

# Codeword Error Rate and (Hyper)-Spherical TDECQ

*Addressing comments 1, 343, 345, 347 and 349 against P802.3dj D2.0*

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# AGENDA

- 1. TDECQ Challenges**
- 2. TDECQ in the Current Draft (P802.3dj D2.0)**
- 3. Codeword Error Rate TDECQ**
- 4. A Geometric Perspective**
- 5. Approximating CER TDECQ – Hyper-Spherical TDECQ**
- 6. Experimental Data**

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# TDECQ Challenges

- Concerns are being raised around TDECQ's correlation to link performance has been raised amongst members of the IEEE 802.3dj task force.
- [chayeb\\_3dj\\_01\\_2505](#) documented some of the TDECQ challenges and potential improvements.
- [chayeb\\_3dj\\_01\\_2505](#) demonstrated experimental data showing lack of correlation between test methodology (TDECQ) and actual link performance (post FEC errors).
- [chayeb\\_3dj\\_01\\_2505](#) and [ghiasi\\_3dj\\_03\\_2501](#) proposed a Codeword Error Rate TDECQ calculated at a target CER (codeword error rate) instead of the current SER (symbol error rate) target.
- This presentation proposes a TDECQ implementation that captures correlation of errors due to effects that are present in the waveform in an economically feasible way.
- This presentation addresses comments *1, 343, 345, 347* and *349* against IEEE P802.3dj D2.0.

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# TDECQ in the Current Draft (P802.3dj D2.0)

**Goal of TDECQ:** Find the maximum intrinsic receiver noise that still achieves the desired error performance (currently specified as SER).

The methodology specified in the standards:

- Combines samples from each sample location, into histograms which represent an estimated pdf of the signal levels at the sample location,  $f_{y_k}(y_k)$
- Let  $n \sim N(0, \sigma^2)$  be a normal gaussian, then the SER is a function of  $\sigma$  defined as:

$$SER(\sigma) = \sum_{l=0}^{l=3} \sum_k f_{y_k}(y_k) \left[ Q\left(\frac{y_k - P_{th(l-1)}}{\sigma}\right) + Q\left(\frac{P_{th(l)} - y_k}{\sigma}\right) \right]$$

- The maximum  $\sigma$  is found such that  $SER(\sigma) \leq SER_{target}$

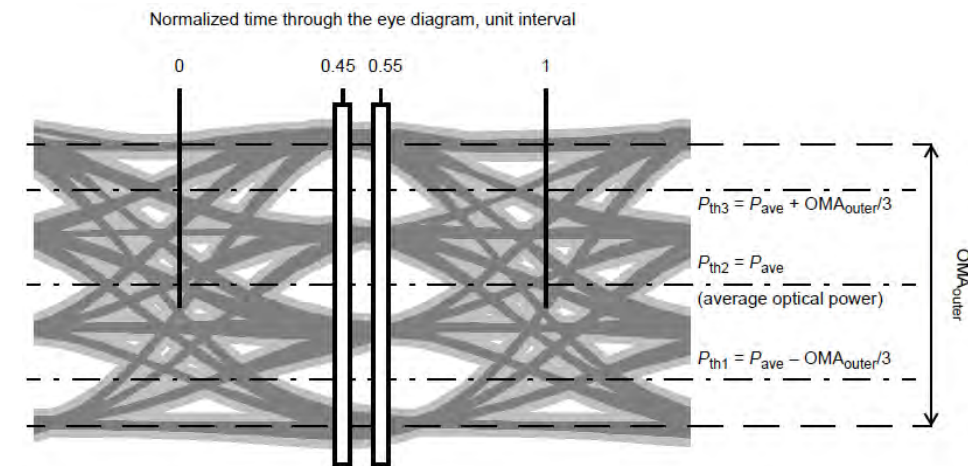


Figure 121-5—Illustration of the TDECQ measurement

IEEE 802.3bs – Clause 121.8.5.1

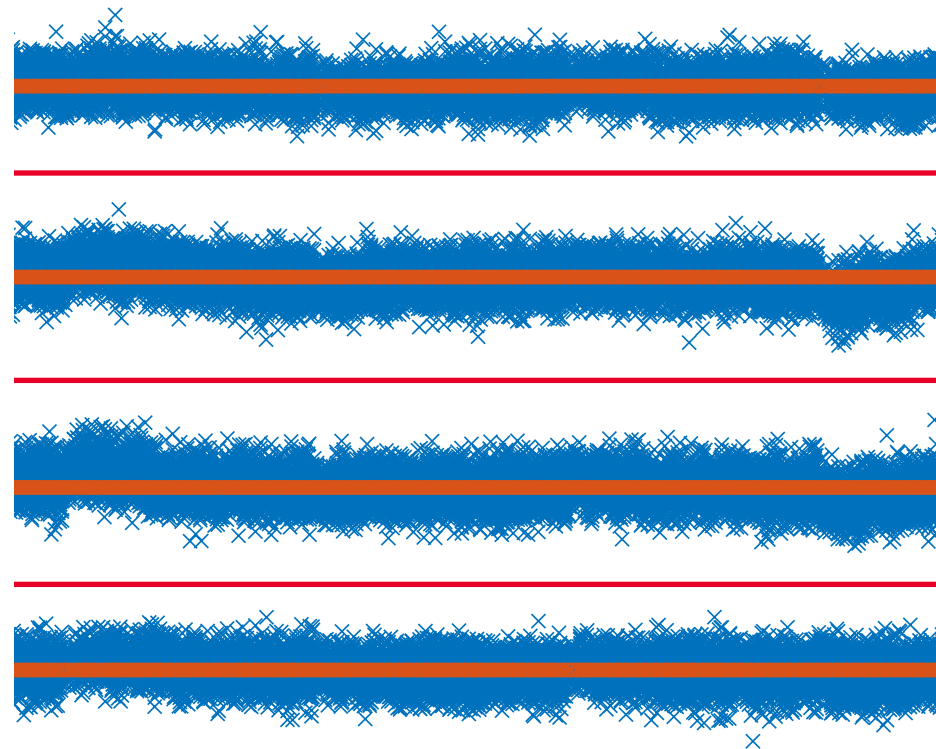
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# Codeword Error Rate TDECQ

## Unwrapping the Histogram

- An alternative, but less efficient, methodology is to unwrap the level histograms as follows:
- Let:
  - $i$  be the symbol index
  - $x_i$  a sample from the target region of the symbol
  - $l_i$  the symbol value (0,1,2,3)
  - $N$  the number of symbols
  - $V(l_i)$  is the nominal level of the symbol
  - $th_{l_i}$  is the optimal threshold below the  $l_i$  level
- The figure shows  $x_i$  in blue,  $V(l_i)$  in orange, and  $th_{l_i}$  in red.





