

# **Significance of Including Quantization Noise in COM Evaluation**

**Hossein Shakiba  
Huawei Technologies Canada  
May 2024**

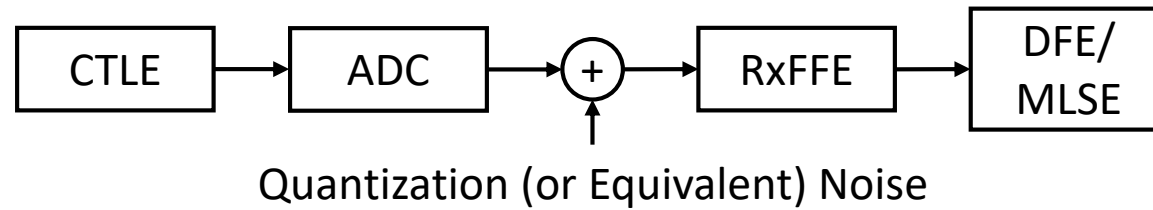
# Outline

- Introduction
- Importance of Quantization Noise
- Quantization Noise Model
- Test Channels
- Test Data
- Quantization Noise Compared to Scaling  $\eta_0$
- Summary and Conclusion

# Introduction

- Majority of the recent receivers for high-loss channels use ADC
- Receiver DSP performs heavy equalization (e.g. FFE and DFE/MLSE)
- Quantization noise is currently not a part of the overall noise during COM evaluation
- Quantization noise affects link performance and COM results:
  - 1) Reduces SNR
  - 2) Influences optimization of equalization (equalization balance among TxFFE, CTLE, RxFFE, and DFE/MLSE)
  - 3) Affects noise coloring (becoming increasingly important in the new MMSE RxFFE method and MLSE)
- Tweaking existing noise components (such as `eta_0`) is not a good approach
- Even without an ADC, it is a good practice to include additional noise before FFE+DFE/MLSE
- A set of 112 channels is used for generating test data
- COM version: `com_ieee8023_93a_450beta3_hs2p0` (see backup slide for COM config)
  - ❖ `_hs2p0` customizes COM to add quantization noise

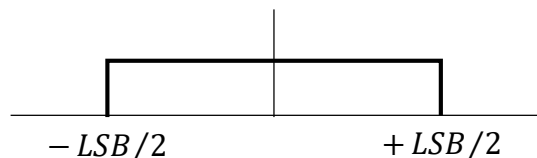
# Importance of Quantization Noise



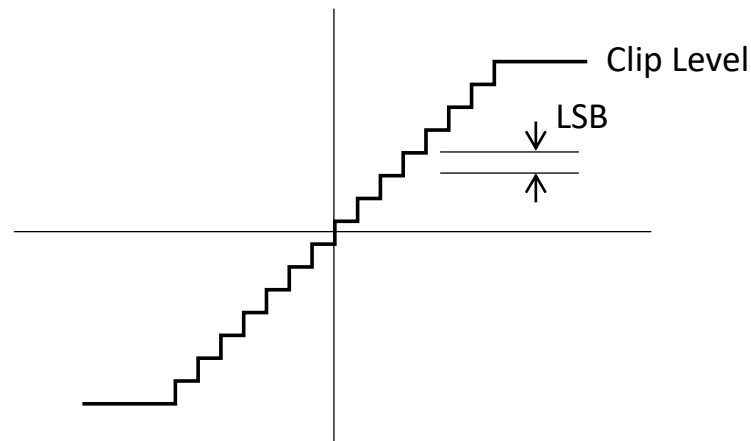
- Since the reference CTLE is a low-order filter (only one high-frequency pole-zero), in the absence of quantization noise, the optimizer tends to favor RxFFE over CTLE
- CTLE equalization will be marginalized to mostly only low-frequencies, for which FFE does not have enough taps
- As a result of CTLE under-utilization, its output could be severely under-equalized
- This forces a large input dynamic range, hence increased number of required ADC bits
- In the absence of ADC (or if it is after RxFFE), considering an 'equivalent' noise between CTLE and FFE is still reasonable and helpful

