL/N_{bx} \approx$   
 0.066 to 0.1  
 dB/ $N_{bx}$

# ERL minimum recommendations

Clause/Annex	type	ERL min (dB) at	
		TP0a	TP0
162.9.3.4	Tx Host	TBD	TBD
162.9.4.5	Rx Host	TBD	TBD
162.11.3	CA	NA	TBD
163.9.1.1	Tx Chip	TBD	TBD
163.9.2.1	Rx Chip	TBD	TBD
163.10.2	Channel	NA	TBD
120F.3.1.1	Tx Chip	TBD	TBD
120F.3.2.1	Rx Chip	TBD	TBD
120F.4.3	Channel	TBD	TBD
120G.3.1.3	Host/Module	TBD	TBD

# Summary

- ❑ Lock down  $\beta_x$  and  $\rho_x$  for all clauses and annexes
- ❑ Adopt N for all clauses and annexes because electrical channel lengths are known
- ❑ Use  $N_{bx} = 0$  for Clause 162, Annex 120G
- ❑ Keep  $N_{bx} = \text{TBD}$  in the next draft for Clause 163 and Annexes 120F and 120G
- ❑ Adopt  $\text{ERL}_{\min}$  as TBD for all clause and channels
  - Confirmation and proposals from vested parties required

**THANK YOU !**

# Backup Data

# Background

- ❑ Reviewing all the ERL presentations over the past year suggest determining simultaneously optimizing ERL parameters ( $\beta_x$ ,  $\rho_x$ ,  $N_{bx}$ ) and ERL limits in the context of performance (COM or  $\Delta$ COM) for all possible channels and packages is a daunting problem
- ❑ Data also suggest single settings for ERL parameters are useful (*mellitz\_3ck\_adhoc\_01a\_030420*)
  - ERL depends on independent parameters,  $\beta_x$  and  $\rho_x$ 
    - ERL strongly tracks  $\rho_x$  which is only weakly scaled by  $\beta_x$
  - Treat  $\rho_x$  and  $\beta_x$  as just computational parameters and lock down for all ports
    - This simplification moves the project along
  - Then set the ERL limit accordingly
- ❑ For Clause 163 and Annex 120G DFE reach tracks COM and ERL through  $N_{bx}$

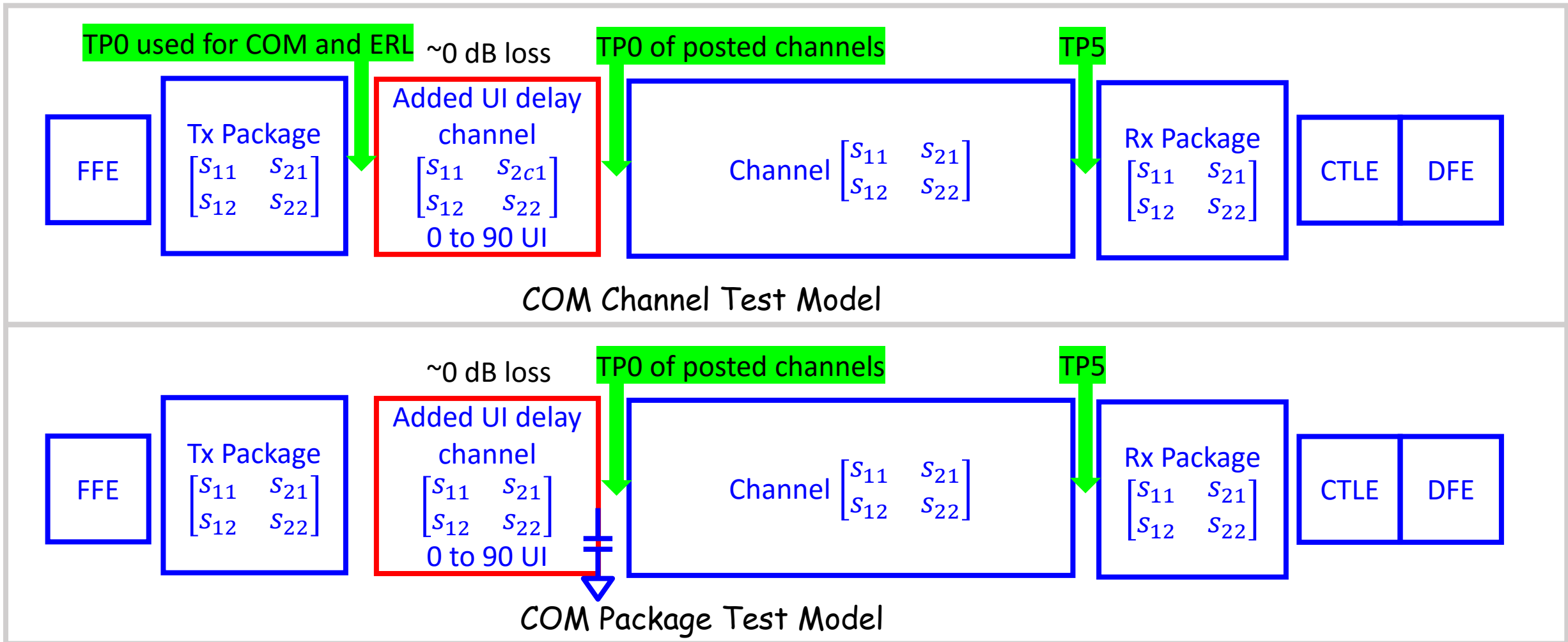
# Does ERL track $\Delta$ COM and COM

- ❑ The answer is “It depends” 😞
- ❑ For Tx/Rx KR and C2C packages answer is yes
- ❑ For channels where reflections are near the end of the channel the answer is yes
- ❑ For all others the tracking is looser and dominated by end terminations
- ❑ The best tact here it to set the ERL limit close to the lowest ERL for channels near 3 dB COM

ERL and COM tracking for bga(tp0) to bga(tp5)



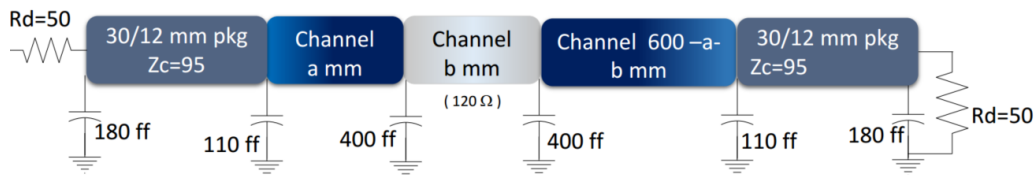
# Package and Channel COM added delay experiments



# Similar to ERL experiment done in mellitz\_3cd\_01b\_1117

Would changing the amount of reflection alter the trends of the results?

- Change the capacitor to 400 ff from 200 ff
- Shorten the total length to 450 mm from 600 mm to keep around 3 dB of COM

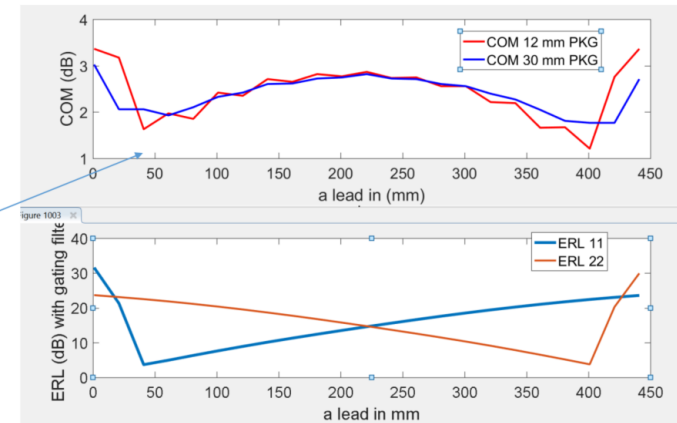


mm	ff	$\Omega$	mm	ff	mm	ff	mm	ff	mm	ff	$\Omega$	ff	$\Omega$	
30	180	95	12	110	a	400	b	400	450-a-b	110	30	95	180	50

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Similar trends regardless of reflection magnitude



