### 802.3dj D1.0 Comment Resolution - Electrical track

Adee Ran, Cisco

### Introduction

- This slide package was assembled by the 802.3dj electrical editorial team to provide background and detailed resolutions to aid in comment resolution.
- Acknowledgement to Howard Heck, Adam Healey, Chris Diminico, Mike Dudek, Matt Brown, and Kent Lusted for reviewing and contributing to this work.
- Text in red indicates editors' explanation of the comments. Italics indicate an editor's observation which might be subjective.
- The responses to the comments may point to the slides for reference but the comment report is the official record.

### **Cross-Clause topics**

### Bessel-Thomson measurement filter bandwidth [15 comments, 178/179/176D/176E]

# 230

Done

C/ 178 SC 178.9.2 P275 L48 Mellitz, Richard Samtec Comment Type Comment Status X The Bessel-Thomson filter should track fr. Between 0.5 fb and 0.6 fb have been shown in presenations SuggestedRemedy change TBD to 67GHz

P275

L48

Proposed Response Response Status O

SC 178.9.2

Li. Mike Intel

Comment Type Comment Status X 3dB BW is TBD

SuggestedRemedy Change it to 65 GHz.

C/ 178

Rational, considering the common and cost effective 1.85mm connector BW, and associated ~7% measurement error, give rise to this number.

Proposed Response Response Status O

# 225 C/ 179 SC 179.9.4 P309 L23 Google

Comment Type Comment Status X

Adopted baseline https://www.ieee802.org/3/dj/public/24\_01/ran\_3dj\_01a\_2401.pdf has BT filter bandwidth as TBD but D1.0 has 40GHz. 3dB bandwidth of 40GHz is insufficient for 200Gbps/lane PAM4

SugaestedRemedy

Noujeim, Leesa

Increase to 65GHz, consistent with test equipment capabilities and demonstrated channel rolloff eg in https://www.ieee802.org/3/dj/public/23\_11/weaver\_3dj\_01\_2311.pdf and https://www.ieee802.org/3/dj/public/24\_01/benartsi\_3dj\_01\_2401.pdf OR change to TBD

Proposed Response Response Status O Comment #60 is against 178.9.2 (Transmitter characteristics) and suggests 67 GHz. Comment #32 proposes a similar bandwidth for receiver test calibration (178.9.3.3). The reasoning for both is that "The Bessel-Thomson filter should track fr".

Comments #131 and #133 against 176E suggest 58.4375 GHz with similar reasoning based on different assumed f r.

Note that there has never been a requirement to have the bandwidth of the measurement B-T filter equal to the COM f r parameter (part of the Rx model). (4 comments)

Comment #230 and comment #245 suggest 65 GHz with reasoning based on connector availability.

Note that the B-T filter is implemented in the test equipment and the cost of other setup components is likely less significant.

(2 comments)

Comment #225 highlights an error in D1.0 - the value 40 GHz appears once although it has not been part of the baseline (and thus has not been adopted).

The proposed value here and in #217 is 65 GHz based on "test equipment capabilities and demonstrated channel rolloff".

Comments #124 and #388 also address the same issue but suggest changing to TBD. (4 comments)

Comments #399, #410, #412, #422, #425 suggest the value **62 GHz**. (5 comments)

Summary: The suggested values are 58,4375, 62, 65, and 67 GHz.

Editorial team proposal: use a BW of X GHz for signal measurements in 178, 179, 176D, 176E. Replace all TBDs and the "40 GHz" that wasn't adopted. Proposed value for X is 65.

### ERL (14 comments, 178/179/179B)

Done

C/ 178 SC 178.9.2.2 P278 L26 Mellitz, Richard Samtec Comment Type Comment Status X scale ERL parameter form 0.3ck SuggestedRemedy in table 163-7 change TBD's as follows Tr 0.005 ns Bx 0 GHz px 0.618 N 400 UI Proposed Response Response Status O

#### There are several similar comments:

- #29 against 178.9.2.2 (KR Tx/Rx ERL) with N=400 (x2 of 802.3ck)
- #28 against 178.9.2.1.2 (KR test fixture ERL) with N=400 (x2 of 802.3ck)
- #43 against 178.10.3 (KR channel ERL) with N=7000 (x2 of 802.3ck)
- #48 against 179.9.4.8 (CR Tx/Rx ERL) with N=1600 (x2 of 802.3ck)
- #51 against 179.11.3 (CR cable assembly ERL) with N=4500 (same as 802.3ck)
- #58 against 179B.4.2 (Mated test fixtures ERL) with N=1600 (x4 of 802.3ck), tw=1, DER0=2e-5

All propose values for ERL parameters, dividing T\_r by 2 and multiplying N by as noted above, relative to the values in 802.3ck (100G).

Additional comments #237, #238, #239, #240 (Tr, beta\_x, rho\_x, N for KR Tx/Rx ERL, respectively) suggest some of the ERL parameters with the same values as above:

#### Other comments

- #241 proposes 44 for N\_bx for KR Tx/Rx ERL
- #231 and #244 propose -3 dB for min dERL (KR Tx/Rx)
- #252 proposes 11 dB for min ERL (KR channel)

The justification provided for ERL parameters is very basic but may be sufficient. (are the values of N in consensus?)

No comments address ERL parameters in Annex 176D (C2C component and channel) and 176E (C2M host and module) which are mostly TBD.

### Editorial team proposal:

- Accept the suggested remedies of comments 28, 29, 43, 48, 51, 58, 241, 244, 252
- Refer 237, 238, 239, 240 to #29
- Use the suggested values from clause 178 and 179 to fill in TBDs for annex 176D and annex 176E respectively, except for the value of N and minimum ERL/dERL which will stay TBD.

### **ERL Tfx (5 comments, 179/176E)**

Done

CI 179 SC 179.9.4.8 P315 L35 Noujeim, Leesa Google Comment Type Comment Status X Practical test fixtures may have discontinuities close to 0.2ns from the host-facing connection (mating interface). If the intent is to remove the test fixture discontinuities from the ERL calculations, we should adjust the 0.2ns SuggestedRemedy Change text to "...Tfx equal to twice the delay between the test fixture connector and the test fixture host -facing connection minus 0.2ns or as needed to remove test-fixture discontinuities from the ERL result" Proposed Response Response Status O

The comment suggests that the value 0.2 ns may not be correct.

#### There are several similar comments:

- #227 against 179.9.4.8 (CR host transmitter ERL)
- #218 against 179.11.3 (CR cable assembly ERL)
- #219 against 179.9.5.5 (CR host receiver ERL)
- #220 against 176E.3.3.3 (C2M Host output ERL)
- #221 against 176E.3.4.2 (C2M Module output ERL)

### Effect of the suggested change:

### 179.9.4.8 Transmitter effective return loss (ERL)

The ERL of the transmitter at TP2 is defined by the procedure in 93A.5 using the values in Table 179–9 and Table 179–15, and with the value of  $T_{fx}$  equal to twice the delay between the test fixture connector and the test fixture host-facing connection minus 0.2 ns or as needed to remove test-fixture discontinuities from the ERL result"

For module:
"module-facing"
For cable assembly:
"cable facing"

Note that the 0.2 in the existing text is a reduction from "twice the delay between the test fixture connector and the test fixture host-facing connection".