# Quantization Noise Model

- Uniformly distributed over  $-LSB/2$  to  $+LSB/2$



$$\sigma_{qn} = \frac{LSB}{\sqrt{12}}$$



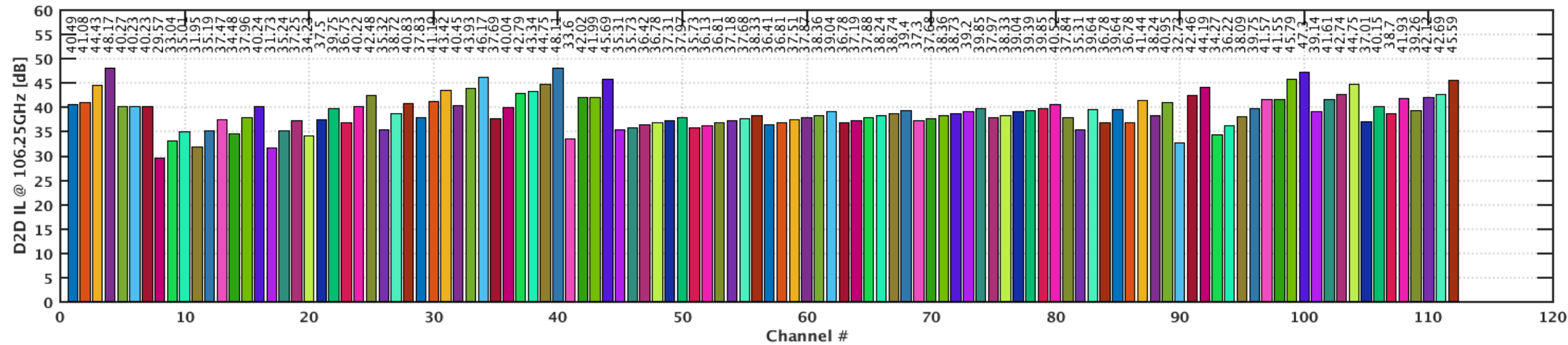
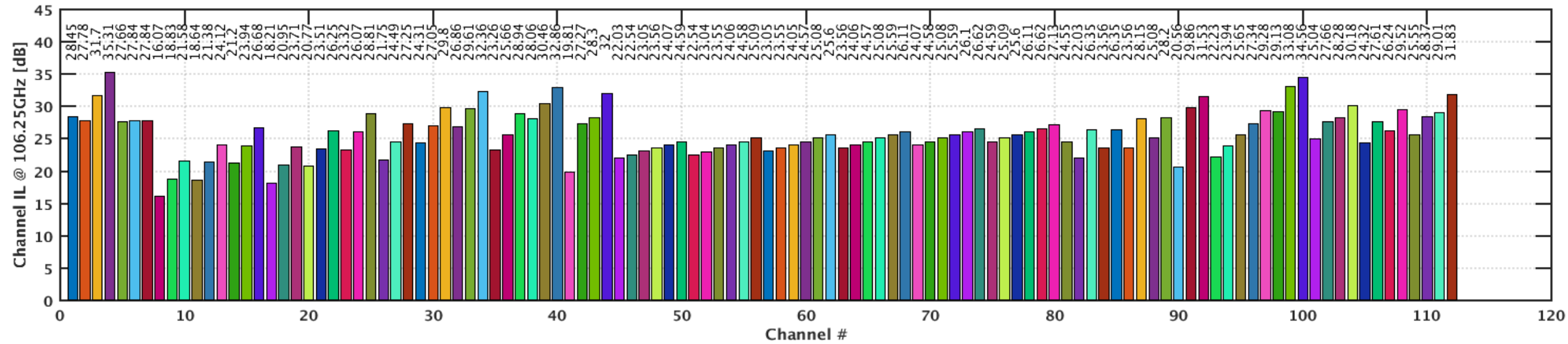
- To optimally utilize the ADC input dynamic range, its clip level is set so that clipping does not happen too frequently
- “too frequently” is defined relative to the target error rate
- ADC clip level is chosen so that clipping frequency is equal to the target error rate
- At this early stage, the analysis assumes ADC delivers  $\overline{ENOB}$  bits on “average”
- In future, actual ENOB may be used that results in quantization noise being colored

# Test Channels

- Same set as in [shakiba\\_3dj\\_01a\\_2403.pdf](#)