$T_b(N_b + 1)$

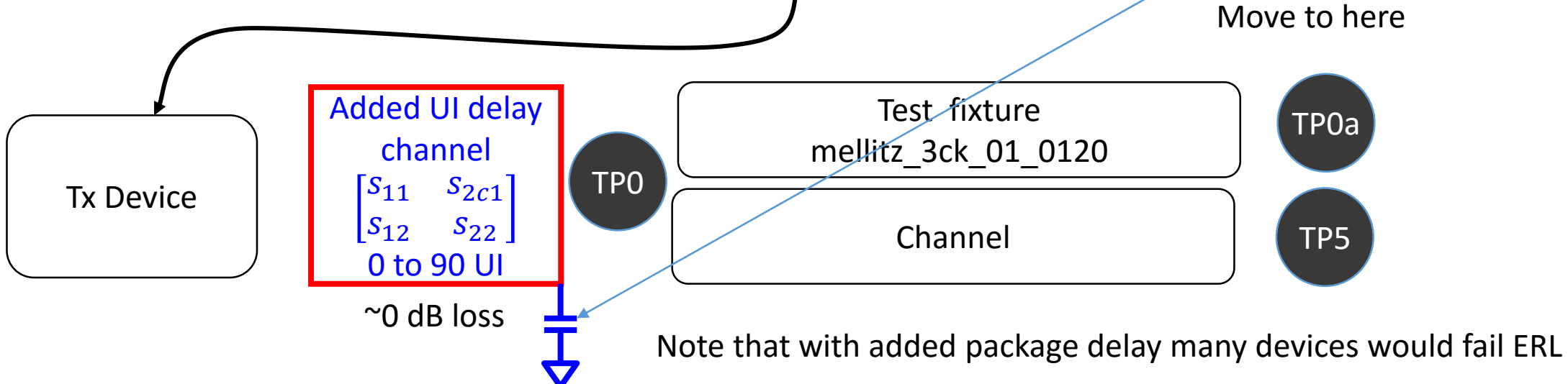
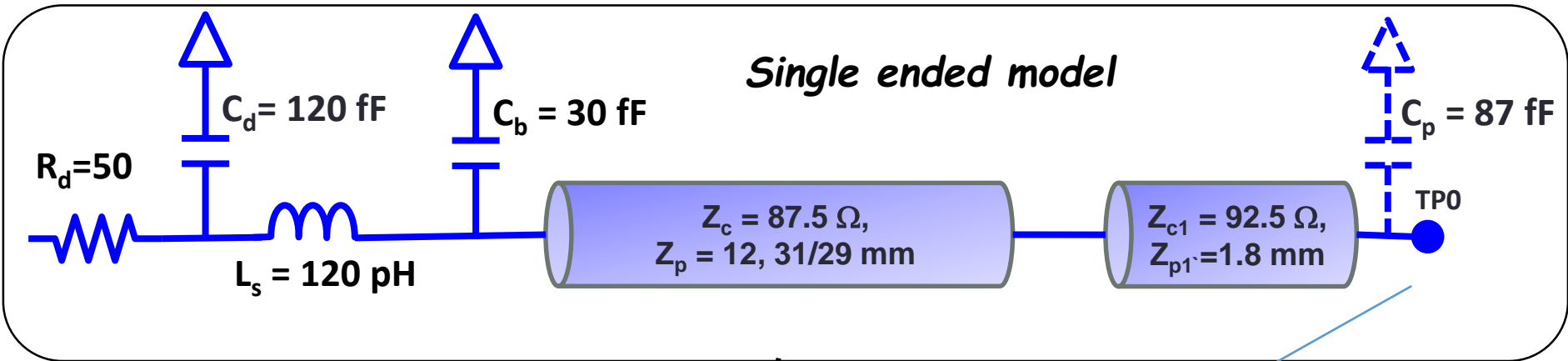
- Takeaway:**
- Similar effects on either end of the channel
  - A key factor is the distance a channel reflection is from either Tx or Rx package

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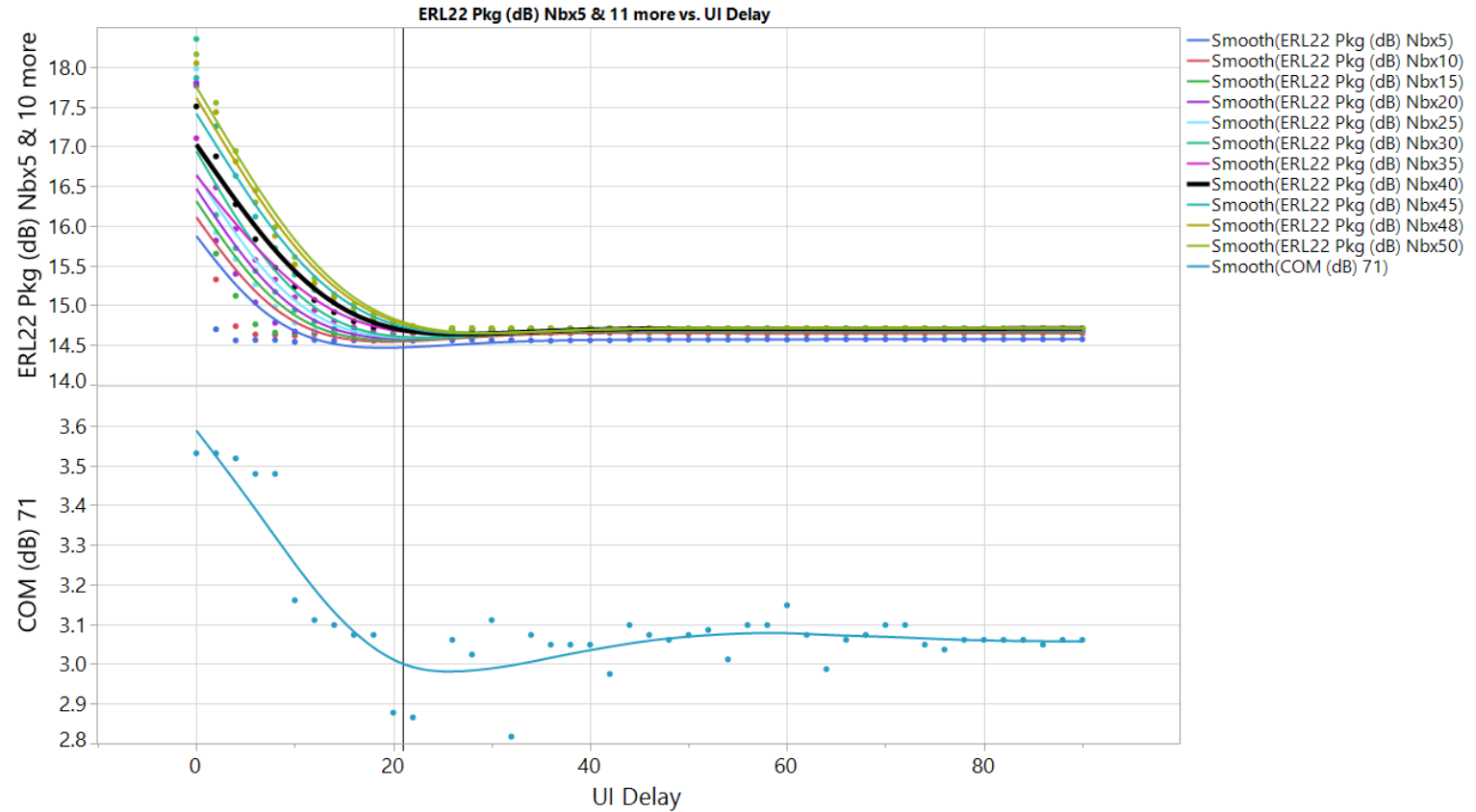
# Package added delay experiment.