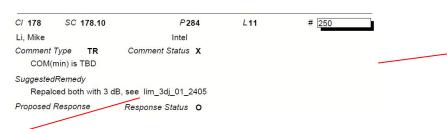
This means the measurement is time gated to remove reflections from the whole text fixture except for a length equivalent 0.1 ns (assuming "twice" has precedence over "minus"). 0.1 ns is approximately 17 mm, so for hosts, the time-domain response starts at 17 mm before the HCB's card end, and likely includes the pad and via. Using different values may cause the via to be excluded, which may affect the ERL result.

For modules and cable assemblies, the MCB has a receptacle, which may require different time gating, but the proposal would leave Tfx open for interpretation and creates ambiguity.

Editorial team proposal: REJECT #218, #220, #227, #219, #221.

- For host ERL, the existing specification of Tfx is appropriate.
- For module and cable assembly ERL, the proposal is not specific enough.

## COM (minimum/target value) [13 comments, 178/179/179A/176D]



Done

This comment and multiple others suggest using 3 dB as the minimum COM for channels

- 178.10 channel: #33, #250, #402, #253
- 179.11 cable assembly: #50, #413
- 176D.4 channel: #430
- 179A.7 TP0d-TP5d: #57

3 dB is also suggested for Rx test channel calibration:

178.9.3.3: #249, #400 179.9.5.3: #49, #411 176D.3.4.4: #427

The referenced presentation, <a href="https://www.ieee802.org/3/dj/public/24\_05/lim\_3dj\_01\_2405.pdf">https://www.ieee802.org/3/dj/public/24\_05/lim\_3dj\_01\_2405.pdf</a>, does not include a rationale for the proposed value of COM

The minimum COM should be based on expected implementation penalty with respect to the new reference receiver, for which many parameters are TBD. Specifically, the presentations <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/shakiba 3dj 02 2405.pdf are relevant in this context.

However, the same value is proposed in several comments, so it may be in consensus and enable moving forward.

Editorial team proposal: Use the value **X** for minimum COM for channels and for test setup calibration in KR/CR/C2C. Proposed value for X is 3 dB.

## Reference impedance for specifications [2 comments, 178/179]

In agenda

CI 178 SC 178.9.1 P275 # 395 L39 Kocsis, Sam Amphenol Comment Type Comment Status X The reference impedance should match the system impedance, Rd as defined in COM spreadsheets. SuggestedRemedy 92-ohm, TBD, or straw poll based on proposed values presented in Task Force contributions Proposed Response Response Status O

2 comments (#395, #387) about the reference impedance for specifications suggest that it should match the system impedance, and that 46 Ohm (92 Ohm differential) is a possible value.

The reference impedance for specifications (not R\_0) affects ERL and other RL results (S-parameter values depend on the reference impedance), and if test equipment is matched to this impedance it will also affect time-domain measurements (R peak, SNR ISI).

Designs may take this into consideration, and it would be preferable if these specs do not penalize components matched to the "system impedance" (assumed impedance of the other system components). ⇒ see next slide.

### 178.9 Electrical characteristics

#### 178.9.1 Reference impedance

The reference impedance for differential specifications is  $100 \Omega$ . The reference impedance for common-mode specifications is  $25 \Omega$ .

## **COM R\_d parameter** [8 comments, 178/179/176D/176E]

In agenda

Cl 178 SC 178.10.1 P285 L40 # 396

Kocsis, Sam Amphenol

Comment Type T Comment Status X

Rd(t) = "TBD"

SuggestedRemedy

Change "TBD" to "92-ohm" to match majority of contributions to the Task Force, and better align with Zc definition in package

Proposed Response Response Status O

This comment and 3 similar ones (#396, #397, #391, #392) are about the R\_d parameter, which is used as the terminations (Tx and Rx) in the COM pulse response calculation.

The value in the suggested remedy seems to be differential, and would become 46 Ohm single ended.

Other comments suggest different values:

- #255, #256: 46.25 Ohm
- #141. #137: 50 Ohm

The value of R\_d affects the reference ERL (and thus dERL) and the reflections within the reference package (and thus COM and dR\_peak).

ERL is also affected by the **reference impedance for measurements** (which is now, effectively, 50 Ohm single-ended - see previous slide) so these specifications are closely tied with each other.

Note that if  $R_d$  is different from the termination used in time-domain measurements (e.g. scope) then  $A_v$ ,  $A_f$  e and  $A_f$  ne need to change to create the minimum compliant  $v_f$  when measured on a scope. Additionally, Tx output specifications may need to be changed.

### Editorial team proposal:

- Change Rd (both t and r) from TBD to **X Ohm** in COM device parameters tables (Table 178–12, Table 179–15, Table 176D).
- Change the reference impedance statements (178A.1.3, 178.9.1, 179.9.3, 179.11.1, and 176D.3.2) to define a reference **single-ended** impedance of **X Ohm** for all specifications, e.g., insertion loss, return loss, and ERL. Add a similar statement in 176E.

Proposed value for X: 46 Ohm.

If X is different from 50 Ohm, add a NOTE that s-parameter measurements may be made with a different impedance and converted mathematically to that reference. Add an editor's note that output waveform measurements on 50 Ohm scopes need to be addressed.

If the reference impedance is different from R\_d, add an editor's note after each COM parameter table, noting that the values of A\_v, A\_ne and A\_fe need confirmation.

## COM R\_0 parameter [9 comments, 178/179/176D/176E]

In agenda

Cl 178 SC 178.10.1

L38

# 35

Mellitz, Richard

Samtec

Comment Type TR

Comment Status X

(Table 178–12): Computation can be independent of R0. Add a note to explain. S parameter can utilize any R0. For computation purposes s-parameters are converted to 50 ohms which is the native impedance for the most common test equipment.

P285

#### SuggestedRemedy

Change R0 for TBD to 50 ohms and add a note indicating the imported s-parameter are to be converted into 50 ohm reference before computation.

Proposed Response

Response Status O

### Editorial team proposal:

Align the value of R\_0 with the reference impedance for specifications in all COM tables. Use the table as a specification and refer to it from the subclauses that define the reference impedance for specifications (178.9.1, 179.9.3, 179.11.1, 176D.3.2).

Delete the statement about reference impedance for measurements of S-parameters in 178A.1.3.

Add an informative NOTE as in the suggested remedy in those subclauses.

2 comments (#35, #52) suggests that R\_0 should not affect COM result.

7 comments (#254, #403, #414, #141, #431, #136, #438) suggest the value 50 Ohm for R\_0, as used in previous projects.

Note that the reference impedance for specifications is stated explicitly in 178.9.1, 179.9.3, 179.11.1, and 176D.3.2 (see previous slide), and does not refer to R 0.

178A.1.3 already defines the reference impedance for s-parameters (effectively 50 Ohm single ended). We should consider defining the value fo R\_0 and the reference impedance for s-parameters once (as one parameter).