# Codeword Errors

## Unwrapping the Histogram

- We can define  $SER(\sigma) = \frac{1}{N} \sum_{i=0}^{N-1} \left[ Q \left( \frac{x_i - th_{l_i-1}}{\sigma} \right) + Q \left( \frac{th_{l_i} - x_i}{\sigma} \right) \right]$
- The above can be dissected to get a probability of error per symbol  $i$  assuming  $\sigma$ :
- $P_{err,i}(\sigma) = SER_i(\sigma) = \left[ Q \left( \frac{x_i - th_{l_i-1}}{\sigma} \right) + Q \left( \frac{th_{l_i} - x_i}{\sigma} \right) \right]$
- $P_{err,i}(\sigma)$  is then the probability of error for the specific instance of  $x_i$  assuming receiver noise power  $\sigma$
- Each  $x_i$  includes ISI, transmitter and scope noise.
- If symbols are organized into codewords of length  $d$  and we assume that the FEC can correct up to  $K$  errors a probability of codeword error for a given  $\sigma$  can be calculated using  $P_{err,i}(\sigma)$

# Codeword Errors

## Probability of $k$ Errors in a Codeword of length $d$ , $C_{k,d}(\sigma)$

- For a set of  $d$  symbols,  $P_{err,i}(\sigma)$  is the corresponding probability of error of the  $i$ th symbol
- The number of errors in the codeword follows a Poisson Binomial distribution
  - An extension of the Binomial distribution, where the probabilities of each trial are not identical
- This distribution,  $C_{k,d}(\sigma)$ , can be calculated by convolving the individual symbol PMF's together:
  - The  $i$ th symbol has the PMF:  $p_i(k) = \begin{cases} 1 - P_{err,i}(\sigma) & k = 0 \\ P_{err,i}(\sigma) & k = 1 \end{cases}$ , where 1 indicates an error
  - $C_{k,d}(\sigma) = p_0(k) * p_1(k) * \dots * p_{d-1}(k)$
- Once the PMF is calculated, the probability of a correctable codeword is found by summing the first  $K$  entries of  $C_{k,d}(\sigma)$
- And the probability that the  $j$ th codeword is in error is:

$$P_{fail,j}(\sigma) = 1 - \sum_{k=0}^{K-1} C_{k,d}(\sigma)$$

## Codeword Errors (con't)

- A  $d$  symbol codeword can be made up of any  $d$  distinct symbols from the pattern
  - There will be  $M = \left\lfloor \frac{N}{d} \right\rfloor$  available blocks
  - The Codeword Error Rate is then the mean of  $P_{fail,j}(\sigma)$  over all blocks:

$$CER(\sigma) = \frac{1}{M} \sum_{j=0}^{M-1} P_{fail,j}(\sigma)$$

# Codeword Error Rate Complexity

- The inner FEC codeword is 64 symbols long
  - An SSPRQ is  $2^{16} - 1$  symbols
  - So, at least  $M = 1023$  codewords need to be evaluated
  - Each codeword PMF requires 63 convolutions given a specific  $\sigma$  value
  - So, for each iteration using a  $\sigma$  value 64,449 convolutions are required
- The existing TDECQ method requires 1 convolution per  $\sigma$  value
  - There is additional computation to convert samples into histograms
- The complexity of the algorithm to exactly compute the Codeword Error Rate appears to be daunting

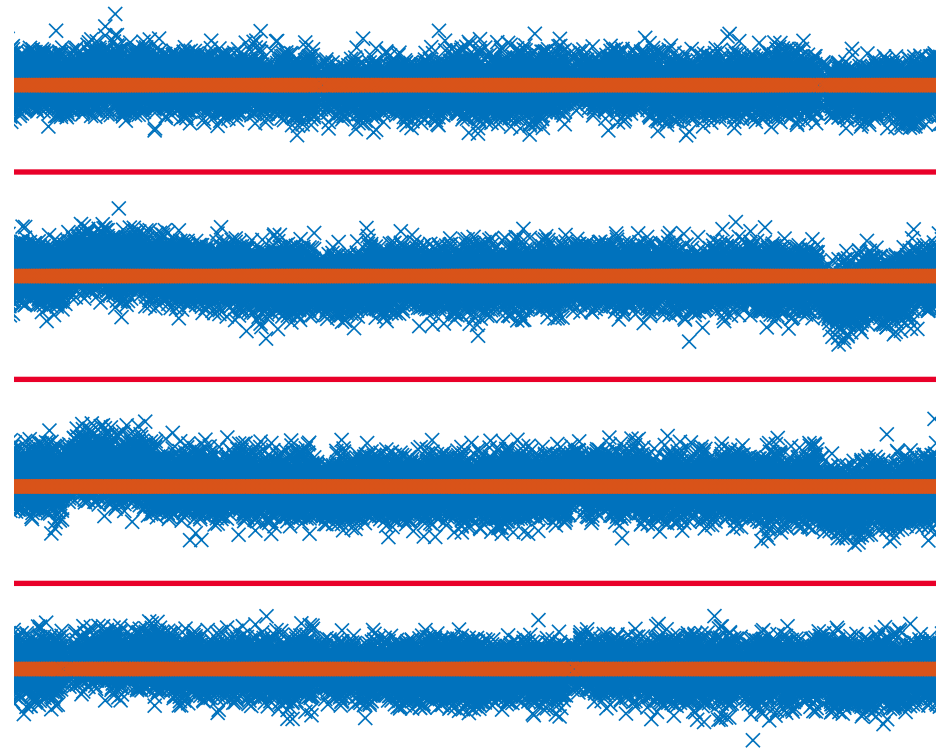
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# A Geometric Perspective