**Use this to test the  $N_{bx}$  Hypothesis for packages**



# Added Package delay vs COM and ERL of package for selections of $N_{bx}$

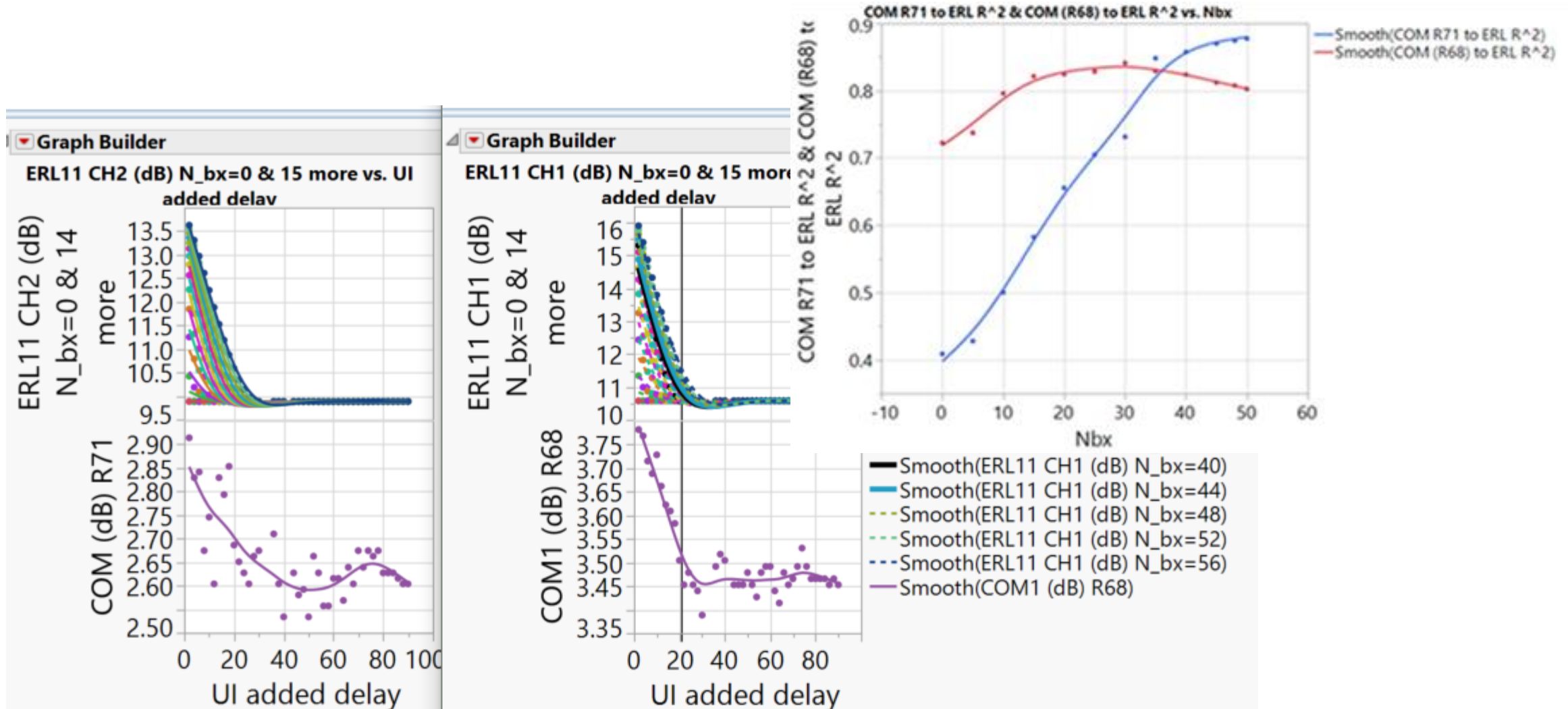
Visual estimate suggests  $N_{bx}$  should be about 42



# A more pragmatic method uses $R^2$

- ❑ If we use all the COM values vs delay for R71 and R68. That gives me 2 vectors,  $\text{COM}(R\#, \text{UI}(0:1:90))$
- ❑ For each case (either R71 or R68) determine ERL for each delay case for a give  $N_{bx}$ , where  $N_{bx} = 0:2:60$  (31 cases)
- ❑ For each case that results in 31 vectors of ERL vs delay  $\text{ERL}(R\#, \text{UI}(0:1:90), N_{bx})$
- ❑  $R^2$  is correlation for each case between  $\text{COM}(R\#, \text{UI}(0:1:90))$  and delay  $\text{ERL}(R\#, \text{UI}(0:1:90), N_{bx})$
- ❑ Then we plot  $R^2$  vs  $N_{bx}$  for each case.
- ❑ Computation for ERL used Annex 93A.5 with  $\beta_x=0$ ,  $\rho_x=-0.618$ , and  $N = 2000$