#### 178A.1.3 Measurement of the channel under test

The S-parameters for each signal path are measured between the test points specified by the clause or annex that utilizes this calculation. It is recommended that the scattering parameters be measured with a uniform frequency step from a start frequency no greater than 10 MHz (TBC) to a stop frequency of at least TBD GHz. The measurement frequency step corresponds to the time span of the pulse response derived from the S-parameters (see 178A.1.6). The frequency step should be chosen to be small enough so that all significant components of the pulse response are included.

The reference impedance for the measurement of differential-mode S-parameters is 100  $\Omega$ .

### COM Tx FFE [10 comments, 178/179/176D/176E]

In agenda

Presentations so for have not shown the need for Tx FFE. Change to no TXFFE until further data is provided.

Rx noise may suggest a need for the TXFFE which would improve performance. It's not clear from a channel perspective that the TX FFE is not a zero sum gain compared to the Rx noise loss of COM. Until Rx FFE noise is better defined zero out TXFFE

SuggestedRemedy

Change TBDs for c(-3),c(-2),c(-1), and c(1) to zero. Set C(0) tp 1.

Proposed Response Response Status O

Coefficient	#37 (C178)	#142, #138 (A176D, A176E)	#258-#262 (C178)	#405, #416 (C178, C179)
c(-3)	0	-	-	-0.06 : 0.02 : 0
c(-2)	0	0:0.02:0.14	0:0.02:0.16	0:0.02:0.12
c(-1)	0	-0.3 : 0.02 : 0	-0.4 : 0.02 : 0	-0.34 : 0.02 : 0
c(0)	1	0.65	0.54	0.5
c(1)	0	-0.14 : 0.02 : 0.14	-0.2 : 0.02 : 0	-0.2 : 0.02 : 0

The comment suggests that COM should use no Tx equalization.

The contributions so far are not sufficient evidence for not needing a Tx FFE as suggested in comment #37. The chosen Tx FFE values ("no equalization") in analysis presented so far is due to the unlimited Rx FFE coefficients used in the analysis, but these limits are still TBD. The possible inclusion of an ADC model (with dynamic-range-dependent quantization noise), presented in https://www.ieee802.org/3/dj/public/24\_05/healey\_3dj\_01b\_2405.pdf, may change the chosen Tx FFE.

There are several proposals in different sets of comments on the same topic However, it may be possible to reach consensus and enable moving forward.

### Editorial team proposal:

- Adopt the highlighted values in the tables above for the COM tables in 178, 179, 176D, and 176E
- no c(-3)

## **COM f\_r parameter**[9 comments, 178/179/176D/176E]

In agenda

Cl 178 SC 178.10.1 P286 L12 # 36

Mellitz, Richard Samtec

Comment Type TR Comment Status X

T (able 178–13) Presentations so far have used fr of 0.5, 0.55, 0.58, and 0.6. 67 Ghz limits on test equipment and cabling/connector modal physics suggest at least a 9 dB loss is required for good measurements at 67 GHz. Set fr to 0.6 or lower to achieve this.

SuggestedRemedy change TBD to 0.6.

Proposed Response Response Status O

Comment #36 suggests that the values 0.5, 0.55. 0.58 and 0.6 are candidates, and suggests 0.6. The suggested relationship between f r and measurement bandwidth has not been established.

Several values are suggested in various comments:

- #36, #53: 0.6
- #404, #415, #432, #439: 0.58 (no rationale)
- #137, #141; 0.55 (no rationale)
- #257: 0.5 (refers to lim\_3dj\_01\_2405, but no rationale in this presentation)

Previous reference receivers used the value 0.75, but several past contributions suggest that a lower value would both be more realistic and yield better COM results.

No comment suggests that there should be a difference between interfaces.

With not-too-different values proposed in several comments, it may be possible to reach consensus and enable moving forward.

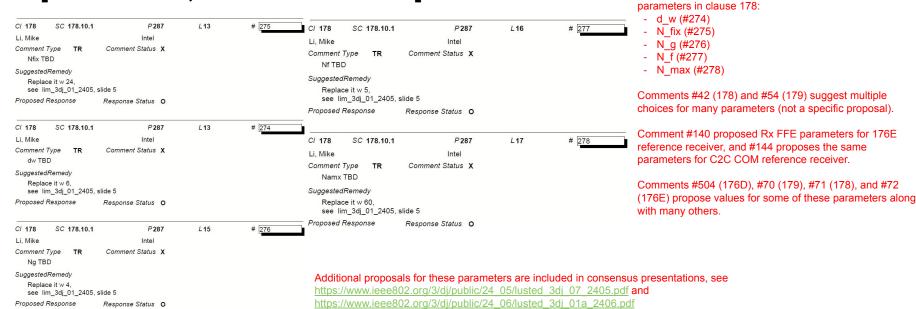
Editorial team proposal:

Use  $\underline{f}_r = X^*f_b$  as reference receiver bandwidth for all electrical interfaces (Table 178-13, Table 179-16, Table 176D-6, and Table 176E–7). Proposed value of X: 0.55

# COM Rx FFE length parameters [9 comments, 178/179/176D/176E]



5 separate comments against COM Rx FFE length



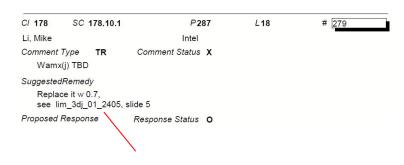
### Editorial team proposal:

Use d\_w=5 (as proposed in <u>lusted\_3dj\_07\_2405</u>), N\_fix=14, N\_g=2, N\_f=4, and N\_max=50 (as proposed in <u>lusted\_3dj\_01a\_2406</u>) to replace TBDs in Table 176D–7 and Table 176E–7.

Add editor's notes similar to that in slide 4 of <u>lusted 3dj 07 2405</u> and slide 5 of <u>lusted 3dj 01a 2406</u> to denote that these values need further analysis. Keep these values TBD in clauses 178 and 179.

## COM Rx FFE and DFE coefficient limits [14 comments, 178/179/176D/176E]

In agenda



The referenced presentation, <u>lim 3dj 01 2405</u>, does not include an explanation of the proposed value.

4 separate comments against COM Rx FFE coefficient limits in clause 178:

- Wmax: 0.7 (#279) and Wmin: -0.7 (#280)
- bmax: 0.85 (#281) and bmin: 0.3 (#282)
- Floating tap min/max (#283 and #284)

For bmax, comment #140 (176E) and #144 (176D) propose 0.75 (but also d\_w)

Comments #42 (178) and #54 (179) suggest multiple choices for many parameters (not a specific proposal).

Comments #504 (176D), #70 (179), #71 (178), and #72 (176E) propose values for some of these parameters along with many others.

The presentation <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/heck 3dj 01b 2405.pdf was provided in support of #504. Slides 25-26 include proposed values for Wmin/Wmax (per index) and Bmax=0.75

A wide range allowed for FFE coefficients, combined with not accounting for quantization noise, effectively assumes a high-resolution ADC and causes the Tx FFE coefficients not to be utilized in COM optimization. This implies that receivers do not need Tx equalization. The penalties of real receivers having more limited FFE equalization, which would some Tx equalization, have not been analyzed. Further analysis in this area is encouraged.