Treat codewords as vectors in a d-dimensional space

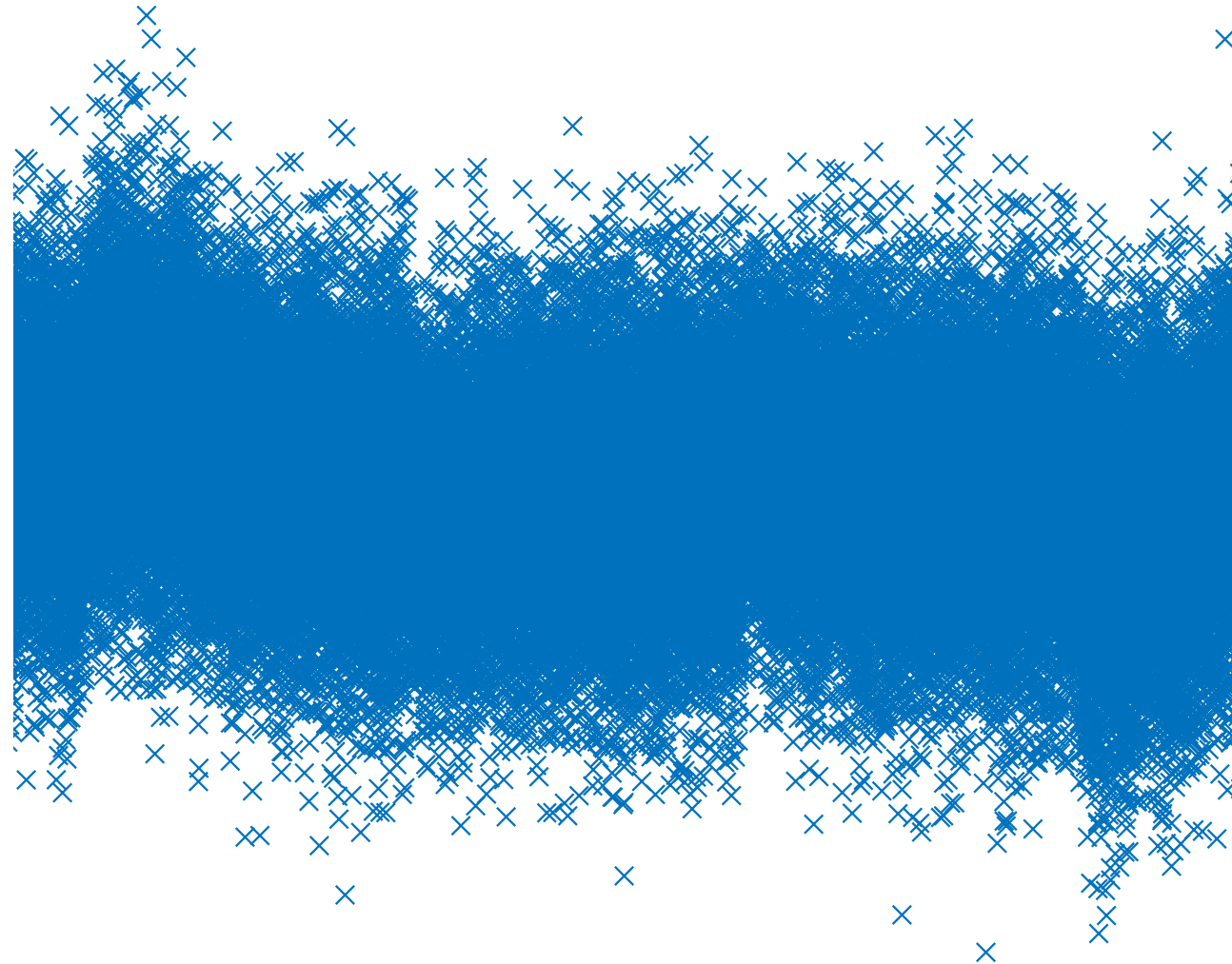
- Let:
  - $i$  be the symbol index
  - $x_i$  a sample from the target region of the symbol
  - $l_i$  the symbol value (0,1,2,3)
  - $N$  the number of symbols
  - $V(l_i)$  is the nominal level of the symbol
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- The figure shows  $x_i$  in blue,  $V(l_i)$  in orange, and  $th_{l_i}$  in red.



# Defining the Residual Vector

Center analysis at the origin

- Define the residual,  $\mu_i = x_i - V(l_i)$
- We can group the residuals to create a d-dimensional residual vector,  
➤ eg  $\bar{\mu} = [\mu_0, \mu_8, \dots, \mu_{8(d-1)}]$
- The above example uses the interleaving of 8 symbol as specified in the standard for FECi.



# A Geometric Perspective

## The Noise Vector

- Let  $\bar{n} \sim N(0, \Sigma)$  be a d-dimensional vector of the noise to be added,  $\Sigma$  is the covariance matrix of the added noise.
  - Assumed to be AWGN filtered by a 4<sup>th</sup> Order Bessel Thompson and any specified equalizers
- Combining the residual vector and noise vector results in the random variable  $\bar{y} = \bar{\mu} + \bar{n}$ 
  - Where the residual is the mean and having a pdf,  $f_{\bar{y}}(\bar{y}) = \frac{1}{\sqrt{(2\pi)^d |\Sigma|}} e^{-\frac{1}{2}(\bar{y}-\bar{\mu})^T \Sigma^{-1}(\bar{y}-\bar{\mu})}$
- **Note:** If the data samples come from sufficiently separated UI and only linear equalizers are used then independence of noise samples can be assumed, i.e.  $\Sigma \approx \sigma^2 \mathbf{I}$ , where  $\mathbf{I}$  is the identity matrix.



# Probability of k Errors

- Can “easily” calculate the probability of 0 errors as:

$$P(0 \text{ errors}) = \int_R \frac{1}{\sqrt{(2\pi)^d |\Sigma|}} e^{-\frac{1}{2}(\bar{y}-\bar{\mu})^T \Sigma^{-1}(\bar{y}-\bar{\mu})} d\bar{y}$$

- where  $R$  is a  $d$ -dimensional region centered at 0, where each dimension is bound by the upper and lower threshold (or +/- infinity) for that symbol.
- **NOTE: Each dimension has its own limit based on  $l_i$**
- For  $k > 0$ , things are much more complicated
  - Need to take all regions  $R$  such that  $k$  dimensions are outside the thresholds and the other  $d-k$  dimensions are inside their respect thresholds.
  - If  $\Sigma \approx \sigma^2 \mathbf{I}$  this results in the same results as the previous section
- For simplicity of analysis, assume that all thresholds are  $\pm \frac{OMA}{6}$  from the nominal symbol level
  - Now, all symbol errors can be thought of as either too high or too low.