# A more pragmatic approach place $N_{bx}$ at 35



# CL 163 Rx/Rx ERL estimate for package reference package variations

# ERL and Pmax/Vf of reference package at TP0a

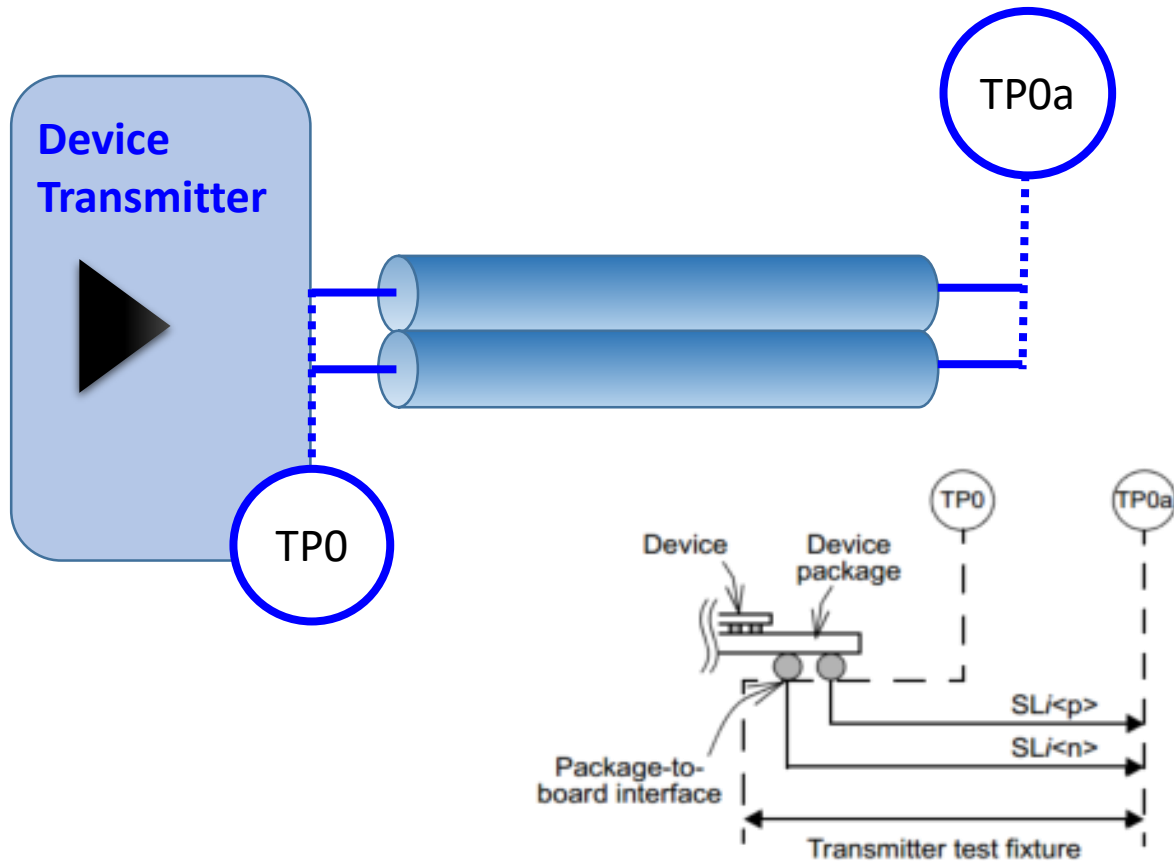


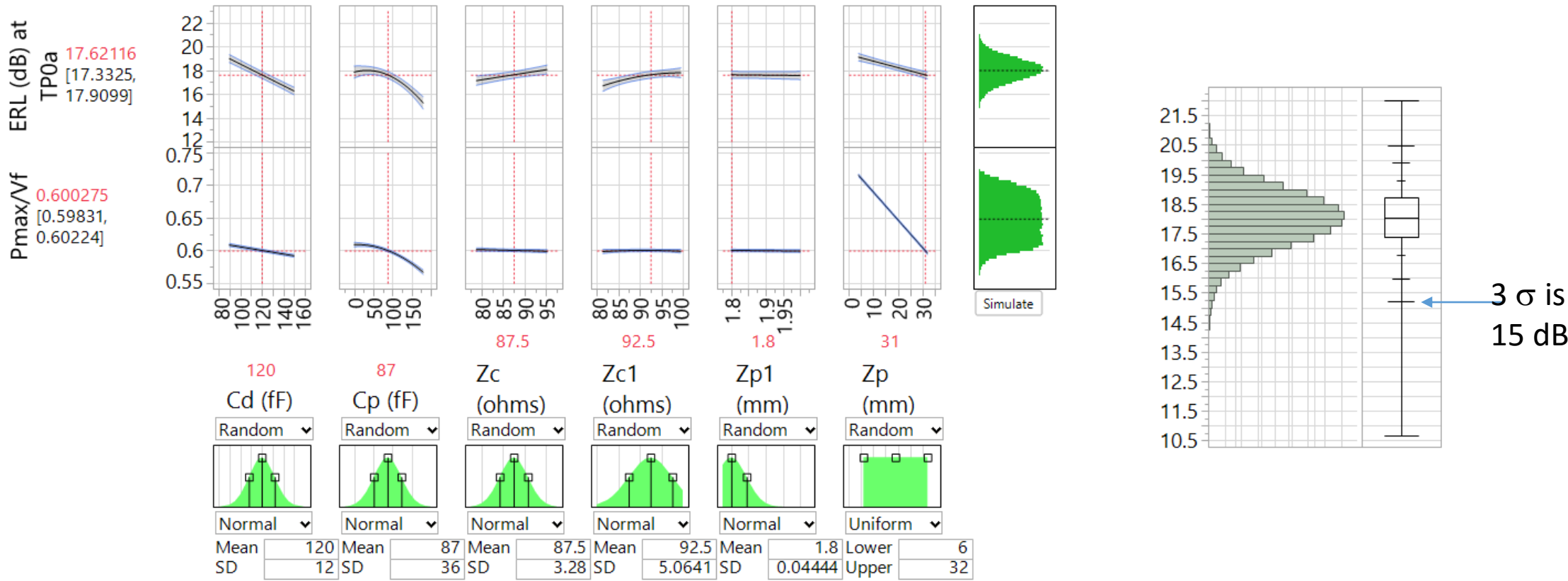
Figure 93-5—Transmitter test fixture and test points

Zp	Zp Pmax/Vf TP0a	Zp Pmax/Vf TP0	ERL (dB) Tp0a	ERL (dB) Tp0
				$N_{bx}=42$
12 mm	0.68	0.727	18.28	14.28
31 mm	0.540	0.577	17.67	14.72

Zp	Zp Pmax/Vf TP0a	Zp Pmax/Vf TP0	ERL (dB) Tp0a	ERL (dB) Tp0
				$N_{bx}=48$
12 mm	0.68	0.727	18.55	14.5
31 mm	0.540	0.577	18.07	15.05



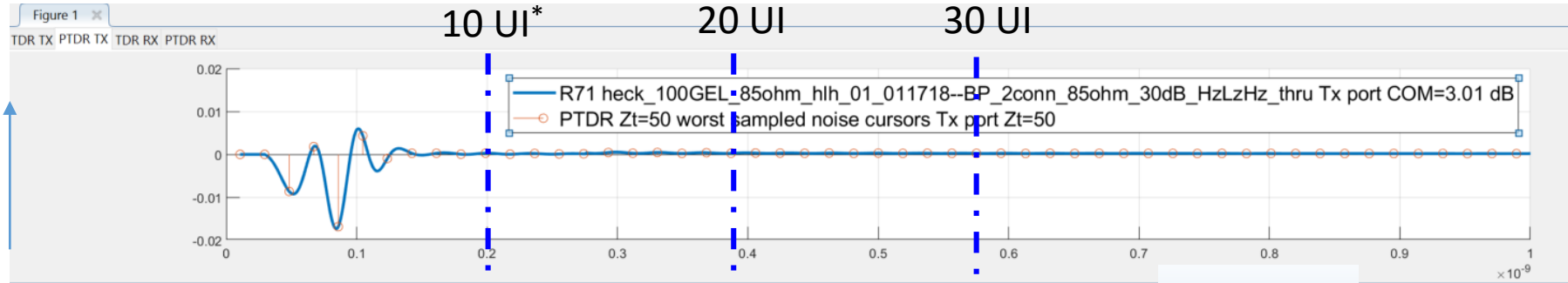
# Distribution suggest Tx/Rx ERL limits at tp0a



# Selecting channel with reflection near TP0 for experimenting

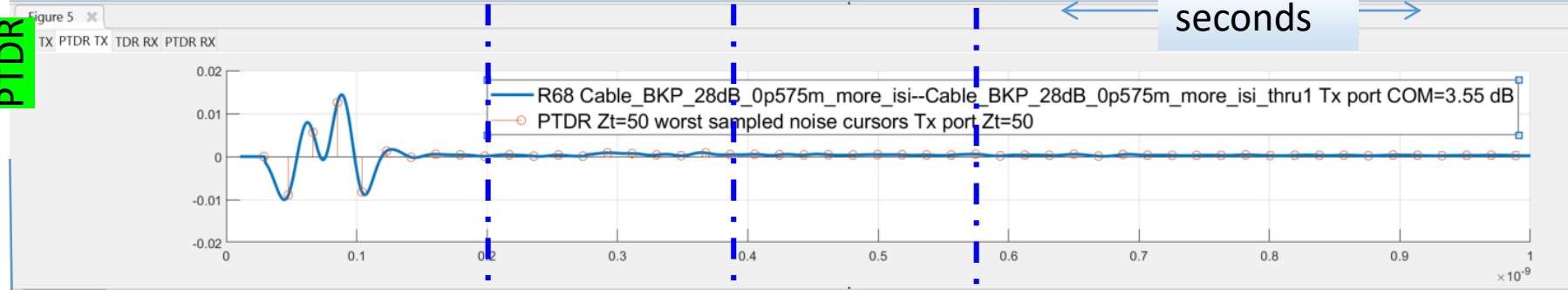
# PTDR of Clause 163 Channels with Reflections near TPO and with COM ~ 3 dB

Larger reflections

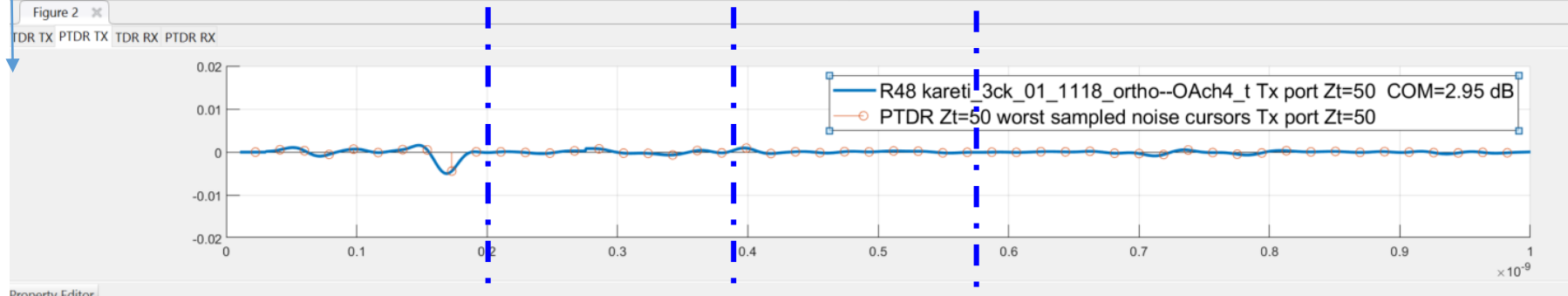


PTDR

Larger reflections



Smaller reflections



seconds

\*Delay through 30 mm package is ~ 10 UI

# PTDR the rest of the story: All the channel have reflections



There are reflection away from TP0



Most all return loss is all comprehended in COM

# Kappa ( $\kappa$ ) may be used to determine impact of reflections (from healey\_3ck\_01a\_0120 )

- The transfer function from the transmitter input to the receiver output is the following

$$H_{21} = \frac{s_{21}^{(t)} s_{21}^{(r)} s_{21}^{(r)}}{1 - s_{22} s_{11}^{(r)} - s_{11} s_{22}^{(t)} + s_{11}^{(r)} s_{22}^{(t)} \Delta S}$$