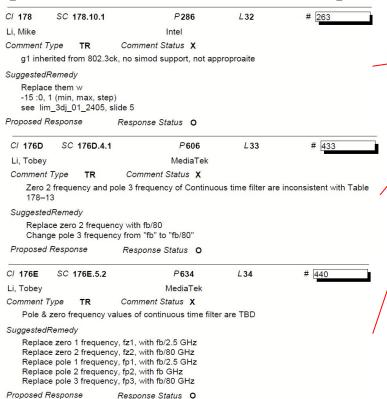
However, if consensus can be achieved, having numbers instead of TBDs is a step forward.

#### Editorial team proposal:

Use the proposed values of w\_max, w\_min, b\_max, b\_min in comments #279-#282 to replace TBDs in COM tables in 178, 179, 176D, and COM table + reference receiver in 176F.

Add editor's notes similar to that in slide 4 of https://www.ieee802.org/3/dj/public/24 05/lusted 3dj 07 2405.pdf to denote that these values need further analysis.

# COM CTLE parameters [6 comments, 178/176D/176E]



Done

4 separate comments against COM CTLE parameters in Table 178-13 (non-TBD):

- g1: -15:1:0 (#263) (-15 instead of -20)
- g2: -5:1:0 (#264) (-5 instead of -6)
- fz1, fz2: fb/4.233, fb/80 (#265) (fb/4.233 instead of fb/2.5)
- fp1, fp2, fp3: fb/1.8973, fb/2.6562, fb/80 (#266) (fb/1.8973 instead of fb/2.5 and fb/2.6562 instead of fb)

The referenced presentation <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/lim 3dj 01 2405.pdf does not include a rationale for the proposed changes (highlighted) to the existing values of these COM parameters, nor clarifies why they should be made different from those of clause 179.

Comment #433 suggests changes in Table 176D-6 (C2C) to align it with Table 178-13 (KR)

- Suggests fz2 = fb/80 (replacing TBD)
- Suggests fp3 = fb/80

Comment #440 suggests CTLE parameters for the C2M reference receiver in Table 176E–7 (all currently TBD) that are aligned with the COM parameters of clause 178 and 179.

In addition, the editorial team noticed that Table 176D-7 includes the parameter fLF which is not defined in Annex 176A (it is replaced by fp3 and fz2).

### Editorial team proposal:

- Use the identical CTLE parameters from Table 178-13 and Table 179-16, without change, in C2C (Table 176D-6) and C2M (Table 176E-7)
- Align fp1, fz1, fp2, fz2, fp3 in all tables
- Remove fLF from Table 176D-7

(AIP #433, AIP #440, REJECT #263-#266)

## COM T\_r parameter [6 comments, 178/179/176D/176E]

In agenda

Cl 179 SC 179.10.1 P286 L50 # 39

Mellitz, Richard Samtec

Comment Type TR Comment Status X

scale Tr from .3ck. Understand that this is not the Tr at TP0d.

SuggestedRemedy
set Tr to 0.00375 ns

Response Status O

This comment suggests that the transition time T\_r used in COM should be scaled down from the value 7.5 ps used in 802.3ck, resulting in 3.75 ps; and states that "this is not the Tr at TP0d".

Other comments (#268, #407, #418, #435, #441) suggest the value 4 ps instead.

Note that T\_r (modeled prior to the device termination) contributes to the transition time at TP0d (after the device termination), and will therefore affect Tx specifications at TP0v (specifically, the reference R\_peak). The sensitivity of these specifications to the suggested values of T\_r (combined with the device and package model) has not been addressed by any contributions. Further analysis in this area is encouraged.

However, having numbers instead of TBDs is a step forward, if consensus can be achieved.

### Editorial team proposal:

Proposed Response

- Change T\_r from TBD to **X** ps in Table 178–13, Table 179–15, Table 176D–7, and Table 176E–7.
- Add editor's notes similar to that in slide 4 of <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/lusted 3dj 07 2405.pdf to denote that this value needs further analysis.

Proposed value of X: 4

## SNDR/SCMR/SNR\_TX [6 comments, 178/179]

In agenda

CI 179 SC 179.9.4.1.1 P312 L42 Mellitz Richard Samtec Comment Type Comment Status X SNDR reduces with loss and used that way for equation 178A-18. SuggestedRemedy Insert a subsection e) Loss correction factor for fitted pulse measurements. See presentation Proposed Response Response Status O C/ 179 SC 179.9.4.6 P315 L17 Mellitz, Richard Samtec Comment Type Comment Status X SNDR reduces with loss and used that way for equation 178A-18. SuggestedRemedy The transmitter SNDR is defined by the measurement method described in 120D.3.1.6 The transmitter SNDR is defined by the measurement method described in 120D.3.1.6 plus a power loss factor defined in xxxx Proposed Response Response Status O

If the proposal is adopted, implementing this proposal would preferably done with broad editorial license.

Other comments (shown on subsequent slides) are based on this proposed change.

The following presentation was reviewed by the task force at the May interim meeting: https://www.ieee802.org/3/dj/public/24 05/mellitz 3dj 02 2405.pdf

The presentation suggested effectively changing the definition of the "signal" component of SNDR as shown in the excerpts below.

The motivation is that this way the SNDR measurement at different losses between the source and the measurement point yield consistent results.

The presentation suggests a specific way of writing this definition as a correction factor "So we don't change prior standards", but this can be done specifically for the clauses in this project without affecting other standards.

Comment #47 seems to suggests essentially the same change.

- ☐ For the "S" in SNDR use the power variance of the signal at the measurement point as follows which is the in time and frequency domain
  - $\sigma_p^2 = \sum_{1}^{M(N_p Dp 1)} p(n)^2$
  - Instead of p<sub>max</sub>
- ☐ Consider SNDR as a ratio of signal power variance to noise power variance
  - Perhaps: SNDR should be  $10*log_{10}\left(\frac{\sigma_P^2}{\sigma_e^2+\sigma_n^2}\right)$

## SNDR/SCMR/SNR\_TX [6 comments, 178/179]

Comment Type

SuggestedRemedy

Proposed Response

adjust SCMR with loss correction factor

add + loss correction factor to equation 178-1

In agenda

18

C/ 178	SC 178.9.2	P276	5 L34	# 27
Mellitz, Ric	hard	Samted	0	
Comment 7 adjust 9		Comment Status )	en and an enter the same and	on prior assumptions
Suggestedl change	Remedy SNDR to 33,5	dB.		
C/ 178	SC 178.10.2	P287	L5	# 41
Mellitz, Ric	hard	Samted		
Comment T SNR_T	Parent Marian	Comment Status ) when loss correction i		
Suggestedl Change	Remedy e TBD to 33.5 dB	3		
Proposed F	Response	Response Status (	)	
C/ 178	SC 178.9.2.6	P279	9 L <b>22</b>	# 31
Mellitz, Ric	hard	Samte	С	

Comment Status X

Response Status O

- #27 suggests setting the minimum SNDR in clause 178 to 33.5 dB based on the expected improvement from the new definition.
- #41 suggests using the same value for SNR\_TX.
- #31 suggests that the correction factor (essentially the new numerator) be applied to SCMR in clause 178.
- #270 suggests 33 dB for SNDR in clause 178 (not related to the proposed redefinition of SNDR).