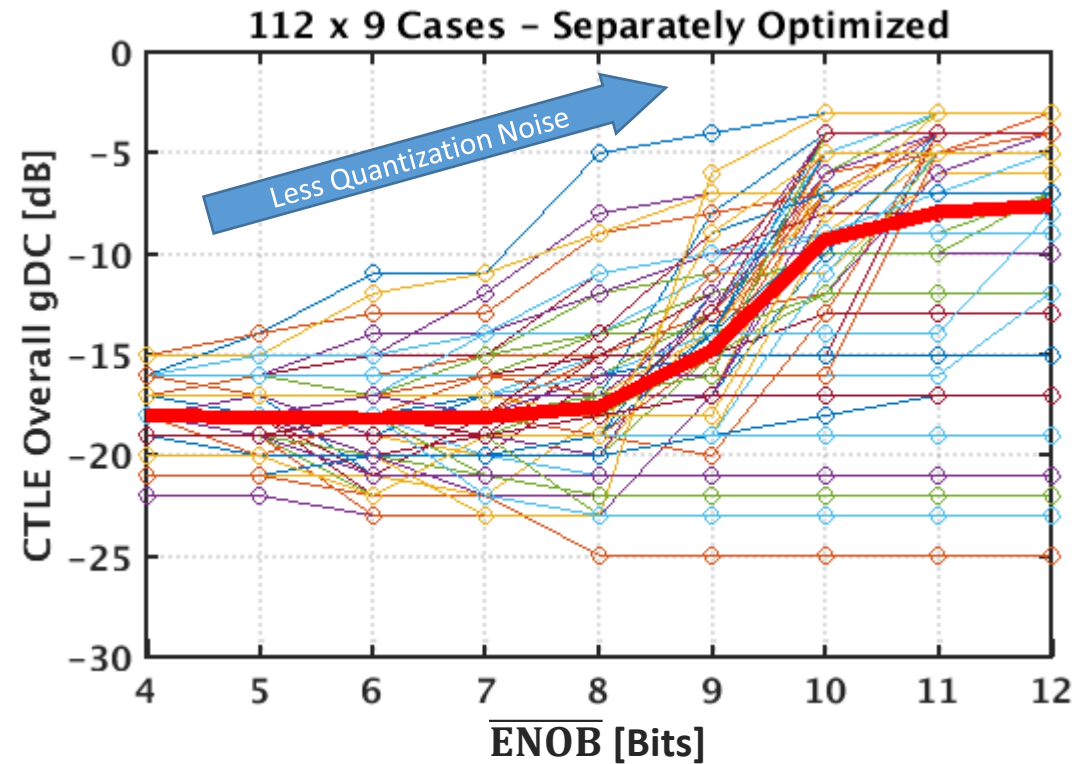
Channel #	Channel Source
1	<a href="https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_03_230629.zip">https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_03_230629.zip</a>
2	<a href="https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_04_230629.zip">https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_04_230629.zip</a>
3 – 7	<a href="https://www.ieee802.org/3/dj/public/tools/CR/kocsis_3dj_02_2305.zip">https://www.ieee802.org/3/dj/public/tools/CR/kocsis_3dj_02_2305.zip</a>
8 – 34	<a href="https://www.ieee802.org/3/dj/public/tools/KR/mellitz_3dj_02_elec_230504.zip">https://www.ieee802.org/3/dj/public/tools/KR/mellitz_3dj_02_elec_230504.zip</a>
35 – 40	<a href="https://www.ieee802.org/3/dj/public/tools/CR/shanbhag_3dj_01_2305.zip">https://www.ieee802.org/3/dj/public/tools/CR/shanbhag_3dj_01_2305.zip</a>
40 – 44	<a href="https://www.ieee802.org/3/dj/public/tools/KR/shanbhag_3dj_02_2305.zip">https://www.ieee802.org/3/dj/public/tools/KR/shanbhag_3dj_02_2305.zip</a>
45 – 80	<a href="https://www.ieee802.org/3/dj/public/tools/KR/weaver_3dj_02_2305.zip">https://www.ieee802.org/3/dj/public/tools/KR/weaver_3dj_02_2305.zip</a>
80 – 88	<a href="https://www.ieee802.org/3/dj/public/tools/KR/weaver_3dj_elec_01_230622.zip">https://www.ieee802.org/3/dj/public/tools/KR/weaver_3dj_elec_01_230622.zip</a>
89	<a href="https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_07_2309.zip">https://www.ieee802.org/3/dj/public/tools/CR/lim_3dj_07_2309.zip</a>
90 – 96	<a href="https://www.ieee802.org/3/dj/public/tools/KR/akinwale_3dj_01_2310.zip">https://www.ieee802.org/3/dj/public/tools/KR/akinwale_3dj_01_2310.zip</a>
97 – 100	<a href="https://www.ieee802.org/3/dj/public/tools/CR/akinwale_3dj_02_2311.zip">https://www.ieee802.org/3/dj/public/tools/CR/akinwale_3dj_02_2311.zip</a>
101 – 112	<a href="https://www.ieee802.org/3/dj/public/tools/CR/weaver_3dj_02_2311.zip">https://www.ieee802.org/3/dj/public/tools/CR/weaver_3dj_02_2311.zip</a>