$$\Delta S = s_{11} s_{22} - s_{21} s_{12}$$

$$s_{21} = s_{12}$$

- Note that  $1/(1+x) \cong 1-x+x^2-\dots$  for  $|x| < 1$

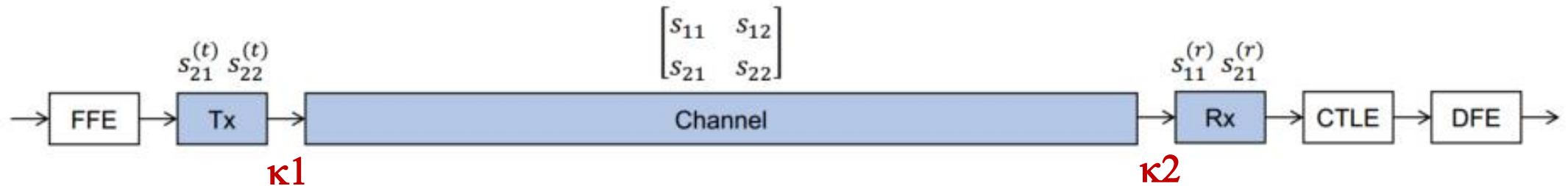
$$H_{21} \cong s_{21}^{(t)} s_{21}^{(r)} s_{21}^{(r)} \left( 1 + \frac{\kappa_2}{s_{22} s_{11}^{(r)}} + \frac{\kappa_1}{s_{11} s_{22}^{(t)}} + \frac{\kappa_1 \kappa_2}{s_{21} s_{21}^{(t)}} \right)$$

If  $\kappa_1$  and  $\kappa_2$  are zero, the reflection at port 1 and 2 are not permitted to occur  
 $\kappa_1$  and  $\kappa_2$  can also apportion reflection

Rx-Tx re-reflection  
 Tx re-reflection  
 Rx re-reflection

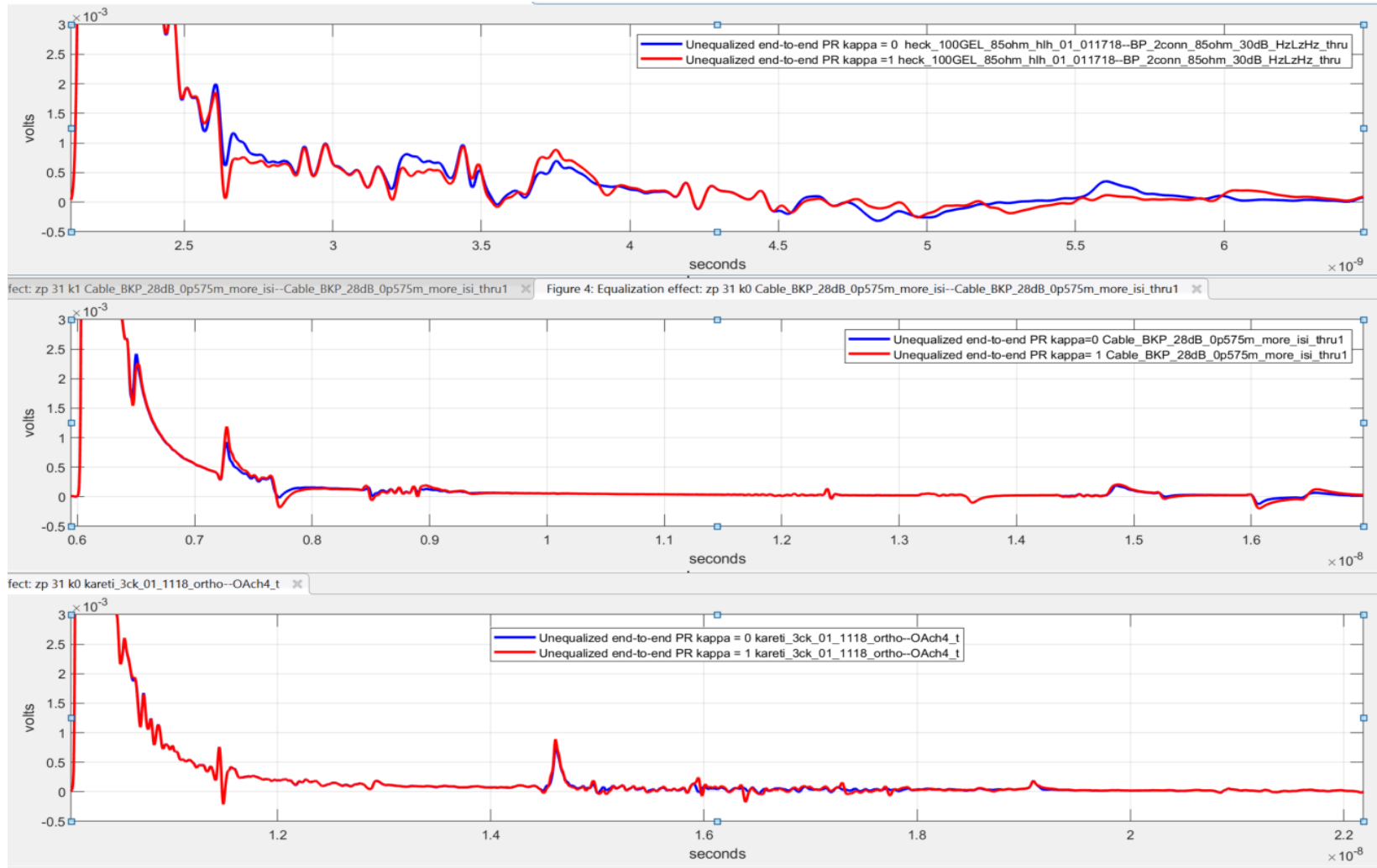
- Constraints on  $s_{11}$ ,  $s_{22}$ ,  $s_{11}^{(r)}$ , and  $s_{22}^{(t)}$ , e.g., ERL, are imposed to limit the re-reflection

Delta COM is the difference between COM with:  
 $\kappa_1$  and  $\kappa_2$  set to one minus COM with  $\kappa_1$  and  $\kappa_2$  set to some other value