### Editorial team proposal:

### For #45:

- Change the definitions of SNDR in 179.9.4.6 and SCMR in 178.9.2.6 to use a numerator based on the suggested equations in <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/mellitz 3dj 02 2405.pdf slides 12 and 13, with editorial license.
- Change SNDR (min) to **X** in transmitter characteristics in 178, 179, and 176D.
- Change SNR\_TX in COM tables to X.

Suggested value for X: 33.5 dB.

#47, #31, #27, #41, #270: resolve using the response to #45.

# **Jitter**[4 comments, 178/179]

CI 178 SC 178.9.2 P276 L38 Li. Mike Intel Comment Type TR Comment Status X Output litter (max) TBD SuggestedRemedy reapcle TBDs with: Jrms: 0.023 UI J2 7u03: 0 102 UI J2 7u: 0 110 UI Even-odd iitter, pk-pk; 0.025 UI See lim\_3dj\_01\_2403a, lim\_3dj\_01\_2405, and [1], [2], [3] Proposed Response Response Status O SC 179.9.4.7 Cl 179 P310 L25 # 204 Ran, Adee Cisco Comment Type TR Comment Status X Jitter specification is TBD.

#### Based on

https://www.ieee802.org/3/dj/public/adhoc/electrical/24\_0104/calvin\_3dj\_elec\_01a\_240104. pdf, the jitter measurement methodology of existing clauses 162, 163, and 120G (specifically using the two edges R03/F30) is feasible for measurements with a loss 30 dB. It is expected that the same method can be used for higher losses as long as the scope can maintain CDR lock.

This methodology should be used for all electrical interfaces, with adequate adjustments.

#### SuggestedRemedy

A detailed proposal will be provided.

Proposed Response Response

Response Status O

In agenda

- #236 refers to the following presentations:

https://www.ieee802.org/3/dj/public/24 03/lim 3dj 01a 2403.pdf, https://www.ieee802.org/3/dj/public/24 05/lim 3dj 01 2405.pdf

#204 refers to

https://www.ieee802.org/3/dj/public/adhoc/electrical/24 0104/calvin 3dj elec 01a 240104.p df and the detailed proposal mentioned is https://www.ieee802.org/3/dj/public/24 05/ran 3di 03 2405.pdf.

- #271 and #272 suggest A DD=0.02 UI and sigma RJ=0.01 UI.

Both #236 and #204 (in <u>ran 3di 03 2405</u>) suggest using jitter specifications based on the dual-Dirac model with parameters in #271 and #272. As a result, both proposals have approximately the same JRMS. EOJ is also similar.

The main difference is that comment #236 suggests using a new parameter J2.7u, while #204 (in ran 3di 03 2405) suggests using J3u and J6u, measured only on the R03/F30 edges.

Additional relevant presentation was reviewed in the May interim meeting: <a href="https://www.ieee802.org/3/di/public/24">https://www.ieee802.org/3/di/public/24</a> 05/zivny 3dj 01a 2405.pdf, suggesting changes to the measurement method to potentially improve the precision. The following straw poll was taken after the presentation:

#### Straw Poll #8

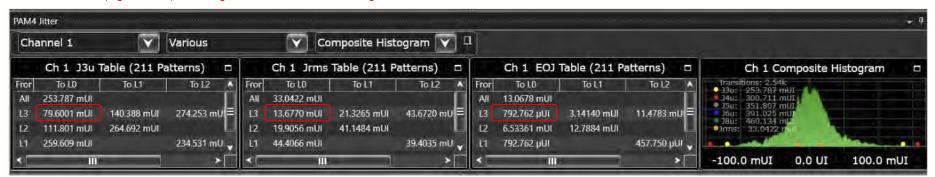
I would support the approach for the C2M and CR measurement specifications in zivny\_3dj\_01a\_2405 Y: 12, N: 11, NMI: 22, A: 36

The results show support for the proposed jitter measurement methodology changes, but consensus is not obvious

# **Jitter (cont.)**[4 comments, 178/179]

In agenda

Based on the results reported on slide 7 of <u>zivny 3dj 01a 2405</u>, it is technically feasible to make jitter measurements after a channel with 33 dB IL using the "large transitions" (e.g. L3 to L0). Including smaller transitions would degrade the results.



In previous projects we had slightly higher J3u/J3u03 limits at TP2 (CR) than in TP0a/TP0v (KR/C2C), to account for degradation due to the host channel loss.

Table 163-5

Jitter (max) <sup>c</sup>			
$J_{RMS}$	162.9.4.7	0.023	UI
J3u <sub>03</sub>	162.9.4.7	0.106	UI
J3u	162.9.4.7	0.115	UI
Even-odd jitter, pk-pk	162.9.4.7	0.025	UI

Table 162-11

Output jitter (max)			
J <sub>RMS</sub>	162.9.4.7	0.023	UI
J3u <sub>03</sub>	162.9.4.7	0.115	UI
J3u	162.9.4.7	0.125	UI
Even-odd jitter, pk-pk	162.9.4.7	0.025	UI

In this project we have **two** Tx package classes for TP0v and **three** host classes for TP2 with different loss to the measurement test point. Following the same reasoning, the jitter specs may be different in each case.

# **Jitter (cont.)**[4 comments, 178/179]

In agenda

A possible set of limits for measurements using only R03/F30 transitions is shown in the tables below. The highlighted values are the same as in 802.3ck KR/CR.

Interface	Max J <sub>RMS</sub>	Max EOJ	Max J3u	Max J6u
C2C and KR transmitter class A	0.023	0.025	0.106	0.138
(TP0v)				
C2C and KR transmitter class B	0.023	0.025	0.108	0.140
(TP0v)				
CR transmitter, host-low (TP2)	0.023	<mark>0.025</mark>	<mark>0.115</mark>	0.147
CR transmitter, host-nom (TP2)	0.023	0.025	0.122	0.154
CR transmitter, host-high (TP2)	0.023	0.025	0.128	0.160

Interface	Max J <sub>RMS</sub>	Max EOJ	Max J3u	Max J6u
C2M module output (TP4)	0.023	0.025	0.106	0.138
C2M host output (TP1a)	0.023	0.025	0.124	0.156

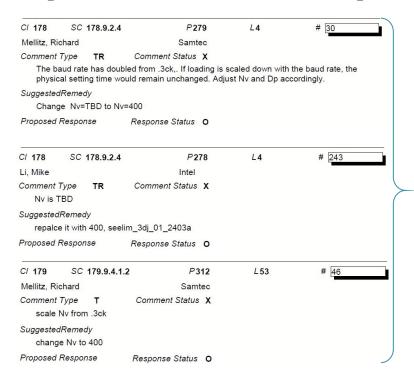
To unify and simplify the specifications, it is suggested that the combination of J3u and J6u is sufficient for AUI specifications; there is no need to specify J4u (or similar specs) for AUI-C2C, nor for AUI-C2M.