# Test Channels



# Test Results – CTLE Utilization

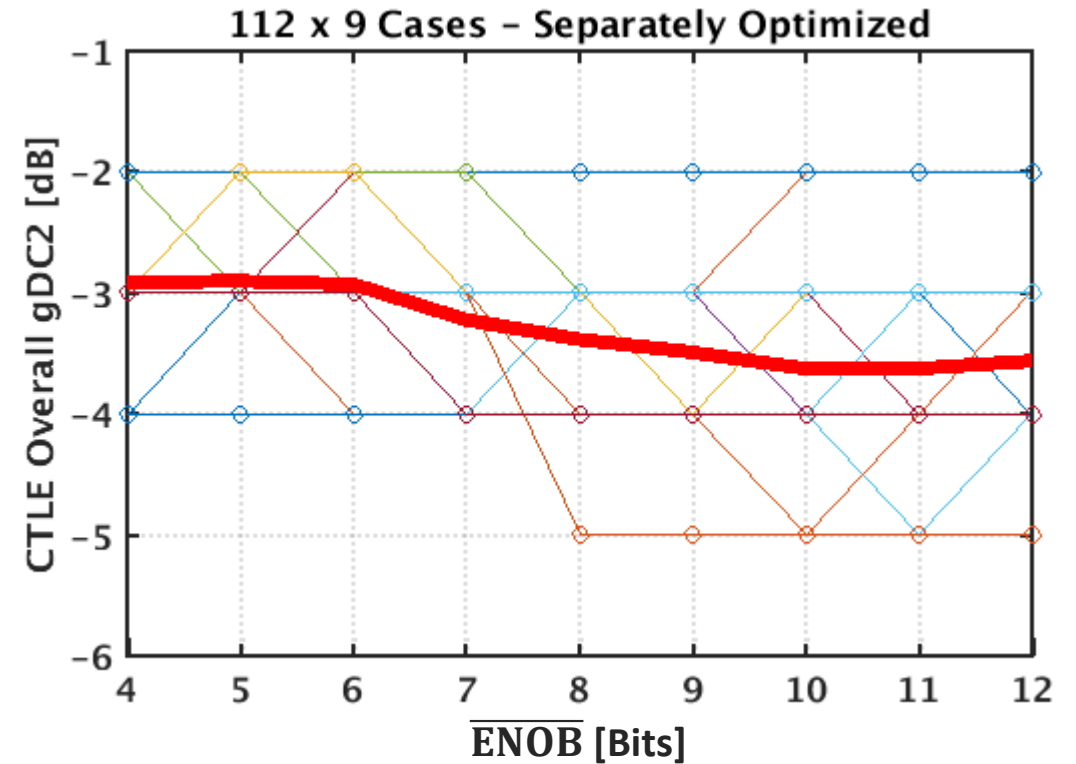
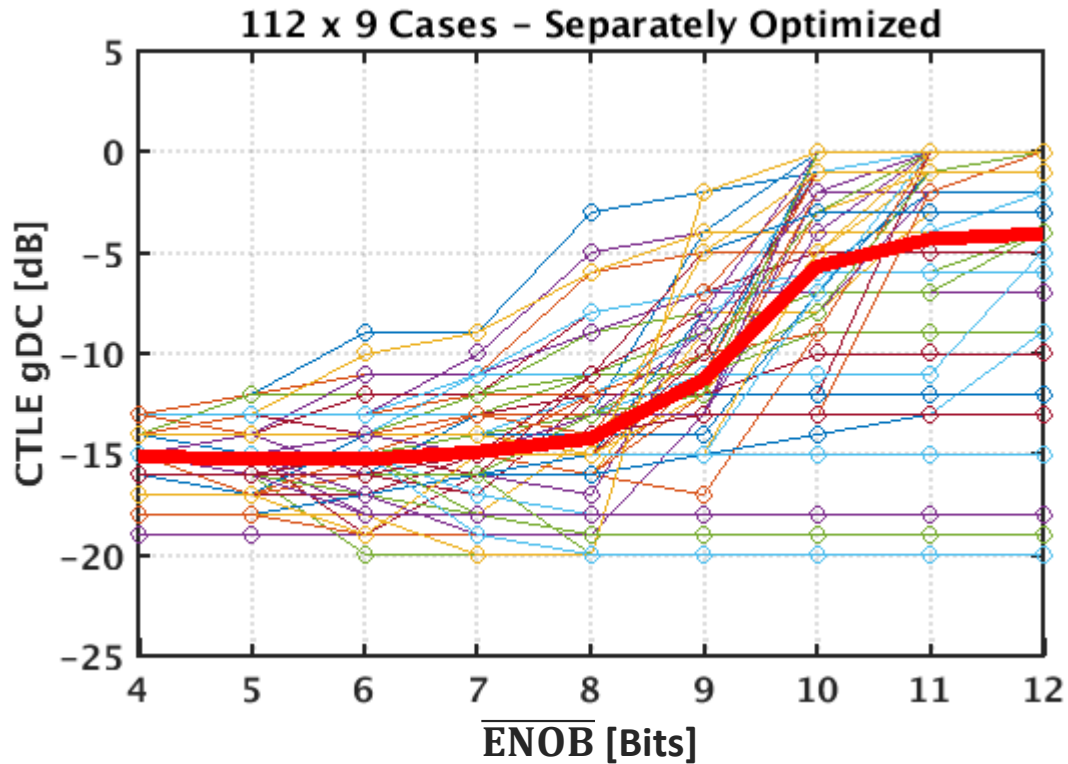
- COM was run on 112 test channel cases, and for each case 9 times for  $\overline{\text{ENOB}}$  range of 4:1:12 bits



- Finer quantization (larger  $\overline{\text{ENOB}}$ ) pushes more equalization to RxFFE and less to CTLE
- CTLE utilization on average reduces from  $\sim 18\text{dB}$  to  $\sim 7.6\text{dB}$  as  $\overline{\text{ENOB}}$  increases from 4 to 12
  - This is what happens if quantization noise is ignored