# Probability of k Errors

## Two-Dimensional Visualization

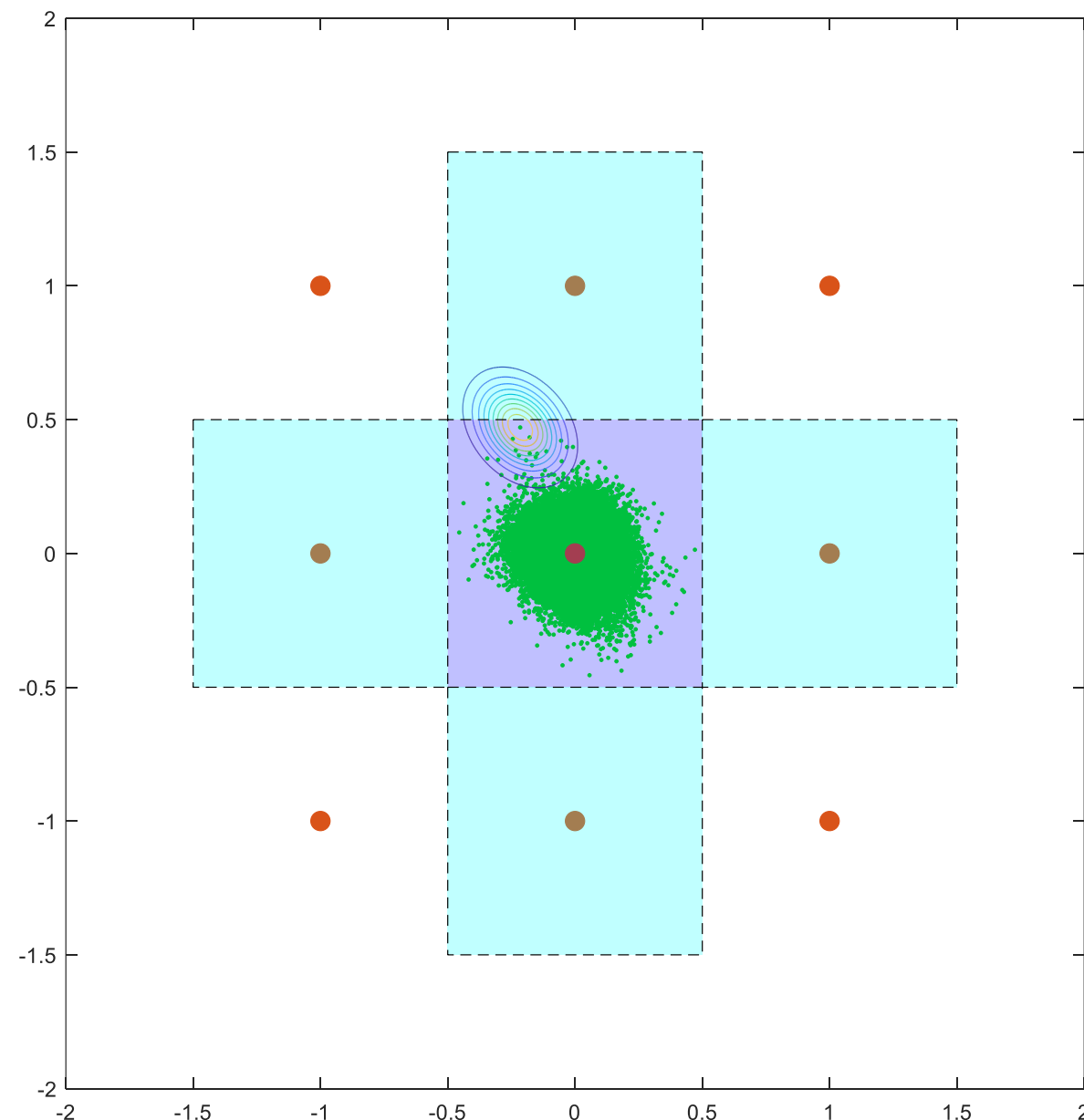
- The probability of exactly k errors is:

$$p_k = \sum_{c \in C_d(k)} \int_c \frac{1}{\sqrt{(2\pi)^d |\Sigma|}} e^{-\frac{1}{2}(\bar{y} - \bar{\mu})^T \Sigma^{-1} (\bar{y} - \bar{\mu})} d\bar{y}$$

- Where  $C_d(k)$  is a d-dimensional hypercube over the codeword space with exactly k symbol errors.