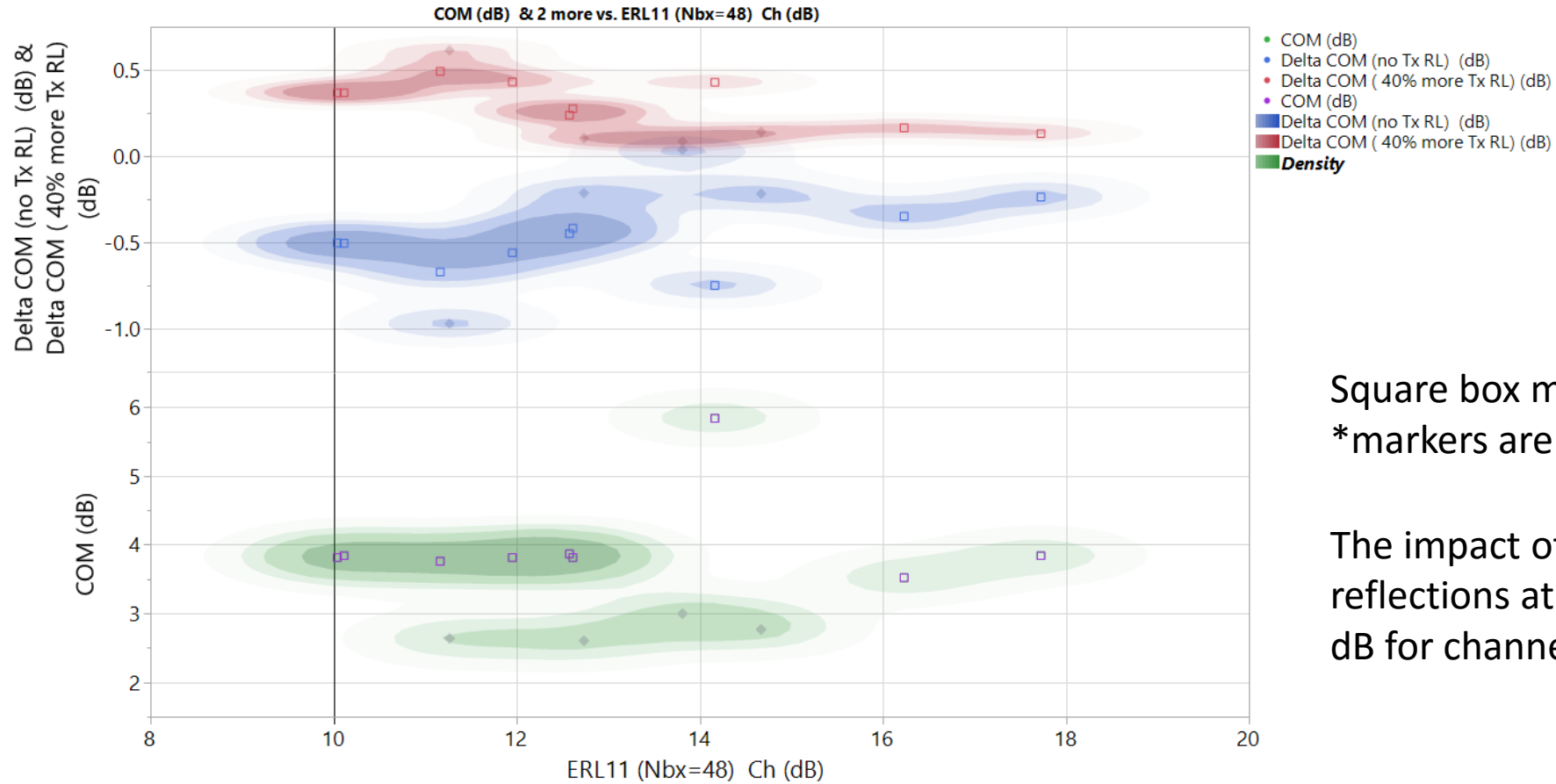


# Delta COM

$\kappa = 0$  ( no tp0 reflection) and  $\kappa=1$  seen at the receiver at tp5



# Start with: $ERL_{\min} = 10$ dB



Square box markers are COM > 3.5 dB  
\*markers are COM ≤ 3dB

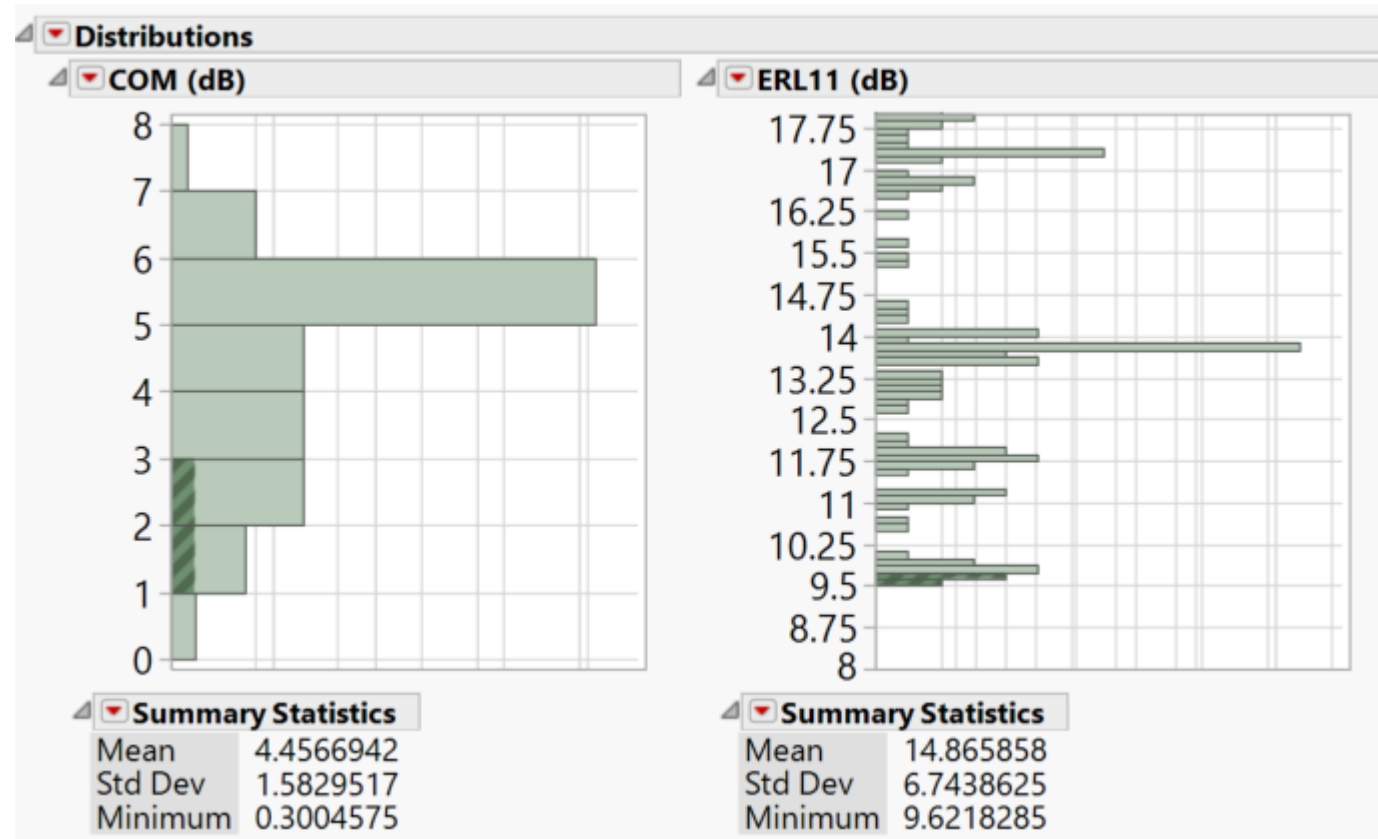
The impact of with and without reflections at  $tp_0$  is only about  $\frac{1}{2}$  dB for channels which pass COM