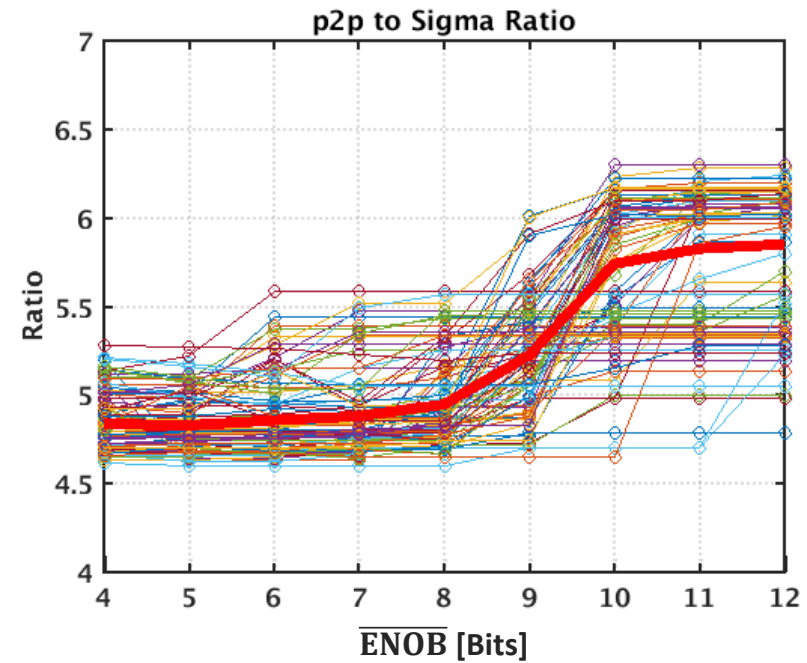
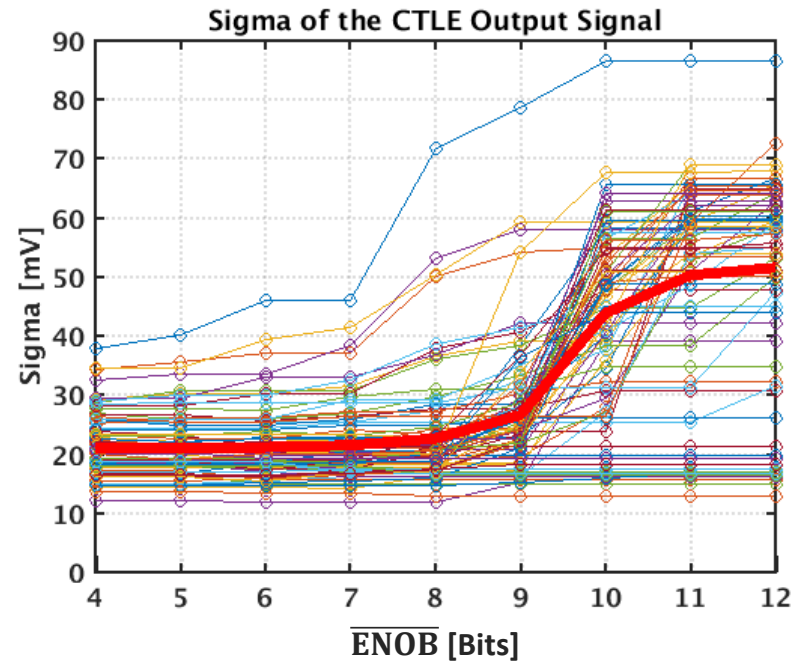
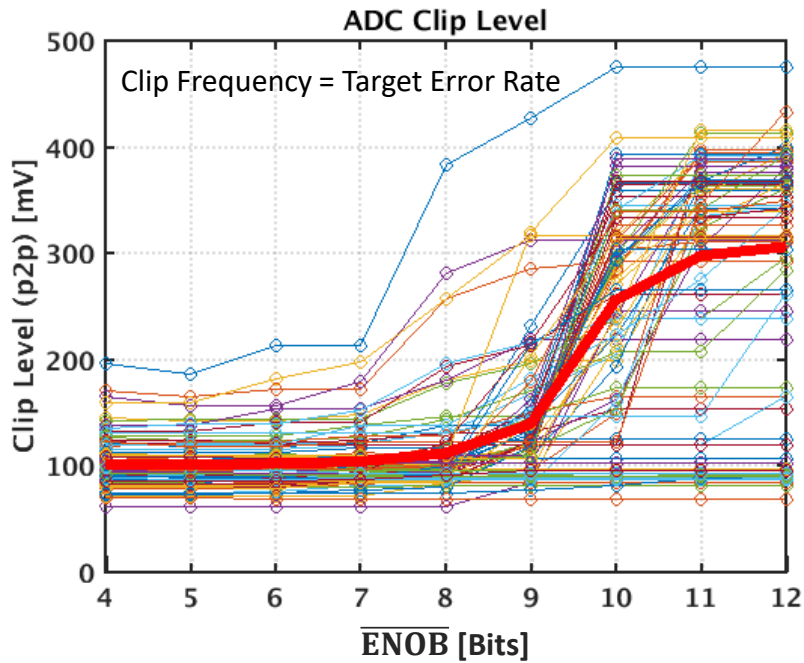