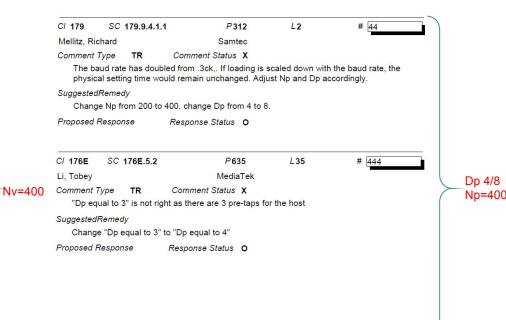
### Editorial team proposal:

- Adopt the COM parameter values A dd=0.02 and sigma RJ=0.01 for all COM tables.
- For clauses 178 and 179 and Annex 176D, adopt the limits in the left table, measured only on R03/F30 edges, for the following parameters:
  - JRMS
  - EOJ
  - J3u
  - J6u
- For Annex 176E (comments #186 and #187), adopt the limits in the right table.
- Add editor's notes that the jitter limits need confirmation.
- For measurement over all 12 edges, encourage additional contributions for feasible values.

# Pulse response linear fit [5 comments, 178/179/176E]

In agenda





## Pulse response linear fit (cont.) [5 comments, 178/179/176E]

### 179.9.4.1.2 Steady-state voltage and linear fit pulse peak ratio

The linear fit pulse peak  $v_{peak}$  and steady-state voltage  $v_f$  are defined using the linear fit pulse response p(1) through  $p(M \times N_v)$ , measured with transmit equalizer set to preset 1 (no equalization)  $N_v$  is set equal to 200. The linear fit procedure for obtaining p(k) and the values of M and  $N_p$  are defined in 179.9.4.1.1.

### 178.9.2.4 Difference steady-state voltage

The difference steady-state voltage of the transmitter at TP0v is computed using the procedure in 163A.3.2.1 with  $N_v = \overline{\text{IBD}}$  and other parameter values specified in Table 178–12. The reference value,  $v_f^{(ref)}$ , is calculated based on the receiver package class to which the device adheres.

The difference steady-state voltage at TP0v shall meet the specification  $dv_f$  (min) in Table 178–6.

#### 176E.5.2 Output signal measurement method

<...>

Perform the following steps for each valid combination of  $g_{DC}$  and  $g_{DC2}$  as specified in Table 176E–7:

- b) Compute the response  $y_2(k)$  by applying the effect of the continuous time filter to  $y_1(k)$  using the associated parameters in Table 176E-7.
- Compute the linear fit pulse response  $p_2(k)$  using the method defined in 179.9.4.1.1 with parameter M the same as for step a)  $D_p$  equal to 3, and  $N_p$  equal to 200.

Editorial team proposal: Use Np=200, Nv=200, and Dp=4 for all interfaces. In 178.9.2.4, change TBD to 200. In agenda

#### 179.9.4.1.1 Linear fit to the measured waveform



Compute the linear fit pulse response p(k) and linear fit error e(k), k=1 to  $M \times N_p$ , from the captured waveform, as specified in 85.8.3.3.5, with  $N_p = 200$  and  $D_p = 4$ ] where the aligned symbols x(n) are assigned normalized amplitudes -1, -ES, ES, and 1 to represent the PAM4 symbol values 0, 1, 2, and 3 respectively. ES is defined as (ESI) + |ESS|/2 where ESI and ESS are calculated according to 120D.3.1.2.

 $D_p$  is the number of precursor in the linear fit and has traditionally been the 1 plus the number of precursor Tx FFE taps, enabling measurement of the coefficients.

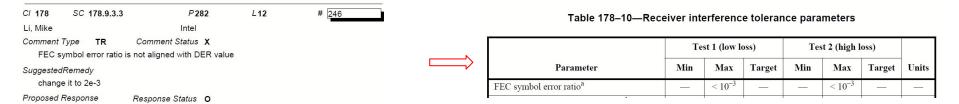
 $N_p$  is the linear fit pulse response length in UI.  $N_v$  is the length in UI taken for calculation of the steady-state voltage  $v_f$ .

Parameter	137.9.2, 120D.3.1	136.9.3.1	(163.9.2.4)	(162.9.4.1)
Np	200	200	200	200
Nv	13	13	200	200
Dp	2	3	4	4

### RX ITOL/JTOL Comment #246

In agenda

24



The comment addresses an inconsistency between FEC SER and DER.

This is a cross-clause topic:

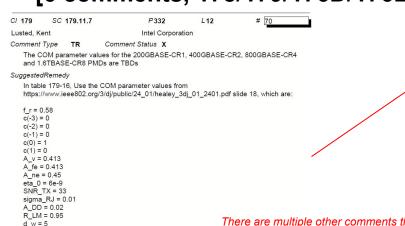
- For CR, the same value 1e-3 appears in Table 179-11.
- For C2C. Table 176D-4 has TBD instead.
- For C2M, the stressed input subclauses 176E.3.5.5 and 176E.3.6.5 have "The host under test shall meet the requirements in 176E.2"...
  - 176E.2 says "A 200 Gb/s per lane AUI-C2M component shall meet the error ratio requirements specified in 174A.X when it operates in DATA mode".

In conjunction with this topic, comment #205 has been resolved with an update of Annex 174A which will specify the data reliability requirement from PMDs and AUIs using the probability of uncorrectable FEC codewords with additional errors (option B). See ran 3di 04a 2405 slide 11 (PMDs) and slide 13 (AUIs).

The response to #205 says "Update clauses/annexes 171, 178, 179, 179D, 179E, 180 to 183, 185, 187 appropriately". The updates of electrical clauses/annexes will include the receiver test requirements.

Editorial team proposal: Resolve #246 using the response to comment #205.

# Multiple COM parameters [6 comments, 178/179/176D/176E]



Comments #70 and #71 propose a large set of COM parameter values together for CR and KR, respectively.

Comment #72 proposes a similar set for C2M and suggests adding a table instead of referring to the parameters of C2C.

The referenced presentation, <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 01/healey 3dj 01 2401.pdf, does not provide rationale for the suggested remedies of these comments. Specifically, slide #18 states that the values used are not a baseline proposal.

Comment #504 proposes a set of parameters for C2C, following analysis in https://www.ieee802.org/3/dj/public/24 05/heck 3dj 01b 2405.pdf.

Comments #42 and #54 do not include a specific proposal (but suggest a series of straw polls).

There are multiple other comments that may result in adopting values for some of the proposed parameters. These comments may to be resolved partially by the responses to the other comments.

Note that the presentations cited did not examine the effect of constraining the Rx FFE parameters (Wmin/Wmax), and disabled Tx equalization (c(i)). The assumption that receiver can operate with large Rx FFE coefficient values and without Tx FFE equalization may not be in consensus.