# All channel with lower than 9.75 dB ERL Fail COM

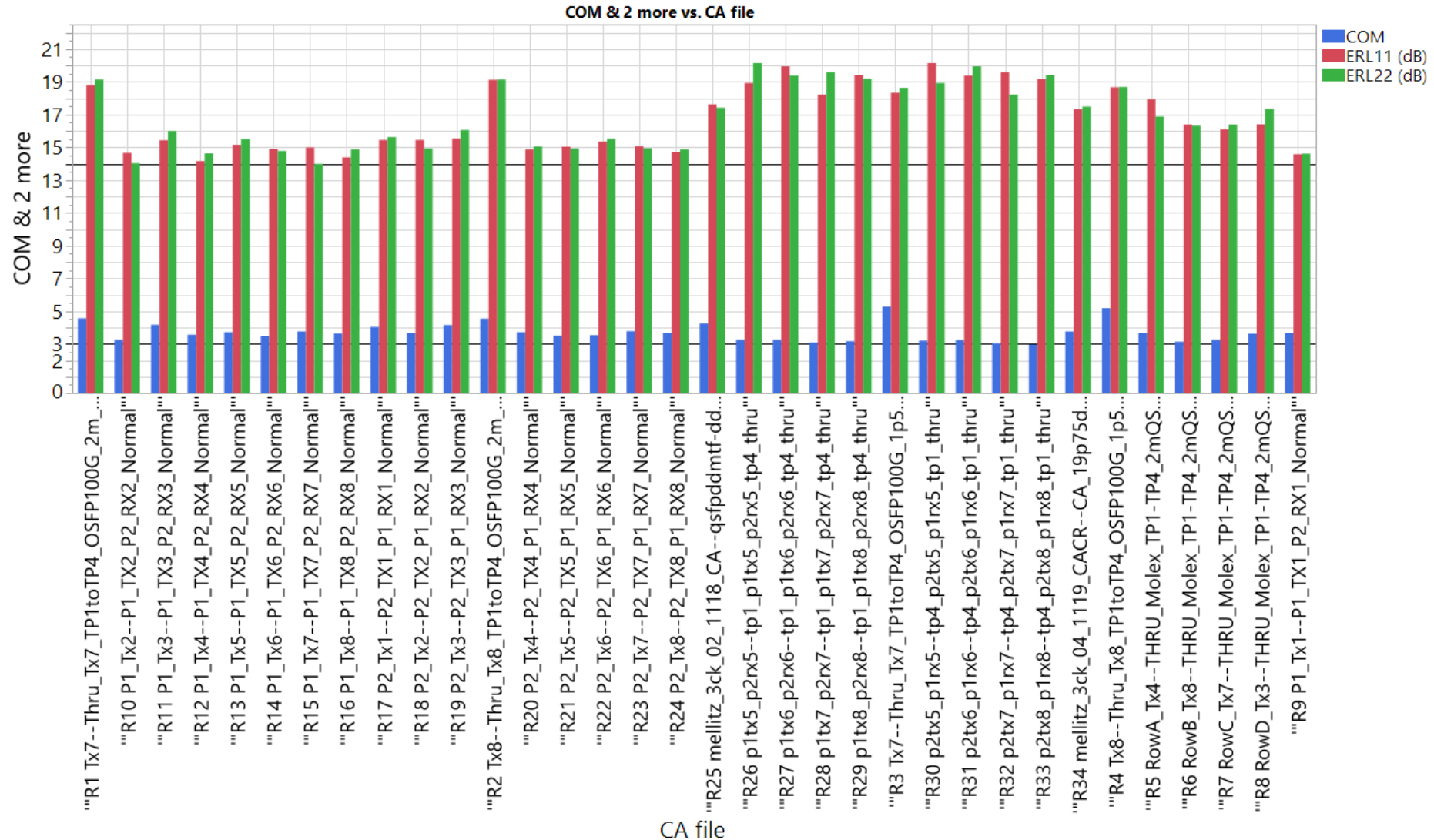
Computed for  
 $N_{bx}=42$

ERL/ $N_{bx}$  ratio  $\approx$   
0.06 dB/ $N_{bx}$



# Cable Assembly Data

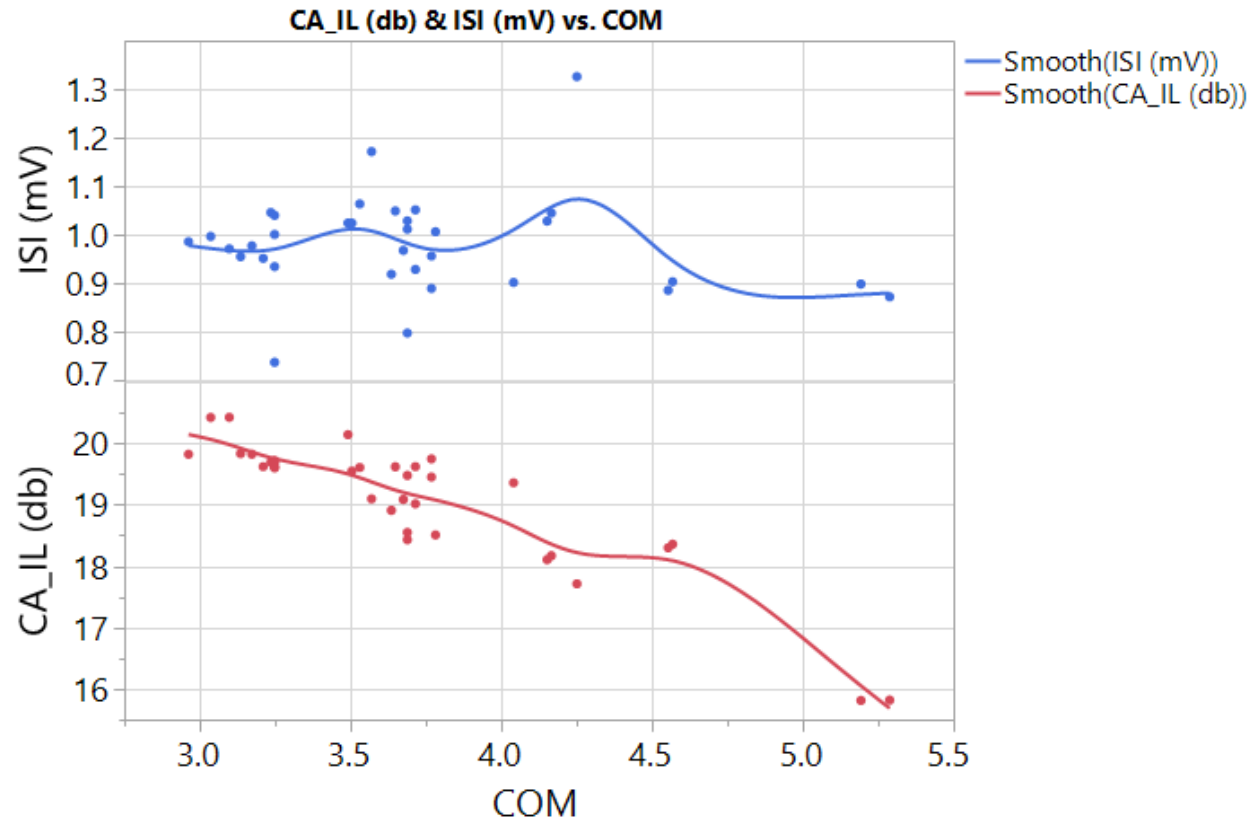
# CA data COM and ERL (Nbx=42)