# Test Results – CTLE Utilization



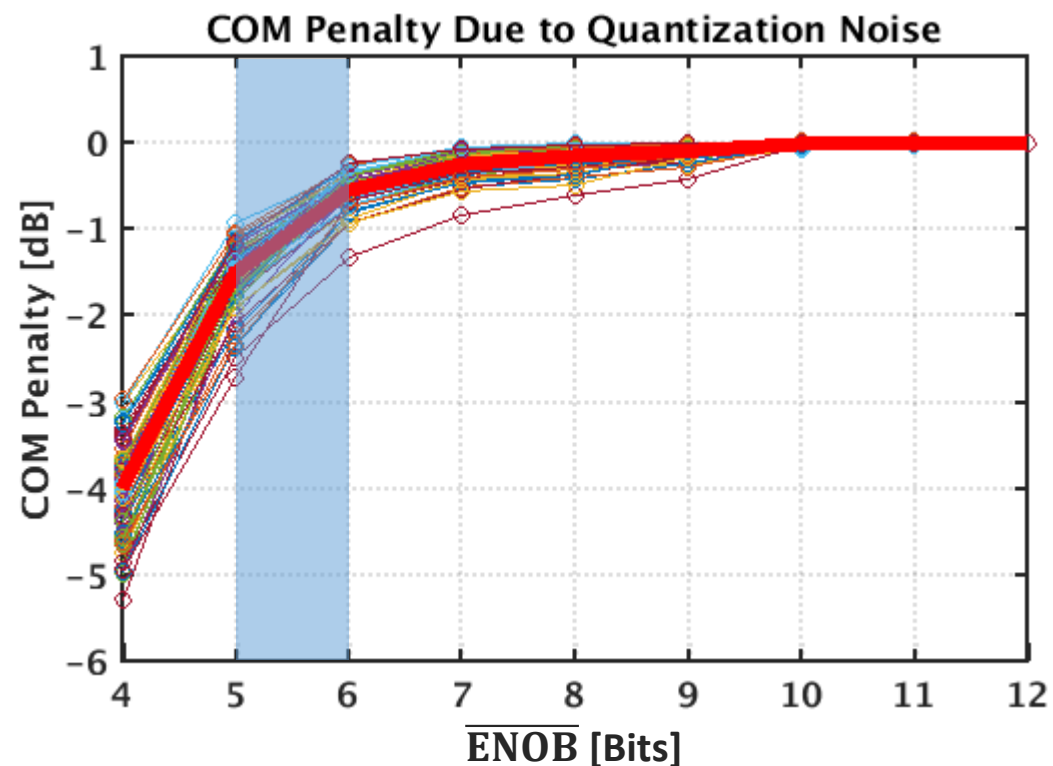
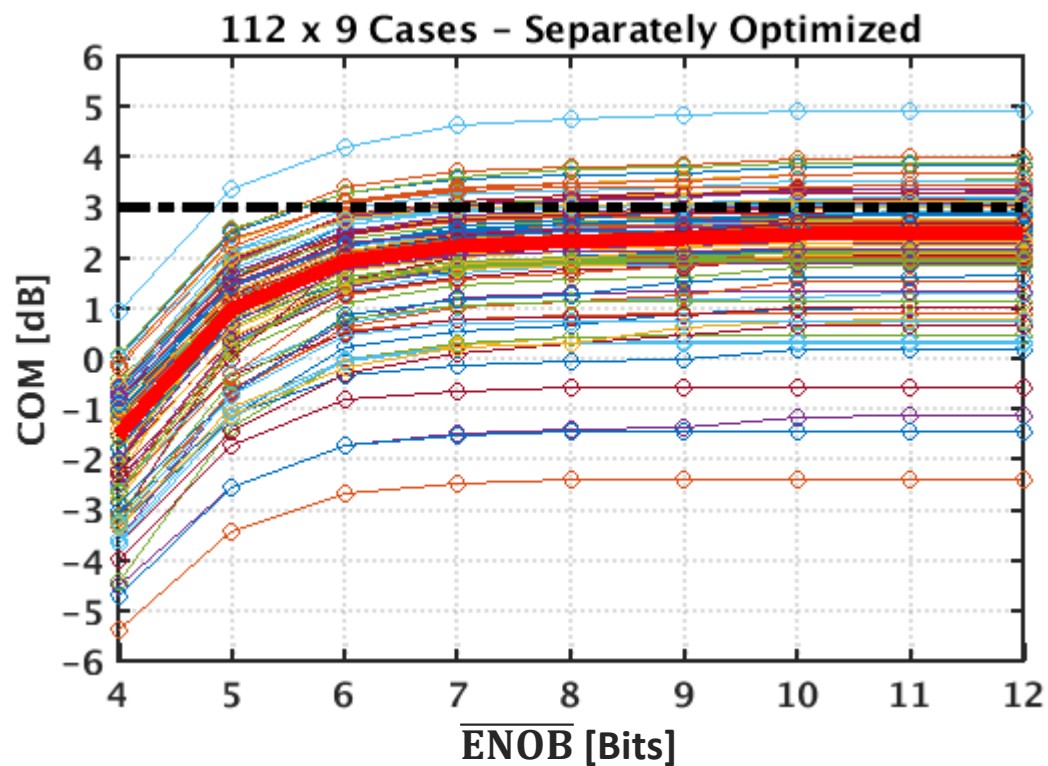
- As expected, CTLE under-utilization is entirely at high frequencies (from 15.1dB to 4.1dB)
- CTLE continues to equalize low frequencies

# Test Results – CTLE Output Signal Statistics



- CTLE marginalization increases p2p and sigma of the (under-equalized) signal as well as their ratio at the CTLE output (ADC input)
- To mimic this in practice, ADCs with large dynamic range and  $\overline{\text{ENOB}}$  are required (beyond what is readily available in today's technologies)