### Editorial team proposal:

additionally, set MLSE = 0 (not enabled)

Response Status O

Nfix = 10

 $b_max(1) = 0.85$  $b_min(1) = 0$ 

Proposed Response

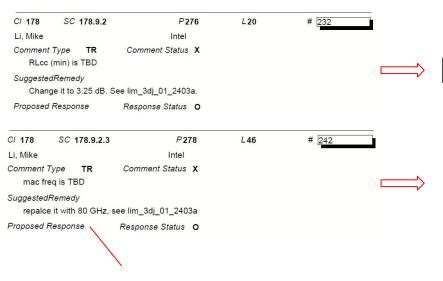
N\_g = 0 N\_f = 0 N max = 0

- REJECT comments #42 and #54 (no actionable remedy within the draft).
- Resolve the other comments after all other electrical comments, to possibly adopt values for parameters that have not been addressed by other comments and may be in consensus.
- Consider adding a COM table in Annex 176E as suggested by comment #72.
- If necessary, add editor's notes similar to that in slide 4 of <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 05/lusted 3dj 07 2405.pdf to denote values that need further analysis.

### Clause 178

### Clause 178, RLcc (#232, #242)

In agenda



The referenced presentation (200G/Lane KR Baseline for 802.3dj, <a href="https://www.ieee802.org/3/dj/public/24">https://www.ieee802.org/3/dj/public/24</a> 03/lim 3dj 01a 2403.pdf) does not include an explanation of the proposed values.

The value suggested by #232 is the same as in 802.3ck. Table 163-5.

### Table 178-6—Summary of transmitter specifications at TP0v

	1	I .	
Common-mode to common-mode return loss, RLcc (min)	178.9.2.3	TBD	dB

#### 178.9.2.3 Transmitter common-mode to common-mode return loss

The common-mode to common-mode return loss shall be greater than or equal to *RLcc* (min) in Table 178–6 at all frequencies between 0.2 GHz and TBD GHz.

Editor's note (to be removed by D2.0, or if values are adopted):
Common-mode to common-mode return loss requirements were left to be determined in the adopted baseline proposal. Contributions in this area are encouraged.

The measurement bandwidth for other specifications is addressed by comment #60, which was resolved with a bandwidth of 60 GHz.

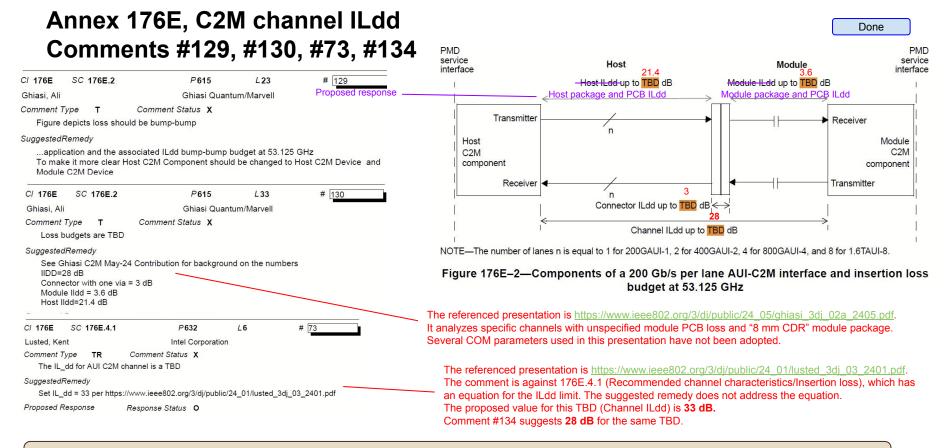
It seems that specifications of Tx RLcc should be within the same bandwidth.

Editorial team proposal:

Accept #232.

Resolve #242 using the response to comment #60.

### Annex 176E



Editorial team proposal:

Reject #130, #73 and #134 due to the large difference. Continue working toward consensus loss budget for C2M in conjunction with COM parameters.

### **Annex 176E, Input specs (#188, #189)**

In agenda

C/ 176E SC 176E.3.5 P624 L3 # [188]

Ran, Adee Cisco

Comment Type TR Comment Status X

Host input characteristics need to be defined with consideration of the availability of training.

This will affect the entire subclause 176E.3.5.

#### SuggestedRemedy

Define the input characteristics using a methodology similar to that of receiver specifications in 179.9.5, with the required changes due to the lack of a cable assembly.

Use a table similar to Table 179-10 but with additional rows for DC common-mode voltage and AC common-mode voltage tolerance.

A detailed proposal will be provided

Proposed Response Response Status O

C/ 176E SC 176E.3.6 P628 L26 # [189 Ran, Adee Cisco

Comment Type TR Comment Status X

Module input characteristics need to be defined with consideration of the availability of training.

This will affect the entire subclause 176F 3.6.

#### SuggestedRemedy

Define the input characteristics using a methodology similar to that of receiver specifications in 179.9.5, with the required changes due to the lack of a cable assembly and usage of MCB instead of HCB.

Use a table similar to Table 179-10 but with additional rows for DC common-mode voltage tolerance and AC common-mode voltage tolerance.

A detailed proposal will be provided.

Proposed Response Response Status O

The detailed proposal mentioned is https://www.ieee802.org/3/di/public/24 05/ran 3dj 01 2405.pdf.

The presentation addressed both host input and module input, and proposes using a methodology based on the CR Amplitude tolerance, Interference tolerance, and Jitter tolerance tests for these specifications, with appropriate changes to the host/module test setups.

The specific proposals are on slides 6-9 of the referenced presentation.

The following straw poll was taken in the May 2024 interim meeting:

#### Straw Poll #2

I would support the approach for the AUI-C2M host and module input specifications outlined in ran\_3dj\_01\_2405 Results (all) Y: 31, N: 15, NMI: 6, A: 39

The results show significant support for using the CR methodology for AUI-C2M host and module input specs.

Note that <a href="https://www.ieee802.org/3/dj/public/24\_06/dawe\_3dj\_01a\_2406.pdf">https://www.ieee802.org/3/dj/public/24\_06/dawe\_3dj\_01a\_2406.pdf</a> seems to oppose this direction, but no other comments related to input specifications were submitted.

### Editorial team proposal:

Implement the proposed changes on slides 6-9 of ran 3dj 02 2405 with editorial license.

### Annex 176E, Output specs (#186, #187, #203)

In agenda

C/ 176E SC 176E.3.3 P617 L10 # 186 Ran, Adee Cisco Comment Type TR Comment Status X Host output characteristics need to be defined with consideration of the variable output settings that can result from training. This will affect the entire subclause 176E.3.3. SuggestedRemedy Define the output characteristics using a methodology similar to that of transmitter specifications in 179.9.4. Use a table similar to Table 179-7 but with different values due to the higher host channel insertion loss budget for C2M. A detailed proposal will be provided. Proposed Response Response Status O C/ 176E SC 176E.3.4 P621 L13 # 187

Ran, Adee Cisco

Module output characteristics need to be defined with consideration of the variable output settings that can result from training.

This will affect the entire subclause 176E.3.4.