- Then the probability of k or more errors is

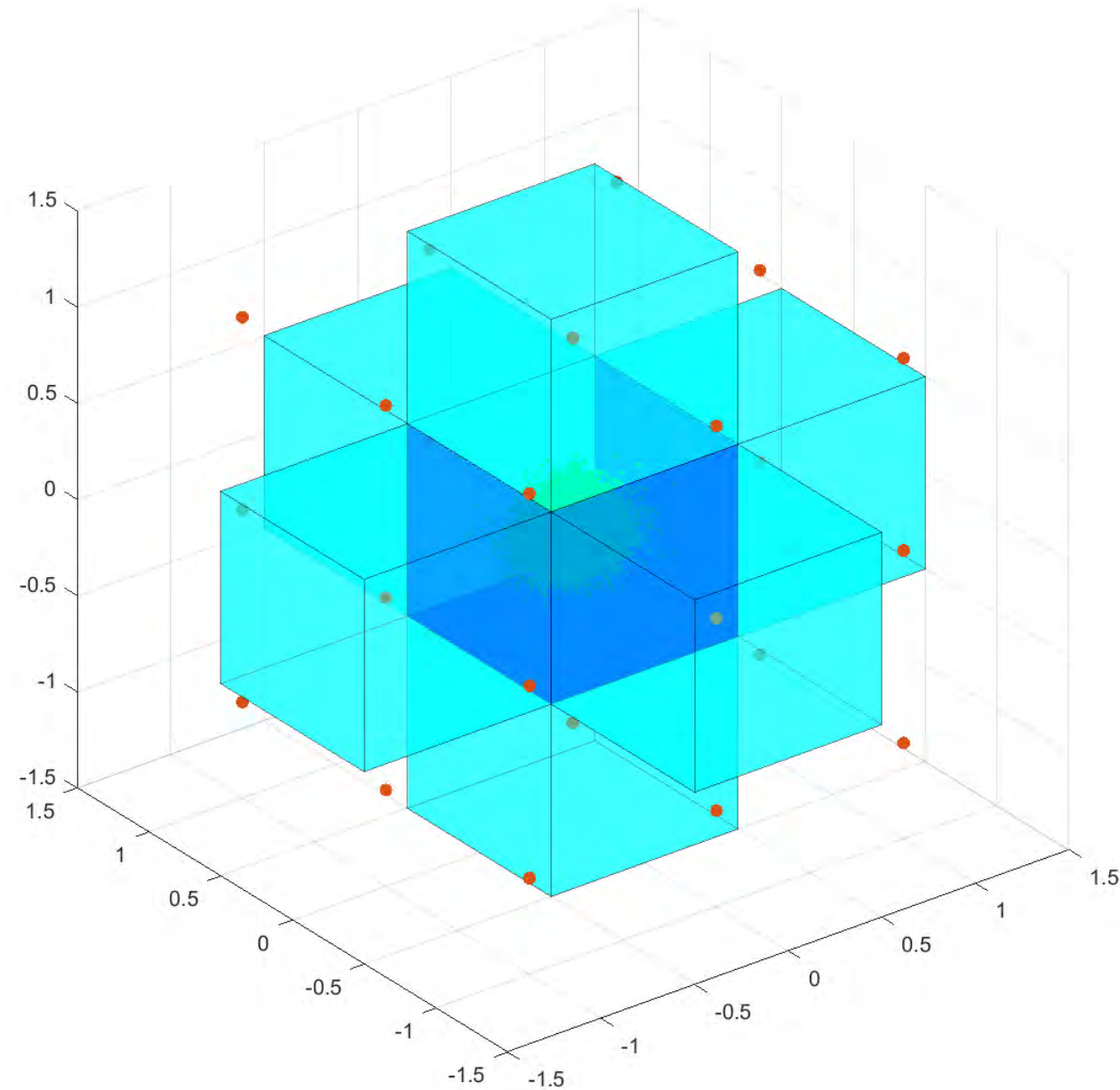
$$P_{fail} = 1 - \sum_{j=0}^{k-1} p_j$$



# Probability of k Errors

## Three-Dimensional Visualization

- An example of our simplified model using just  $d=3$  length codewords.
- The Red dots are the potential codewords
- Blue box indicates region of 0 errors per code word
- Cyan boxes indicate region with 1 error per code word
- Green dots are the residuals,  $\bar{\mu}$ , from data samples



# Regions with k Errors

## Two-Dimensional Visualization

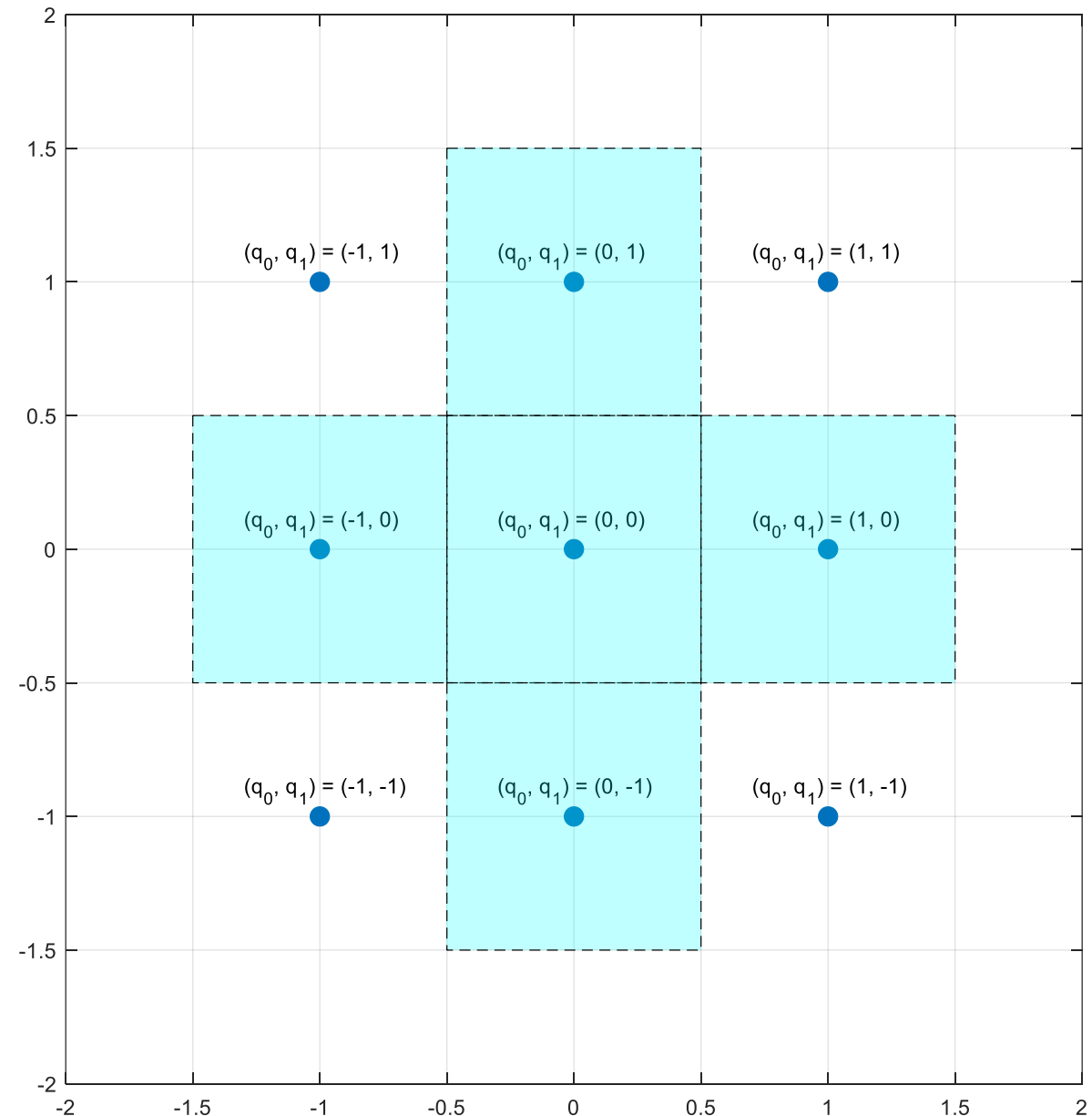
- Define a set of d-dimensional hypercubes,  $C_d(k)$ , all with sides of length  $\frac{OMA}{3}$

- Each region is centered at the point  $\frac{OMA}{3} [q_0, q_1, \dots, q_{d-1}]$ ,  $q_i \in \{-1, 0, 1\}$

where,

$$\sum_{i=0}^{d-1} |q_i| = k$$

- The number of such regions is  $|C_d(k)| = 2^k \binom{d}{k}$
- 64 symbols and up to 3 errors results in over 333312 regions
- This appears to be an even more complex process