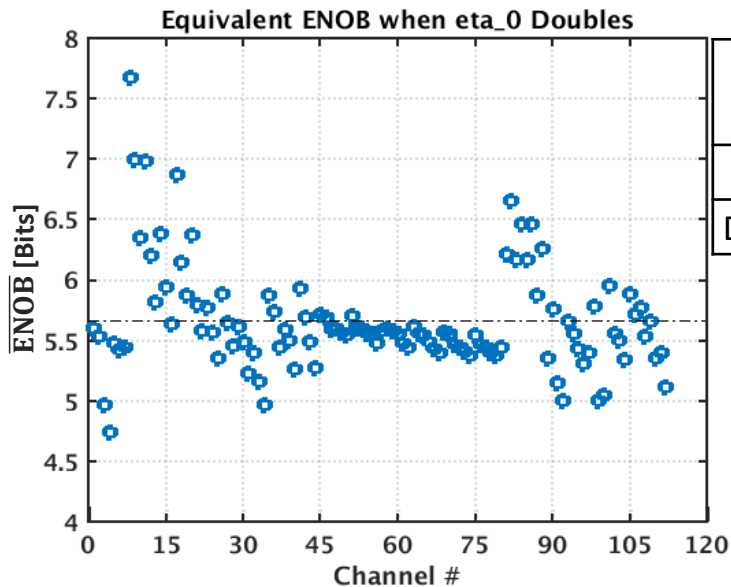
# Test Results – COM



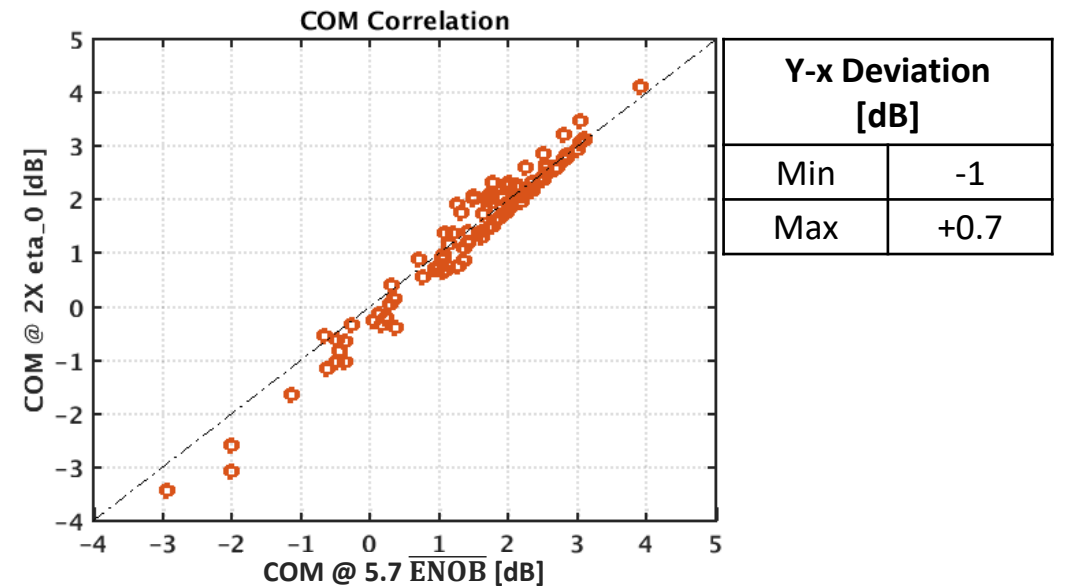
- Expect a considerable (unattended) drop in COM unless much better ADCs become available
- For example, for  $\overline{\text{ENOB}}$  of 5 to 6 bits, average COM drops by 1.5dB to 0.6dB
- Even if  $\overline{\text{ENOB}}$  improves by 1-2 bits, there is still a COM penalty that cannot be easily ignored

# Can I Tweak $\eta_0$ to Mimic Quantization Noise Effect?

- Remove quantization noise and scale up  $\eta_0$  (e.g. X2)
- For each test case, calculate COM and interpolate the previous plot (COM vs.  $\overline{\text{ENOB}}$ ) to obtain equivalent  $\overline{\text{ENOB}}$  that would have resulted in the same COM (left plot)
- Set  $\overline{\text{ENOB}}$  to the average equivalent  $\overline{\text{ENOB}}$  and calculate COM for each test case (right plot)
- For the test channels, if X2  $\eta_0$  is used to represent an  $\overline{\text{ENOB}}$  of 5.7 bits, there will be  $\sim 1\text{dB}$  COM inaccuracy (Likely not acceptable)



Can doubling  $\eta_0$  represent 5.7 bit  $\overline{\text{ENOB}}$ ?



# Summary and Conclusion

- Quantization noise is currently not represented in COM
- Quantization noise decreases SNR, influences noise coloring, and could have a major effect on equalization optimization and distribution
- RxFFE tap optimization MMSE method and MLSE  $\Delta$ COM are affected by noise coloring
- A set of 112 KR/CR channels were used to generate data
  - ❖ An  $\overline{\text{ENOB}}$  of 5/6 bits drops COM by as much as 2.7dB/1.3dB and on average by 1.5dB/0.6dB
  - ❖ This is currently ignored during COM compliance evaluations
  - ❖ If eta\_0 scaling were to be used to represent quantization noise, scaling by 2X corresponds to 5.7bit  $\overline{\text{ENOB}}$ , but results in as much as 1dB of COM inaccuracy
- Considering adding quantization noise as another noise component in the COM flow is suggested