#### SuggestedRemedy

Comment Type

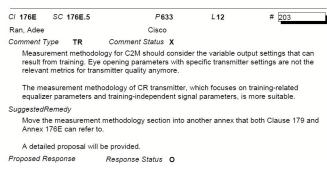
Define the output characteristics using a methodology similar to that of transmitter specifications in 179.9.4.

Comment Status X

Use a table similar to Table 179-7 but with different values due to the lower insertion loss assumed for the module output test.

A detailed proposal will be provided.

Proposed Response Response Status O



The detailed proposal mentioned is https://www.jeee802.org/3/di/public/24 05/ran 3di 02 2405.pdf.

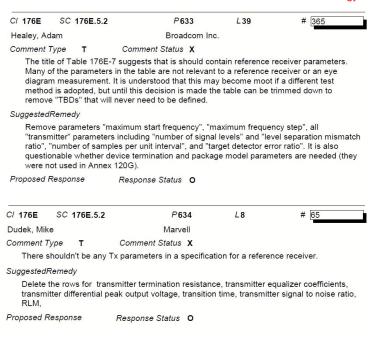
The presentation addressed both host output and module output, and proposes using the CR methodology for these specifications (instead of VEC/EH measurements) and rewriting the "Measurement methodology" section 176E.5. The specific proposals are on slides 6, 8, 10, and 11.

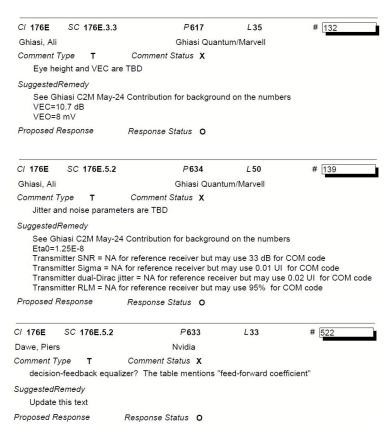
### Annex 176E, Output specs (#65, #132, #139, #365, #522)

In agenda

Rewriting 176E.5 as proposed would remove the "Eye opening reference receiver parameter values" table (Table 176E–7), which is the subject of several comments.

### Comment #132 about the EH/VEC values will be OBE if the CR methodology is adopted.





### **Annex 176E, Output specs**

In agenda

The following straw poll was taken in the May 2024 interim meeting:

#### Straw Poll #3

I would support the approach for the AUI-C2M host and module output specifications outlined in ran 3dj 02 2405 Results (all): Y: 38, N: 9, NMI: 9, A: 42

The results show significant support for using the CR methodology for AUI-C2M host and module output specs.

Note that https://www.ieee802.org/3/dj/public/24 06/dawe 3dj 01a 2406.pdf seems to oppose this direction.

### Editorial team proposal:

- In response to #186, AIP: implement the proposed changes on slides 6, 8, 10, and 11 of ran\_3dj\_02\_2405 with editorial license.
  - For jitter method and parameters: Use the values on slide 21 (jitter comments) or leave the limits TBD if there is no consensus.
- For #187 and #203, AIP: resolve using the response to #186.
- For #65, #139, #365, and #522, AIP: the text subject of the comment will be replaced by the response to #186.
- #132: AIP, the specification is replaced by the response to #186.

### Annex 178A

### Annex 178A, DER0 (1 of 2) Comments #285, #362

Done

Cl 178A SC 178A.1.10.2 P659 L12 # 285

Li, Mike Intel

Comment Type TR Comment Status X

DER0 EQ is wrong

SuggestedRemedy

change P(y0)= DER0 to 1-P(y0) =DER0, see slide 3 of lim\_3dj\_02\_2405, see also a marked version in the support data sheet.

Proposed Response Status O

C/ 178A SC 178A.1.10 P658 L43 # 362
Healey, Adam Broadcom Inc.

Comment Type T Comment Status X

The relationship between "detector error ratio", "PAM-L symbol error ratio", and "bit error ratio" is not documented and, as a result, not generally understood. While these quantities are related, they are not interchangeable. Prior assumptions that they are interchangeable has led to errors in the translation between COM results and expected (measured) receiver performance. This new annex gives us an opportunity to clarify the relationship between DERO and other terms or to replace DERO with a more generally understood term.

#### SuggestedRemedy

Slide 5 of <a href="https://www.ieee802.org/3/dj/public/23\_11/healey\_3dj\_01a\_2311.pdf">healey\_3dj\_01a\_2311.pdf</a> suggest expressions for relationship between detector error ratio and other terms. Either replace "DER0" with a target PAM-4 symbol error ratio (or bit error ratio) and adjust the equations for calculating COM accordingly, or document the relationship between DER0 and the other two terms.

Proposed Response Status O

It may not be clear that the solution for  $y_0 = P^{-1}(DER_0)$  is less than 0 for any  $DER_0 < 0.5$ 

It is specified that the magnitude of the result is used

Expressions like  $y_0 = P^{-1}(1-DER_0)$  or  $DER_0 = 1-P(y_0)$  are used elsewhere in Annex 178A to make  $y_0$  greater than 0

Although also correct, use of different forms may lead to further confusion

It would be better to add clarity to, and consistently use, an expression to map between a CDF value and the corresponding amplitude value

In addition, there has been confusion about the relationship between  $DER_0$  and PAM-L symbol error ratio (or BER) and it has been suggested that this relationship be clarified

### Annex 178A, DER0 (2 of 2) Comments #285, #362

Editorial team proposal:

Implement the changes shown in this slide, with editorial license.

### 178A.1.10.2 Noise and interference amplitude

The probability distribution function of the noise and interference amplitude is calculated using the procedure defined in 93A.1.7.3 using the sampled time-domain responses and noise variance defined in 178A.1.9. The corresponding cumulative distribution function is defined by Equation (178A–35).

$$P(y) = \int_{-\infty}^{y} p(u)du$$
 [Use dummy variable in integrand.] (178A–35)

[Re-write to clarify that the expected result is a positive number.] The noise and interference amplitude,  $A_{ni'}$  is a positive value that satisfies the relationship  $P(\neg A_{ni}) = DER_0$  where  $DER_0$  is the target detector error ratio.

[Add note to clarify the relationship between DER and other error ratio metrics.] NOTE 2—The target detector error ratio  $DER_0$  is  $SER_1 \times L/(2L-2)$  where  $SER_1$  is the maximum allowed probability of the initial error in a PAM-L error burst.

[Change Equations (178A–36) and (178A–37), as modified by comments #211, #212, #286, and #287, to use expressions similar to what is used in 178A.1.10.2.]

$$COM_{MLSD} = COM_{DEF} + 20\log_{10}(-P^{-1}(DER_{MLSD})/A_s) - Q$$
 (178A–36)

$$DER_{MLSD} = \sum_{j=1}^{\infty} \left(\frac{L-1}{L}\right)^{j-1} P_j \left(-A_s \frac{\left(u_j^T u_j\right)^{3/2}}{\left(u_j^T V_j u_j\right)^{1/2}}\right)$$
(178A–37)

Done

## Clause xxx, <topic> Comment #<n>