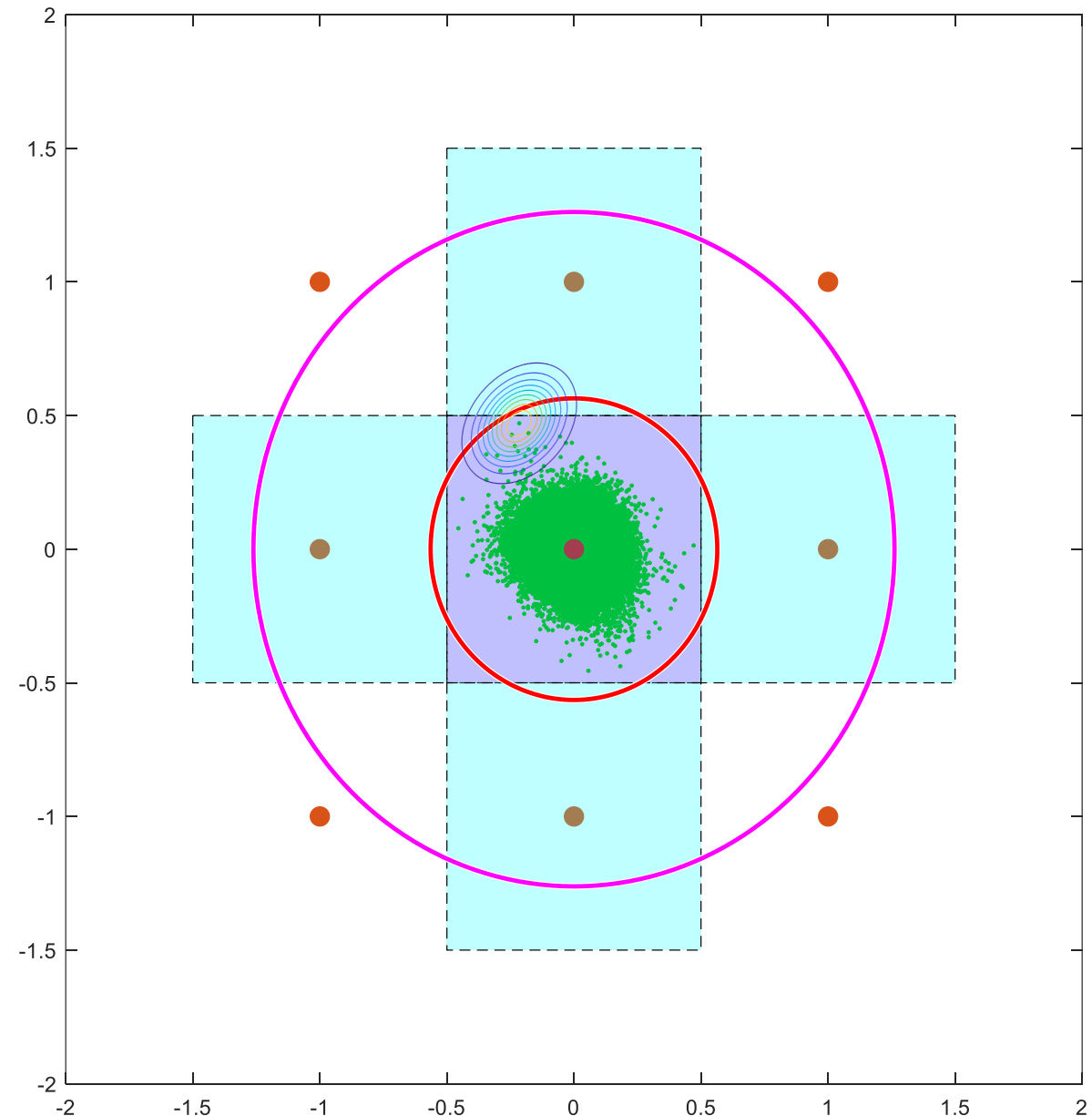
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# Can We Make it simpler?

## Going Back to the Two-Dimensional Example

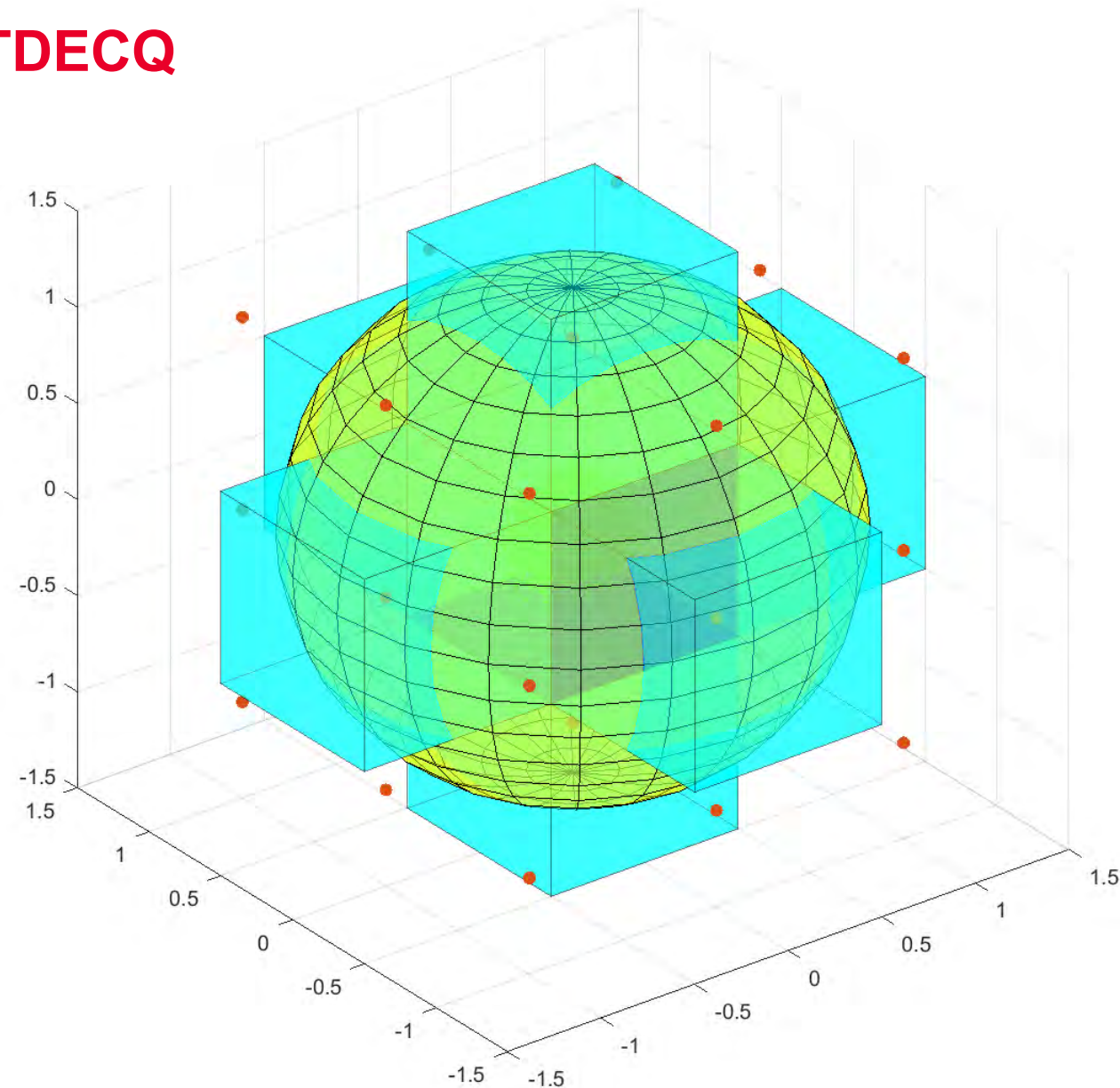
- Approximate the set of hypercubes as a single hypersphere
- Focus instead on the Euclidean distance from the ideal codeword location
- An estimate for the codeword error rate could be the probability that  $\bar{y} = \bar{\mu} + \bar{n}$  is contained within a sphere of radius  $r_k$  centered at the ideal codeword location
- Essentially,  $\text{CER} \approx 1 - \Pr\{\bar{y}^T \bar{y} \leq r_k^2\}$