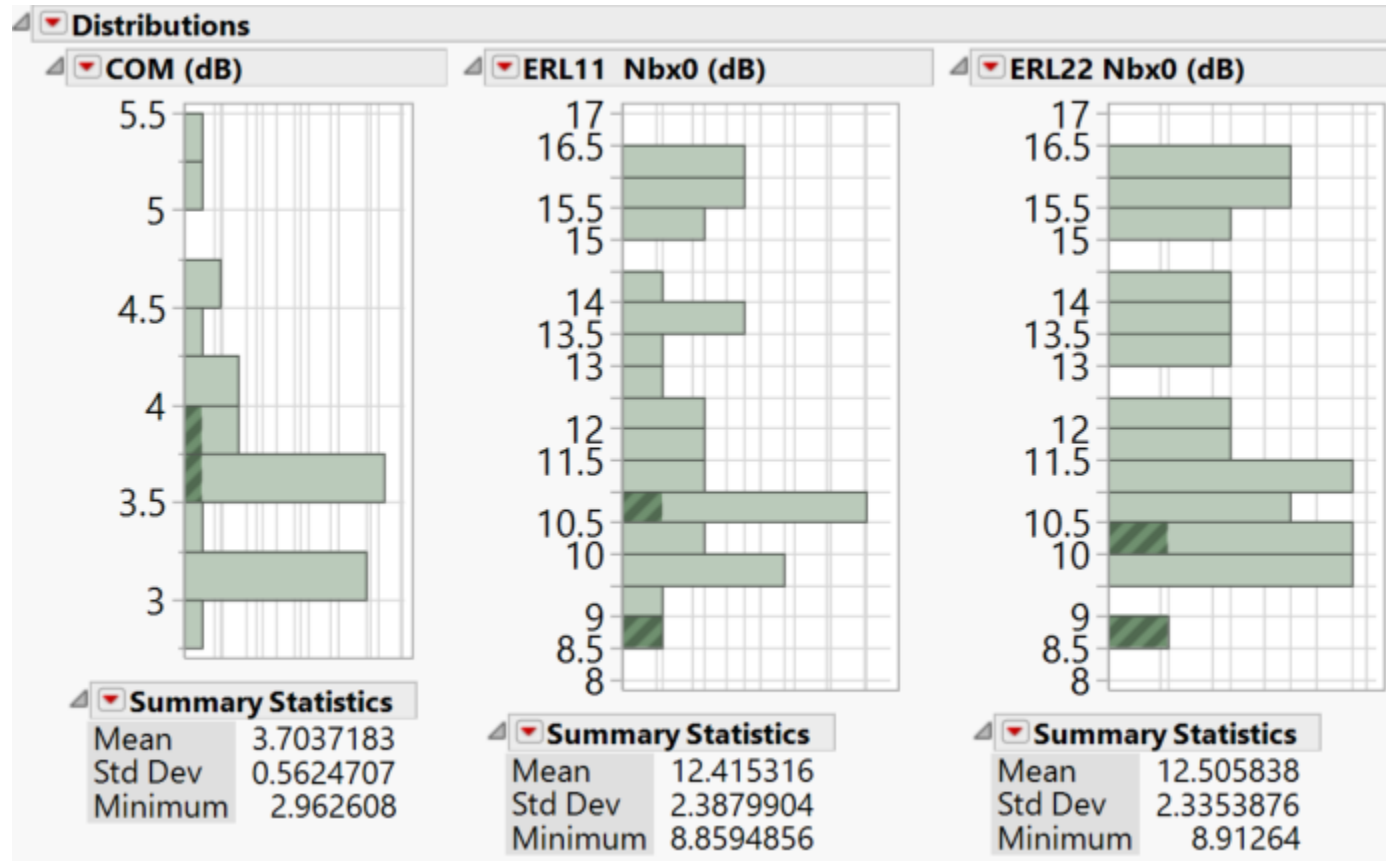
# COM vs IL and ISI ... correlations

COM correlated to  
Cable IL

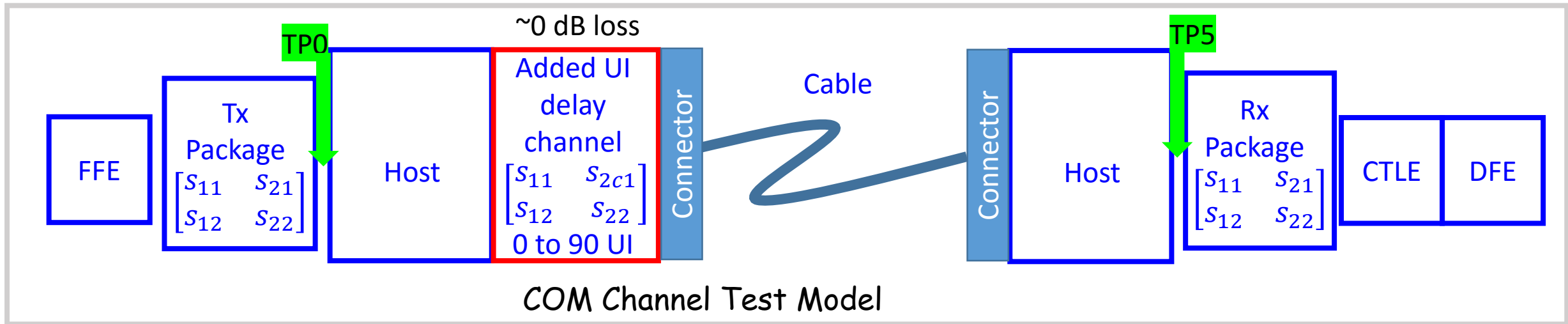
COM not  
correlated to ISI



Based on All posted CA channel ERL min looks like 8.5 dB with  $N_{bx}=0$



# DAC Cable Delay experiment

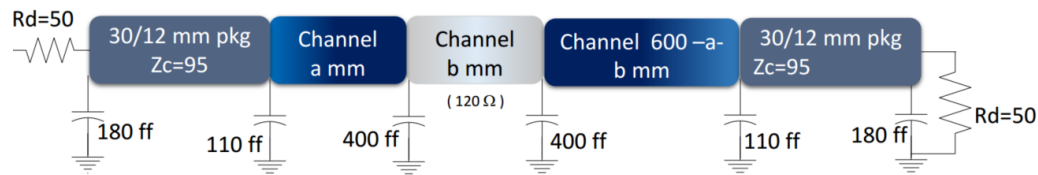


- ❑ Similar experiment what was done for packages and KR channels
- ❑ A similar experiment was done during .3ck ERL development

# Look back at the ERL experiment in mellitz\_3cd\_01b\_1117

Would changing the amount of reflection alter the trends of the results?

- Change the capacitor to 400 ff from 200 ff
- Shorten the total length to 450 mm from 600 mm to keep around 3 dB of COM

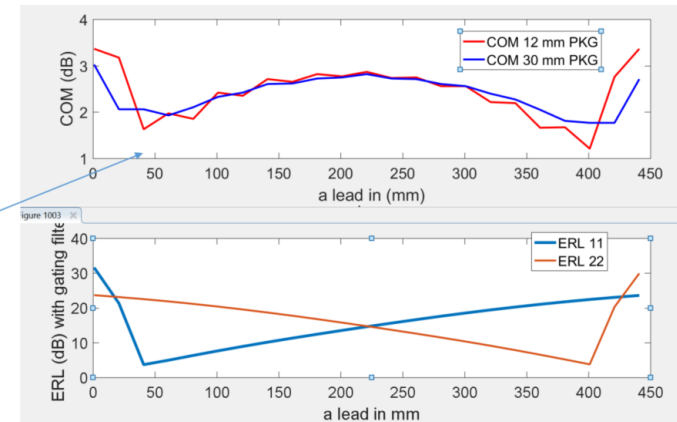


mm	ff	Ω	mm	ff	mm	ff	mm	ff	mm	ff	mm	Ω	ff	Ω
30	180	95	12	110	a	400	b	400	450 -a-b	110	30	95	180	50

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Similar trends regardless of reflection magnitude



$T_b(N_b + 1)$

**Takeaway:**

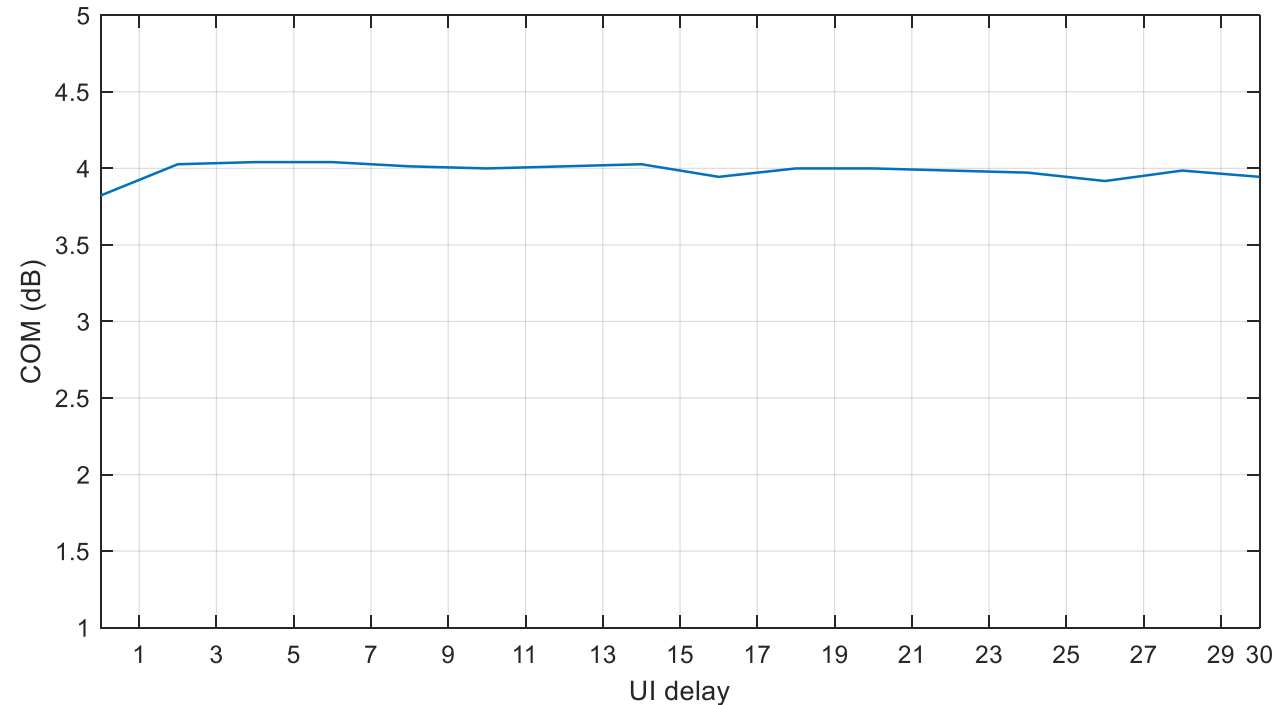
- Similar effects on either end of the channel
- A key factor is the distance a channel reflection is from either Tx or Rx package

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**Hypothesis: Based on the above there should be little affect from the added delay aside from the expected resonances**

As expect there is little impact from the delay and no obvious impact from the DFE reach



Supports the assertion that  $N_{bx}$  should be 0 for CR  
Since the ERL is measured a far distance from the devices



# ERL minimum values to be reviewed by vested parties

Clause/Annex	type	ERL min (dB) at	ERL min (dB )at
		TP0a	TP0
162.9.3.4	Tx Host	TBD	TBD
162.9.4.5	Rx Host	TBD	TBD
162.11.3	CA	NA	8.5
163.9.1.1	Tx Chip	15	13
163.9.2.1	Rx Chip	15	13
163.10.2	Channel	NA	9.75
120F.3.1.1	Tx Chip	13.5	11.5
120F.3.2.1	Rx Chip	13.5	11.5
120F.4.3	Channel	TBD	TBD
120G.3.1.3	Host/Module	TBD	TBD