# Backup – COM Config

Table 93A-1 parameters			
Parameter	Setting	Units	Information
f_b	106.25	GBd	
f_min	0.05	GHz	
Delta_f	0.01	GHz	
C_d	[0.4e-4 0.7e-4 1.1e-4 0.4e-4 0.9e-4 1.1e-4]	nF	[TX RX]
L_s	[0.13 0.15 0.14 0.13 0.15 0.14]	nH	[TX RX]
C_b	[0.3e-4 0.3e-4]	nF	[TX RX]
R_0	5.00E+01	Ohm	
R_d	[50 50]	Ohm	[TX RX]
PKG_NAME	PKG_HI_R_CLASSB		TX RX
A_v	0.413	V	
A_fe	0.413	V	
A_pe	0.608	V	
z_p select	[3]		
L	4		
M	32		
Filter and Eq			
f_r	0.58	'fb	
c(0)	0.55		[min:step:max]
c(-1)	0		[min:step:max]
c(-2)	0		[min:step:max]
c(-3)	0		[min:step:max]
c(-4)	0		[min:step:max]
c(1)	0		[min:step:max]
N_b	1	UI	
b_max(1)	0.75		As/dfe1
b_max(2..N_b)	0.3		As/dfe2..N_b
b_min(1)	0		As/dfe1
b_min(2..N_b)	-0.15	s	As/dfe2..N_b
g_DC	[-20:1.0]	dB	[min:step:max]
f_z	25.16	GHz	
f_p1	40.00	GHz	
f_p2	56.00	GHz	
g_DC_HP	[-6:1.0]		[min:step:max]
f_HP_PZ	1.328125	GHz	
Butterworth	1	logical	include in fr

I/O control		
DIAGNOSTICS	0	logical
DISPLAY_WINDOW	0	logical
CSV_REPORT	0	logical
RESULT_DIR	.\results\CACR_set1_(date)\	
SAVE_FIGURES	0	logical
Port Order	[1 3 2 4]	
RUNTAG	KR_set1_eal	
COM_CONTRIBUTION	1	logical
TDR and ERL options		
TDR	1	logical
ERL	1	logical
ERL_ONLY	0	ns
TR_TDR	0.01	
N	4000	logical
TDR Butterworth	1	
beta_x	0	
rho_x	0.618	
TDR_W_TXPKG	0	UI
N_bx	20	
fixture delay time	[0.0]	
Tukey Window	1	
Noise, jitter		
sigma_RJ	0.01	UI
A_DD	0.02	V <sup>2</sup> /GHz
eta_0	8.00E+09	dB
SNR_TX	33	
R_LM	0.95	
BREAD_CRUMBS	1	logical
DRM_CODE	1000000	2's complement
seed	40	2's complement

Table 93A-3 parameters			
Parameter	Setting	Units	Information
package_tl_gamma0_a1_a2	e-4 0.00065 0.0003		
package_tl_tau	0.006141	ns/mm	
package_Z_c	70.70; 80.80; 10	Ohm	
z_p (TX)	*1 1 1, 1 1 1, 0	mm	[test cases to run]
z_p (NEXT)	*1 1 1, 1 1 1, 0	mm	[test cases]
z_p (FEXT)	*1 1 1, 1 1 1, 0	mm	[test cases]
z_p (RX)	*1 1 1, 1 1 1, 0	mm	[test cases]
C_p	[0.4e-4 0.4e-4]	nF	[test cases]
Operational			
ERL Pass threshold	10	dB	
COM Pass threshold	3	db	
DER_0	1.00E+04		
T_r	0.00400	ns	
FORCE_TR	1	logical	
PMD_type	C2C		
EW	1		
MLSE	3	logical	
ts_anchor	1		
sample adjustment	[-8 8]		
Local Search	2		
Filter: Rx FFE			
ffe_pre_tap_len	6	UI	
ffe_post_tap_len	24	UI	
ffe_pre_tap1_max	1		
ffe_post_tap1_max	1		
ffe_tapn_max	1		
FFE_OPT_METHOD	MMSE		FV_LMS or MMSE
num_ui_RXFF_noise	512		
Floating Tap Control			
N_bg	0	0 1 2 or 3 groups	
N_bf	4	taps per group	
N_f	80	UI span for floating taps	
b_max	0.2	max DFE value for floating taps	
B_float_RSS_MAX	0.1	rss tail tap limit	
N_tail_start	25	(UI) start of tail taps limit	

SAVE_CONFIG2MAT		
Parameter	Setting	Information
RX_CALIBRATION	0	logical
Sigma BBN step	5.00E+03	V
ICN parameters		
f_v	0.278	Fb
f_f	0.278	Fb
f_n	0.278	Fb
f_2	61.625	GHz
A_ft	0.450	V
A_rt	0.450	V
board_tl_gamma0_a1_a2 [0.644084e-4 3.6036e-05] 1.4 db/in @ 53.125G		
board_tl_tau	5.790E-03	ns/mm
board_Z_c	100	Ohm
z_bp (TX)	32	mm
z_bp (NEXT)	32	mm
z_bp (FEXT)	32	mm
z_bp (RX)	32	mm
C_0	[0.2e-4 0]	nF
C_1	[0.2e-4 0]	nF
Include PCB	0	logical
Selections (rectangle, gaussian, dual rayleigh, triangle)		
Histogram_Window_Weight	gaussian	selection
Or	0.02	UI