# Approximating Codeword Error Rate TDECQ

## Reduction to a Singular Dimension

- Define a new variable:  $r^2 = \frac{\bar{y}^T \bar{y}}{\sigma^2}$
- For the case where  $\bar{y} \sim N(\bar{\mu}, \Sigma)$ ,  $r^2$  follows a generalized  $\chi^2$ -distribution
- If  $\Sigma \approx \sigma^2 \mathbf{I}$ , then  $r^2$  follows the non-central  $\chi^2$ -distribution with  $d$  degrees of freedom and non-centrality parameter  $\lambda = \frac{\bar{\mu}^T \bar{\mu}}{\sigma^2}$
- With cdf  $F_{r^2}(r^2; d, \lambda)$
- Then an approximation for the probability of at most  $k$  errors per codeword can be written as:  $F_{r^2}\left(\frac{r_k^2}{\sigma^2}; d, \lambda\right)$ . Where  $r_k$  is the radius of the circle representing at most  $k$  errors.



# An Algorithm

## (Hyper)-Spherical TDECQ

- Let  $r_k$  be the target radius for up to k errors per codeword
- Let  $\bar{\mu}_i$  be the  $i$ th vector of sampled residuals out of N total vectors
- Define the probability of a correctable codeword for a given  $\sigma$  as:

$$G(\sigma) = \frac{1}{N} \sum_{i=0}^{N-1} F_{r^2} \left( \frac{r_k^2}{\sigma^2}; d, \frac{\bar{\mu}_i^T \bar{\mu}_i}{\sigma^2} \right)$$

➤ Find the maximum noise,  $\sigma_g$ , such that  $(1 - G(\sigma_g)) \leq CER_{target}$

➤ Calculate  $\sigma_s = \sqrt{C_{eq}^{-2} \sigma_g^2 + \sigma_c^2}$  to remove noise gain and include intrinsic channel noise

- $TDECQ = 10 \log_{10} \left( \frac{\sigma_{ref}}{\sigma_s} \right)$ , where  $\sigma_{ref}$  is the noise margin of an ideal transmitter
- Note: For d=1, this can be reduced to an equivalent calculation of traditional TDECQ



# Additional Considerations

## Choice of $r_k$

- Choose  $r_k$  such that  $1 - CER_{target} = F_{r^2} \left( \frac{r_k^2}{\sigma_{ref}^2}; d, 0 \right)$ 
  - The probability of a successful codeword is equal to the probability of being within the hypersphere of radius  $\frac{r_k}{\sigma_{ref}}$
  - $r_k = \sigma_{ref} \sqrt{F_{\chi_d}^{-1}(1 - CER_{target})}$
- $\sigma_{ref}$  is calculated similarly to the existing method, assuming an idealized system
  - Assume symbol errors are iid with error probability  $P_e$
  - $\sigma_{ref} = \frac{OMA}{6\sqrt{2} \operatorname{erfc}^{-1}\left(\frac{4}{3}P_e\right)}$
  - $P_e$  can be derived from  $CER_{target}$  as the solution to  $1 - CER_{target} = \sum_{i=0}^k \binom{n}{i} P_e^i (1 - P_e)^{n-i}$

# Additional Considerations

## Efficient Calculation of Non-Central Chi-Square CDF

- The non-central Chi-Square CDF,  $F_{r^2}(r^2; d, \lambda)$ , does not have a closed form solution
- “Exact” Numerical solutions exist, but are computationally complex
- Tables could be generated for pairs of values,  $(r, \lambda)$ , to speed up computations
- For calculations in this presentation the following estimate<sup>1</sup> is used:

$$F_{r^2}(r^2; d, \lambda) \approx \frac{1}{2} \operatorname{erfc} \left[ \frac{1}{\sqrt{2}} \left( r - \sqrt{\lambda} - \frac{d-1}{2(r - \sqrt{\lambda})} (\log(r) - \log(\sqrt{\lambda})) \right) \right]$$

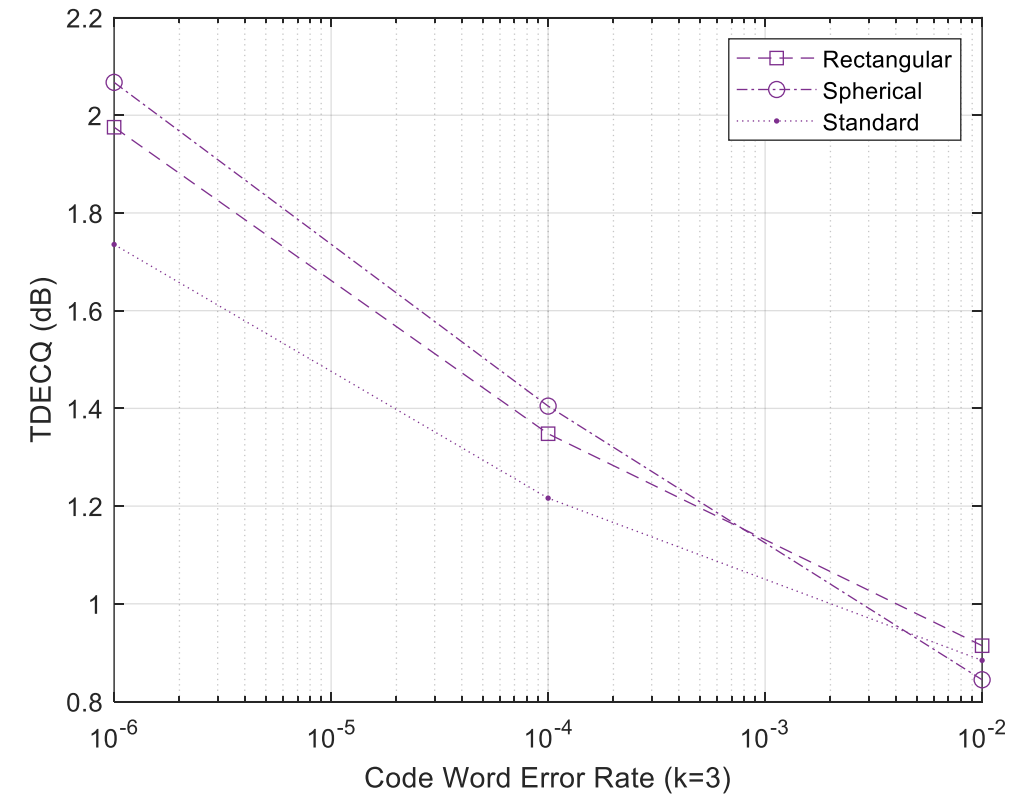
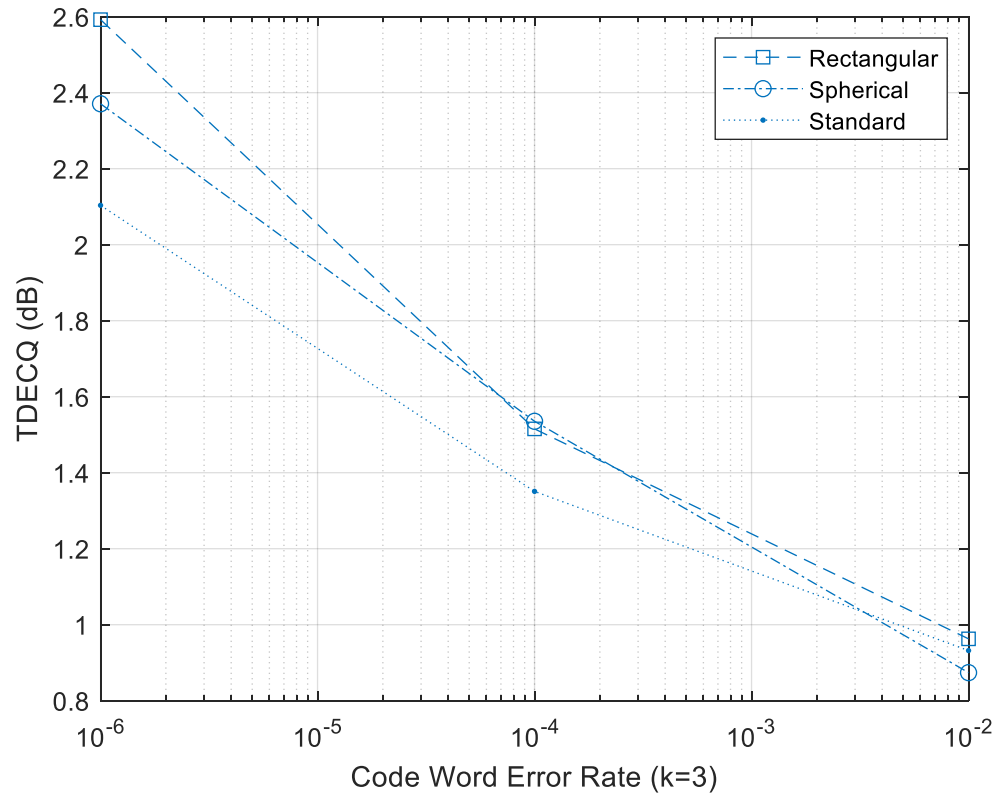
1 - Fraser, D.A.S. & Wu, Jianrong & Wong, Augustine. (1998). An approximation for the noncentral chi-squared distribution. Communications in Statistics - Simulation and Computation. 27. 275-287. 10.1080/03610919808813480.

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# Comparison of Methods

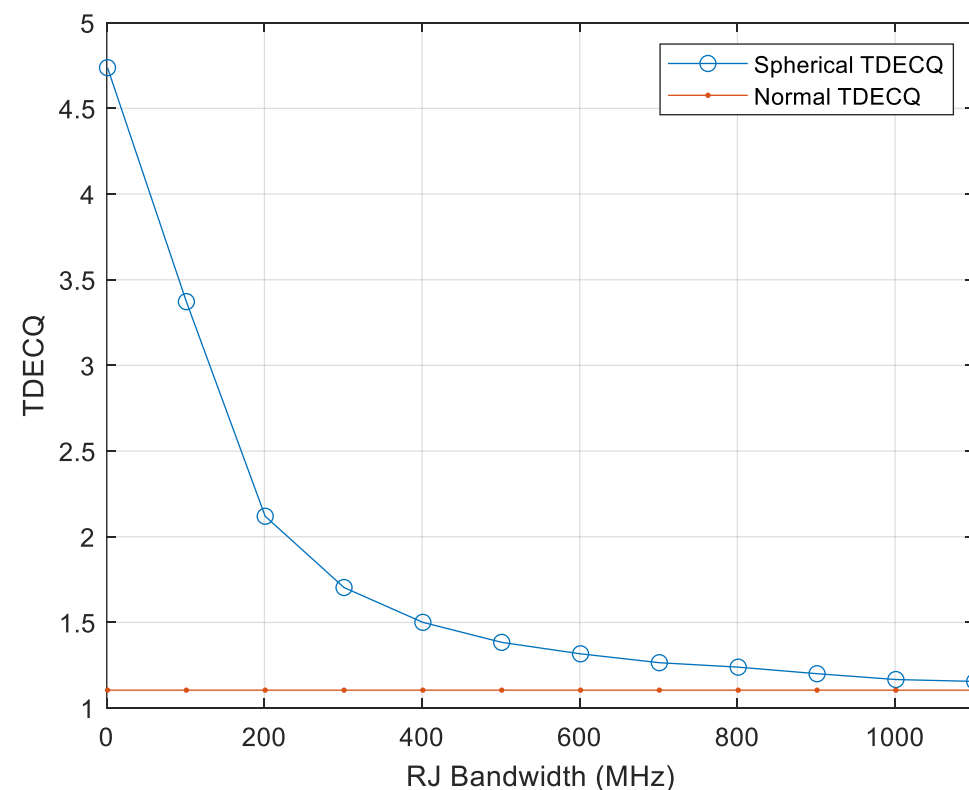
## Two Different Waveforms



# Bandlimited RJ

## Simulated Test Case

- [ran\\_3dj\\_02a\\_2407](#) and [oif2024.449.02](#) demonstrated receivers with performance issue related to bandlimited random jitter
- Using a captured real signal and applied 400fs of bandlimited RJ in MATLAB
- Comparison of Normal TDECQ and Spherical TDECQ.
- Spherical TDECQ is sensitive to the lower frequency bandlimited RJ
- NOTE: On a Sampling Oscilloscope, the spectrum of the RJ will be aliased, and will appear uncorrelated.



# Thank you